FRONTGRADE

DATASHEET

UT24C407

Certus™-NX-RT FPGA

12/13/2024

Version #: 2.0.7

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Acronyms in This Document

A list of acronyms used in this document.

Acronym	Definition
AES	Advanced Encryption Standard
BGA	Ball Grid Array
CDR	Clock and Data Recovery
CRC	Cycle Redundancy Code
DCC	Dynamic Clock Control
DCS	Dynamic Clock Select
DDR	Double Data Rate
DLL	Delay Locked Loops
DSP	Digital Signal Processing
EBR	Embedded Block RAM
ECC	Error Correction Coding
ECDSA	Elliptic Curve Digital Signature Algorithm
ECLK	Edge Clock
FFT	Fast Fourier Transforms
FIFO	First In First Out
FIR	Finite Impulse Response
LC	Logic Cell
LRAM	Large RAM
LVCMOS	Low-Voltage Complementary Metal Oxide Semiconductor
LVDS	Low-Voltage Differential Signaling
LVPECL	Low Voltage Positive Emitter Coupled Logic
LVTTL	Low Voltage Transistor-Transistor Logic
LUT	Look Up Table
MLVDS	Multipoint Low-Voltage Differential Signaling
PCI	Peripheral Component Interconnect
PCS	Physical Coding Sublayer
PCLK	Primary Clock
PDPR	Pseudo Dual Port RAM
PFU	Programmable Functional Unit
PIC	Programmable I/O Cells
PLL	Phase Locked Loops
POR	Power On Reset
SCI	SerDes Client Interface
SER	Soft Error Rate
SEU	Single Event Upset
SLVS	Scalable Low-Voltage Signaling
SPI	Serial Peripheral Interface
SPR	Single Port RAM
SRAM	Static Random-Access Memory
TAP	Test Access Port
TDM	Time Division Multiplexing

General Description

The Certus™-NX-RT family of low-power general purpose FPGAs can be used in a wide range of applications across multiple markets and are optimized for bridging and processing needs in Edge applications. It is built on the Lattice Nexus™ FPGA platform, using low-power 28 nm FD-SOI technology. It combines the extreme flexibility of an FPGA with the low power and high reliability (due to extremely low SER) of FD-SOI technology and offer small footprint package options with a high amount of I/O per mm2. Design security features such as AES-256 encryption and ECDSA authentication are also supported.

Certus™-NX-RT supports a variety of interfaces including PCI Express (Gen1, Gen2), SGMII (Gigabit Ethernet), LVDS, LVCMOS (0.9–3.3 V), and more.

Processing features of Certus™-NX-RT include 39k Logic Cells, 56 multipliers (18 × 18), 2.5 Mb of embedded memory (consisting of EBR and LRAM blocks), distributed memory, DRAM interfaces (supporting DDR3, DDR3L, LPDDR2, and LPDDR3 up to 1066 Mbps × 16 bits data width).

Certus™-NX-RT FPGAs support fast configuration of the reconfigurable SRAM-based logic fabric, and ultra-fast configuration (under 3 ms) of its programmable sysI/O™. In addition to the high reliability inherent to FD-SOI technology (due to its extremely low SER), active reliability features such as built-in frame-based SED/SEC (for SRAM-based logic fabric), and ECC (for EBR and LRAM) are also supported. Dual 12-bit ADCs are available on-chip for system monitoring functions.

Lattice Radiant™ design software allows large complex user designs to be efficiently implemented in the Certus™-NX-RT FPGA family. Synthesis library support for Certus™-NX-RT devices is available for popular logic synthesis tools. Radiant tools use the synthesis tool output along with constraints from floor planning tools to place and route the user design for the Certus™-NX-RT device. The tools extract timing from the routing, and back-annotate it into the design for timing verification.

Lattice provides many pre-engineered IP (Intellectual Property) modules for the Certus™-NX-RT family. By using these configurable soft IP cores as standardized blocks, you are free to concentrate on the unique aspects of your design, increasing your productivity.

Features

- Programmable architecture
  - 39k logic cells
  - 56 sysDSP™ 18 × 18 multipliers
  - 2.5 Mb of embedded memory (EBR, LRAM)
  - 192 programmable sysI/O (High Performance and Wide Range I/O)
- Programmable sysI/O supports wide variety of interfaces
  - High Performance (HP) I/O on bottom I/O banks
    - Supports up to 1.8 V V_{CCIO}
    - Mixed voltage support (1.0 V, 1.2 V, 1.5 V, 1.8 V)
    - High-speed differential up to 1.5 Gbps
    - Supports LVDS, Soft D-PHY (Tx/Rx), LVDS 7:1 (Tx/Rx), SLVS (Tx/Rx), subLVDS (Rx)
    - Supports SGMII (Gb Ethernet)
    - Two channels (Tx/Rx) at 1.25 Gbps
    - Dedicated DDR3/DDR3L and LPDDR2/LPDDR3 memory support with DQS logic, up to 1066 Mbps data-rate and x16 bits data width
  - Wide Range (WR) I/O on left, right, and top I/O Banks
    - Supports up to 3.3 V V_{CCIO}
    - Mixed voltage support (1.2 V, 1.5 V, 1.8 V, 2.5 V, 3.3 V)
    - Programmable slew rate (slow, med, fast)
    - Controlled impedance mode
    - Emulated LVDS support
    - Hot Socketing Support
- Power modes – Low Power versus High-Performance
  - User selectable
  - Low-Power mode for power and/or thermal challenges
  - High-Performance mode for faster processing
- Two channels of CDR up to 1.25 Gbps to support SGMII on HP I/O
  - CDR for RX
  - 10b/8b decoding
  - Independent Loss of Lock (LOL) detector for each CDR block
- sysCLOCK™ analog PLLs
  - Three in 39k LC device
  - Six outputs per PLL
  - Fractional N
  - Programmable and dynamic phase control
- sysDSP enhanced DSP blocks
  - Hardened pre-adder
  - Dynamic shift for AI/ML support
  - Four 18 × 18, eight 9 × 9, two 18 × 36, or 36 × 36 multipliers per block
  - Advanced 18 × 36, two 18 × 18, or four 8 × 8 MACs per block
- Flexible memory resources

  - — Up to 1.5 Mb sysMEM™ Embedded Block RAM
  - — Programmable width
  - — ECC
  - — FIFO
  - — 240k bits distributed RAM
  - — Large RAM Blocks
    - 0.5 Mbits per block
    - Up to five blocks (2.5 Mb total) per device
- SerDes – PCIe Gen2 x1 channel (Tx/Rx) hard IP in 39k LC device
  - — PCIe hard IP supports
    - Gen1 and Gen2
    - Endpoint and Root complex
    - Multi-function up to four functions
    - x1 lane
- Internal bus interface support
  - — APB control bus
  - — AHB-Lite for data bus
  - — AXI4-streaming
- Configuration – Fast, Secure
  - — SPI – x1, x2, x4 up to 150 MHz
    - Master and Slave SPI support
  - — JTAG
  - — I^2C and I3C
  - — Ultrafast I/O configuration for instant-on support (under 3 ms)
  - — Less than 15 ms full device configuration for UT24C407
- Cryptographic engine
  - — Bitstream encryption – using AES-256
  - — Bitstream authentication – using ECDSA
  - — Hashing algorithms – SHA, HMAC
  - — True Random Number Generator
  - — AES 128/256 Encryption
- Single Event Upset (SEU) Mitigation Support
  - — Extremely low Soft Error Rate (SER) due to FD-SOI technology
  - — Soft Error Detect – Embedded hard macro
  - — Soft Error Correction – Without stopping user operation
  - — Soft Error Injection – Emulate SEU event to debug system error handling
- Dual ADCs – 1MSPS, 12-bit SAR, with Simultaneous Sampling
  - — Three Continuous-time Comparators
- System level support
  - — IEEE 1149.1 and IEEE 1532 compliant
  - — Reveal Logic Analyzer
  - — On-chip oscillator for initialization
  - — 1.0 V core power supply

Table 1.1. Certus™-NX-RT Family Selection Guide

Device	UT24C407
Logic Cells*	39k
Embedded Memory (EBR) Blocks (18 kb)	84
Embedded Memory (EBR) Bits (kb)	1,512
Distributed RAM Bits (kb)	240
Large Memory (LRAM) Blocks	2
Large Memory (LRAM) Bits (kb) (512 kbits each)	1024
18 X 18 Multipliers	56
ADC Blocks	2
450 MHz High Frequency Oscillator	1
128 kHz Low Power Oscillator	1
GPLL	3
PCIe Gen2 Hard IP	1
Packages (Size, Ball Pitch)	Wide Range (WR) GPIO (Top/Left/Right Banks)/ High Performance (HP) GPIO (Bottom Banks)/ADC/5G PCIe Lane
caBGA256 (14 mm × 14 mm, 0.8 mm)	111/74/6/1

***Note:** Logic Cells = LUTs × 1.2 effectiveness.

Part-Specific Information

Each Certus™-NX-RT device contains a unique (per device) TraceID that can be used for tracking purposes or for IP security applications. Customers can provide the TraceID to the Frontgrade factory for part-specific information. Customers can learn more about the various methods to read the TraceID in section Trace ID, and in the associated application note (FPGA-TN-02084).

Architecture

Overview

Each Certus™-NX-RT device contains an array of logic blocks surrounded by Programmable I/O Cells (PIC). Interspersed between the rows of logic blocks are rows of sysMEM Embedded Block RAM (EBR) and rows of sysDSP Digital Signal Processing blocks, as shown in Figure 2.1. The UT24C407 devices have two rows of DSP blocks and contain three rows of sysMEM EBR blocks. In addition, UT24C407 devices include two Large SRAM blocks. The sysMEM EBR blocks are large, dedicated 18 kb fast memory blocks and have built-in ECC and FIFO support. Each sysMEM block can be configured to a single, pseudo dual or true dual port memory in a variety of depths and widths as RAM or ROM. Each DSP block supports a variety of multiplier and adder configurations with one 108-bit or two 54-bit accumulators supported, which are the building blocks for complex signal processing capabilities.

Each PIC block encompasses two PIO (PIO pairs) with their respective sysI/O buffers. The sysI/O buffers of the Certus™-NX-RT devices are arranged in seven banks allowing the implementation of a wide variety of I/O standards. The Wide Range (WR) I/O banks that are located on the top, left and right sides of the device provide flexible ranges of general purpose I/O configurations up to 3.3 V V_{CCIOs}. The banks located on the bottom side of the device are dedicated to High Performance (HP) interfaces such as LVDS, MIPI, DDR3, LPDDR2, and LPDDR3 supporting up to 1.8 V VCCIOs.

The Programmable Functional Unit (PFU) contains the building blocks for logic, arithmetic, RAM and ROM functions. The PFU block is optimized for flexibility, allowing complex designs to be implemented quickly and efficiently. Logic Blocks are arranged in a two-dimensional array. The registers in the PFU and sysI/O blocks in Certus™-NX-RT devices can be configured to be SET or RESET. After power up and device configuration, it enters into user mode with these registers SET/RESET according to the user design, allowing the device to power up in a known state for predictable system function.

In addition, Certus™-NX-RT devices provide various system level hard IP functional and interface blocks such as PCIe, I^2C, SGMII/CDR, and ADC blocks. The PCIe hard IP supports PCIe Generation 2.0. Certus™-NX-RT devices also provide security features to help protect user designs and deliver more robust reliability by offering enhanced frame-based SED/SEC functions.

Other blocks provided include PLLs, DLLs, and configuration functions. The PLL and DLL blocks are located at the corners of each device. Certus™-NX-RT devices also include the Lattice Memory Mapped Interface (LMMI) which is a Lattice standard to support simple read and write operations to control internal IP.

Every device in the family has a JTAG port. This family also provides an on-chip oscillator and soft error detect capability. The Certus™-NX-RT devices use 1.0 V as their core voltage.

Figure 2.1. Simplified Block Diagram, UT24C407 Device (Top Level)

PFU Blocks

The core of the Certus™-NX-RT device consists of PFU blocks. Each PFU block consists of four interconnected slices numbered 0–3 as shown in Figure 2.2. Each slice contains two LUTs. All the interconnections to and from PFU blocks are from routing.

The PFU block can be used to perform Logical, Arithmetic, RAM or ROM functions. Table 2.1 shows the functions each slice can perform in either Distributed SRAM or non-Distributed SRAM modes.

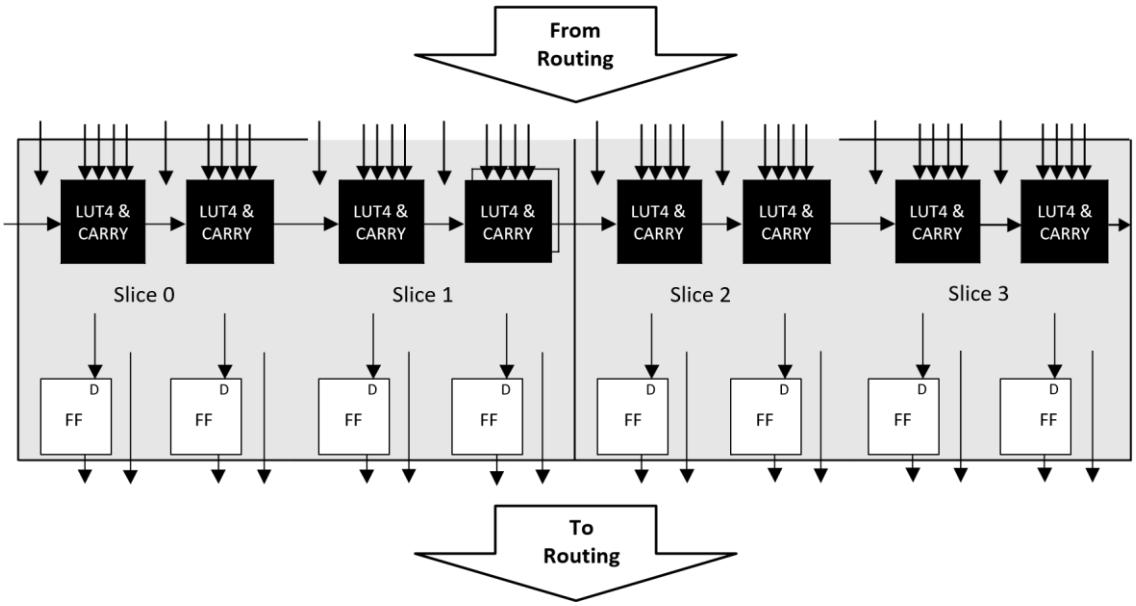


Figure 2.2. PFU Diagram

Slice

Each slice contains two LUT4s feeding two registers. In Distributed SRAM mode, Slice 0 and Slice 1 are configured as distributed memory, and Slice 2 is not available as it is used to support Slice 0 and Slice 1 while Slice 3 is available as Logic or ROM. Table 2.1 shows the capability of the slices along with the operation modes they enable. In addition, each Slice contains logic that allows the LUTs to be combined to perform a LUT5 function. There is control logic to perform set/reset functions (programmable as synchronous/ asynchronous), clock select, chip-select, and wider RAM/ROM functions.

Table 2.1. Resources and Modes Available per Slice

Slice	PFU (Used as Distributed SRAM)		PFU (Not used as Distributed SRAM)	
	Resources	Modes	Resources	Modes
Slice 0	2 LUT4s and 2 Registers	RAM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 1	2 LUT4s and 2 Registers	RAM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 2	2 LUT4s and 2 Registers	RAM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 3	2 LUT4s and 2 Registers	Logic, Ripple, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM

Figure 2.3 shows an overview of the internal logic of the slice. The registers in the slice can be configured for positive/negative edge trigger.

Each slice has 17 input signals: 16 signals from routing and one from the carry-chain (from the adjacent slice or PFU). Three of them are used for FF control and shared between two slices (0/1 or 2/3). There are five outputs: four to routing and one to carry-chain (to the adjacent PFU). Table 2.2 and Figure 2.3 list the signals associated with all the slices. Figure 2.4 shows the slice signals that support a LUT5 or two LUT5 functions. F0 can be configured to have a LUT4 or LUT5 output while F1 is for a LUT4 output.

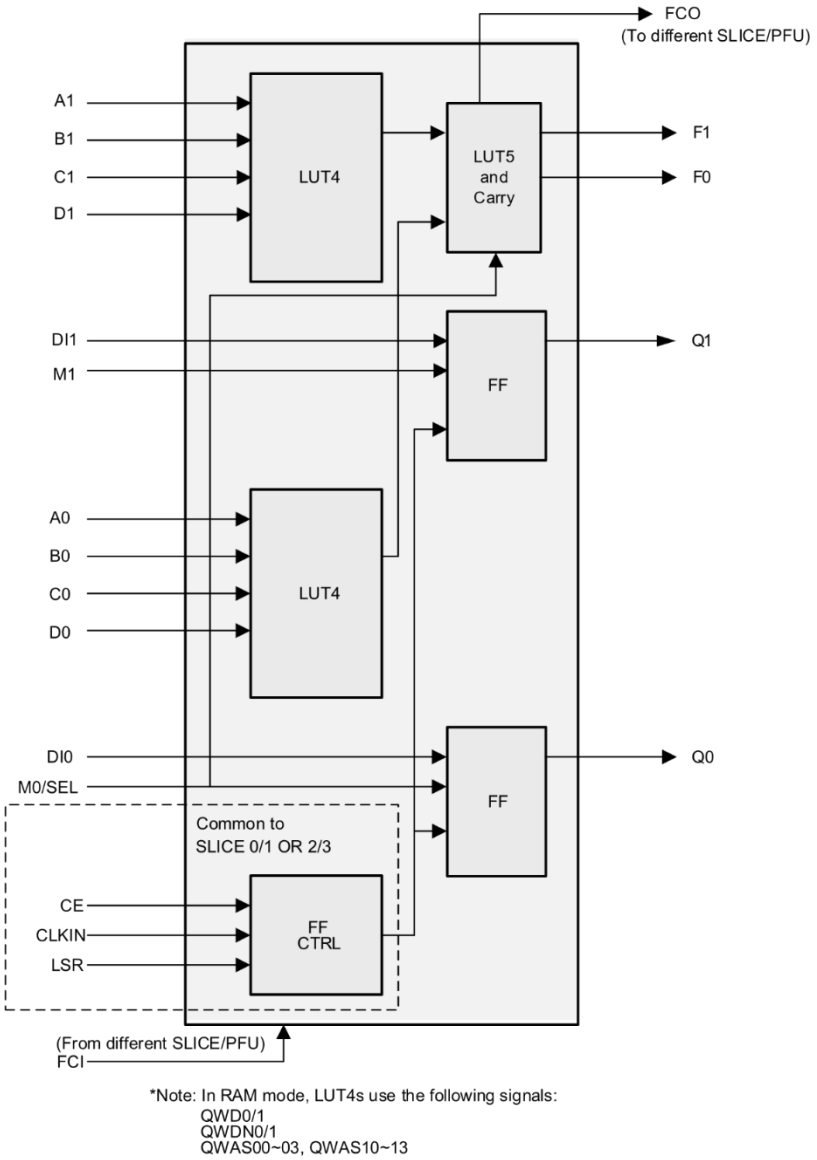


Figure 2.3. Slice Diagram

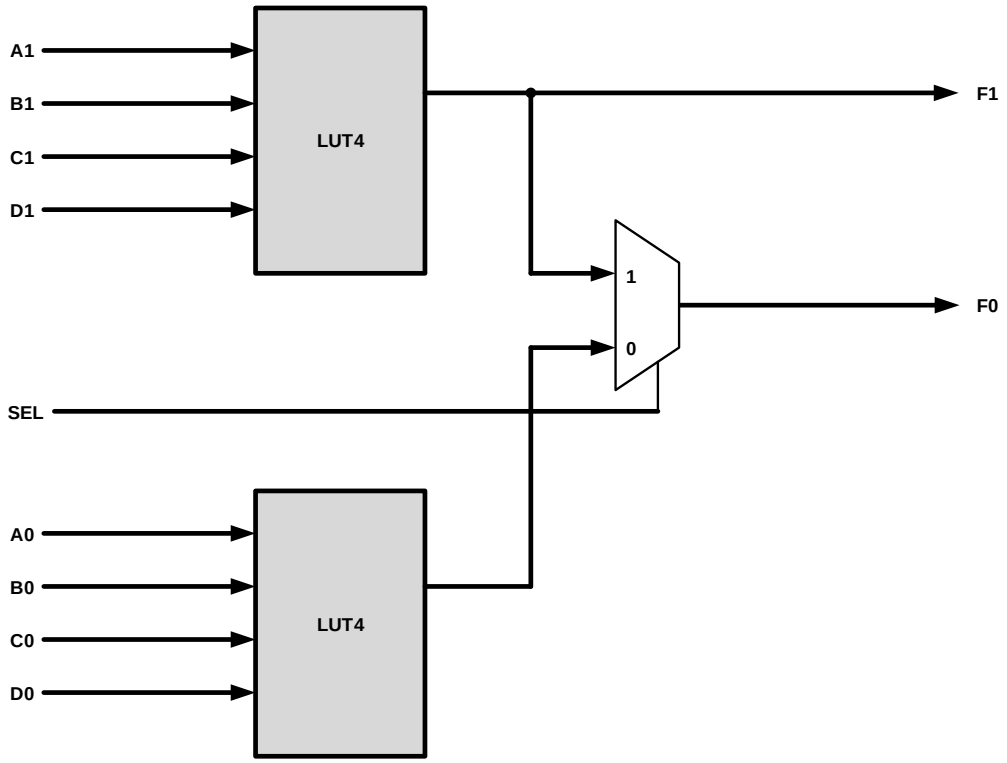


Figure 2.4. Slice configuration for LUT4 and LUT5

Table 2.2. Slice Signal Descriptions

Function	Type	Signal Names	Description
Input	Data signal	A0, B0, C0, D0	Inputs to LUT4
Input	Data signal	A1, B1, C1, D1	Inputs to LUT4
Input	Data signal	M0, M1	Direct input to FF from fabric
Input	Control signal	SEL	LUT5 mux control input
Input	Data signal	DI0, DI1	Inputs to FF from LUT4 F0/F1 outputs
Input	Control signal	CE	Clock Enable
Input	Control signal	LSR	Local Set/Reset
Input	Control signal	CLKIN	System Clock
Input	Inter-PFU signal	FCI	Fast Carry-in
Output	Data signals	F0	LUT4/LUT5 output signal
Output	Data signals	F1	LUT4 output signal
Output	Data signals	Q0, Q1	Register outputs
Output	Inter-PFU signal	FCO	Fast carry chain output

Note: See Figure 2.3 for connection details.

Modes of Operation

Slices 0-2 have up to four potential modes of operation: Logic, Ripple, RAM and ROM. Slice 3 is not needed for RAM mode, it can be used in Logic, Ripple, or ROM modes.

Logic Mode

In this mode, the LUTs in each slice are configured as 4-input combinatorial lookup tables. A LUT4 can have 16 possible input combinations. Any four input logic functions can be generated by programming this lookup table. Since there are two LUT4s per slice, a LUT5 can be constructed within one slice.

Ripple Mode

Ripple mode supports the efficient implementation of small arithmetic functions. In ripple mode, the following functions can be implemented by each slice:

- Addition 2-bit
- Subtraction 2-bit
- Add/Subtract 2-bit using dynamic control
- Up counter 2-bit
- Down counter 2-bit
- Up/Down counter with asynchronous clear 2-bit using dynamic control
- Up/Down counter with preload (sync) 2-bit using dynamic control
- Comparator functions of A and B inputs 2-bit
  - A greater-than-or-equal-to B
  - A not-equal-to B
  - A less-than-or-equal-to B
- Up/Down counter with A greater-than-or-equal-to B comparator 2-bit using dynamic control
- Up/Down counter with A less-than-or-equal-to B comparator 2-bit using dynamic control
- Multiplier support Ai×Bj+1 + Ai+1×Bj in one logic cell with 2 logic cells per slice
- Serial divider 2-bit mantissa, shift 1bit/cycle
- Serial multiplier 2-bit, shift 1bit/cycle or 2bit/cycle

Ripple Mode includes an optional configuration that performs arithmetic using fast carry chain methods. In this configuration (also referred to as CCU2 mode) two additional signals, Carry Generate and Carry Propagate, are generated on a per slice basis to allow fast arithmetic functions to be constructed by concatenating Slices.

RAM Mode

In this mode, a 16 × 4-bit distributed single or pseudo dual port RAM can be constructed in one PFU using each LUT block in Slice 0 and Slice 1 as a 16 × 2-bit memory in each slice. Slice 2 is used to provide memory address and control signals. The Certus™-NX-RT devices support distributed memory initialization.

The Lattice design tools support the creation of a variety of different sized memories. Where appropriate, the software constructs these using distributed memory primitives that represent the capabilities of the PFU. Table 2.3 lists the number of slices required to implement different distributed RAM primitives. For more information about using RAM in Certus™-NX-RT devices, refer to Memory Usage Guide for Nexus Platform (FPGA-TN-02094).

Table 2.3. Number of Slices Required to Implement Distributed RAM

	SPR 16 X 4	PDPR 16 X 4
Number of slices	3	3

Note: SPR = Single Port RAM, PDPR = Pseudo Dual Port RAM

ROM Mode

ROM mode uses the LUT logic; hence, Slice 0 through Slice 3 can be used in ROM mode. Preloading is accomplished through the programming interface during PFU configuration. For more information, refer to Memory Usage Guide for Nexus Platform (FPGA-TN-02094).

Routing

There are many resources provided in the Certus™-NX-RT devices to route signals individually or as busses with related control signals. The routing resources consist of switching circuitry, buffers and metal interconnect (routing) segments.

The Certus™-NX-RT family has an enhanced routing architecture that produces a compact design. The Radiant software tool suites take the output of the synthesis tool and places and routes the design.

Clocking Structure

2.3.2. The Certus™-NX-RT clocking structure consists of clock synthesis blocks (PLLs), balanced clock tree networks (PCLK & ECLK), and efficient clock logic modules: Clock Dividers (PCLKDIV and ECLKDIV), Dynamic Clock Selection (DCS), Dynamic Clock Control (DCC), and DDRDLLs. Each of these functions is described as follows.

Global PLL

The Global PLLs (GPLL) provide the ability to synthesize clock frequencies. The devices in the Certus™-NX-RT family support two or three full-featured General Purpose GPLLs.

The architecture of the GPLL is shown in Figure 2.5. A description of the GPLL functionality follows.

REFCLK is the reference frequency input to the PLL and its source can come from external CLK inputs or from internal routing. The CLKI input feeds into the input Clock Divider block.

CLKFB is the feedback signal to the GPLL which can come from a path internal to the PLL or from FPGA routing. The feedback divider is used to multiply the reference frequency and thus synthesize a higher or lower frequency clock output.

The PLL has six clock outputs CLKOP, CLKOS, CLKOS2, CLKOS3, CLKOS4, and CLKOS5. Each output has its own output divider, thus allowing the GPLL to generate different frequencies for each output. The output dividers can have a value from 1 to 128. Each GPLL output can be used to drive the primary clock or edge clock networks.

The setup and hold times of the device can be improved by programming a phase shift into the output clocks which advances or delays the output clock with reference to the un-shifted output clock. This phase shift can be either programmed during configuration or can be adjusted dynamically using the DIRSEL, DIR, DYNROTATE, and LOADREG ports.

The LOCK signal is asserted when the GPLL determines it has achieved lock and de-asserted if a loss of lock is detected. The lock signal is asynchronous to the PLL clock outputs

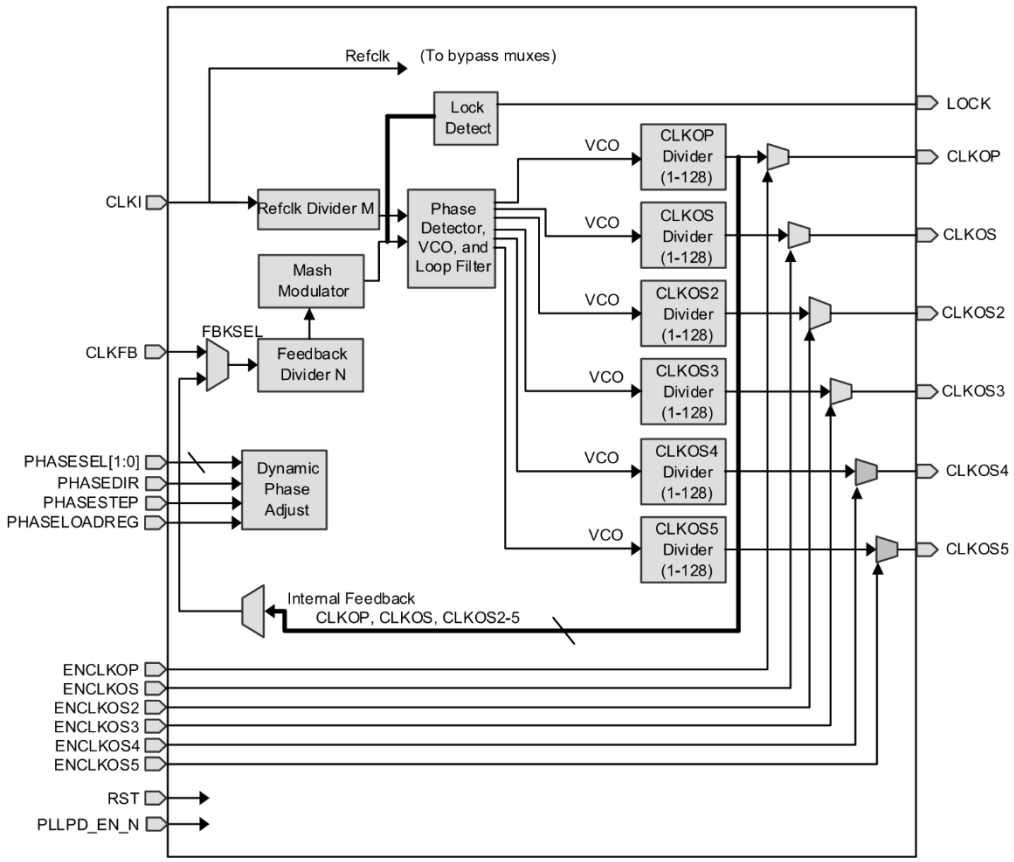


Figure 2.5. General Purpose PLL Diagram

For more details on the PLL, you can refer to the sysCLOCK PLL Design and Usage Guide for Nexus Platform (FPGA-TN-02095).

Clock Distribution Network

There are two main clock distribution networks for any member of the Certus™-NX-RT product family, namely Primary Clock (PCLK) and Edge Clock (ECLK). These clock networks can be driven from many different sources, such as Clock Pins, PLL outputs, DLLDEL outputs, Clock Divider outputs, SerDes/PCS clocks and user logic. There are clock divider blocks (ECLKDIV and PCLKDIV) to provide a slower clock from these clock sources.

Certus™-NX-RT supports glitchless Dynamic Clock Control (DCC) for the PCLK Clock to save dynamic power. There are also Dynamic Clock Selection logic to allow glitchless selection between two clocks for the PCLK network (DCS).

An overview of the Clocking Network is shown in Figure 2.6 for the Certus™-NX-RT device. The shaded blocks (PCIe and upper left PLL) are not available in the 17k Logic Cell Device.

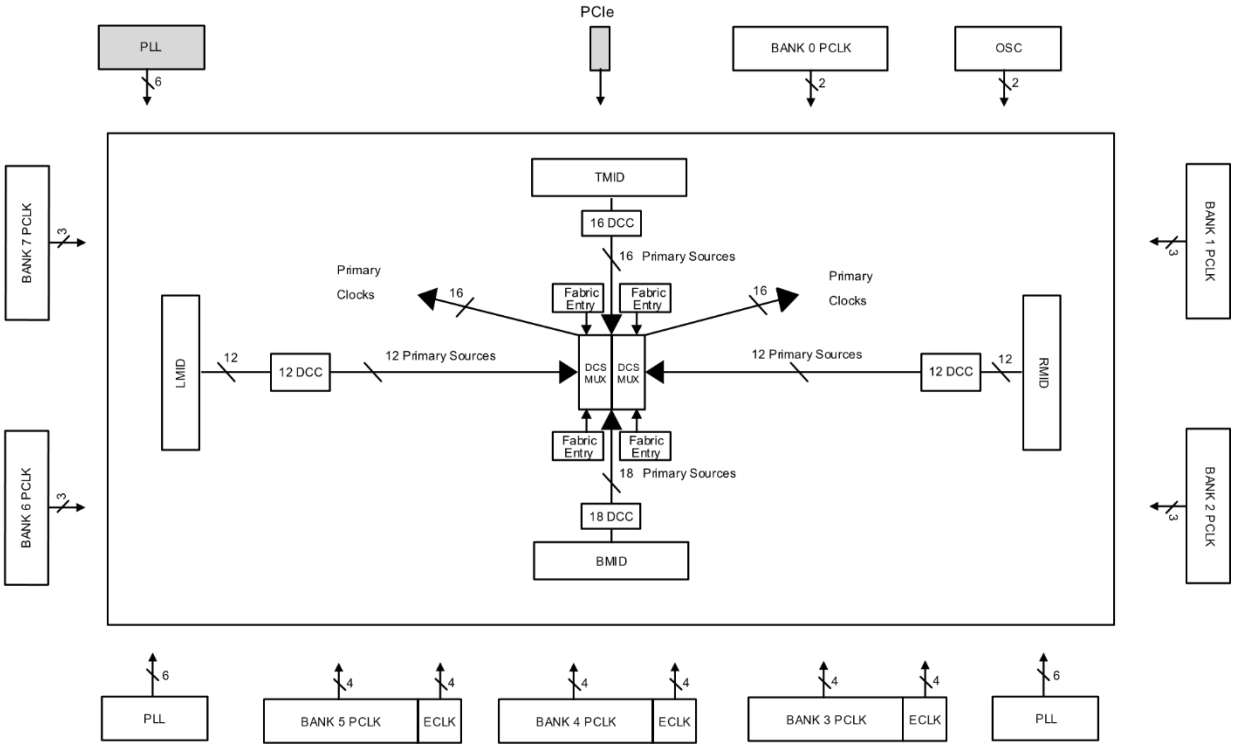


Figure 2.6 Clocking

Primary Clocks

The Certus™-NX-RT device family provides low-skew, high fan-out clock distribution to all synchronous elements in the FPGA fabric through the Primary Clock Network. The Certus™-NX-RT PCLK clock network is a balanced clock structure which is designed to minimize the clock skew across all destinations in the FPGA core.

The primary clock network is divided into two clock domains depending on the device density. Each of these domains has 16 clocks that can be distributed to the fabric in the domain.

The Lattice Radiant software can automatically route each clock to one of the domains up to a maximum of 16 clocks per domain. You can change how the clocks are routed by specifying a preference in the Lattice Radiant software to locate the clock to a specific domain. The Certus™-NX-RT device provides you with a maximum of 64 unique clock input sources that can be routed to the primary Clock network.

Primary clock sources are:

- Dedicated clock input pins
- PLL outputs
- PCLKDIV, ECLKDIV outputs
- Internal FPGA fabric entries (with minimum general routing)
- SGMII-CDR, PCIe clocks
- OSC clock

These sources are routed to each of four clock switches called a Mid Mux (LMID, RMID, TMID, BMID). The outputs of the Mid MUX are routed to the center of the FPGA where additional clock switches (DSC_CMUX) are used to route the primary clock sources to primary clock distribution to the Certus™-NX-RT fabric. These routing muxs are shown in **Error! Reference s ource not found.**. There are potentially 64 unique clock domains that can be used in the largest Certus™-NX-RT Device. For more information about the primary clock tree and connections, refer to sysCLOCK PLL Design and Usage Guide for Nexus Platform (FPGA-TN-02095).

Edge Clock

Certus™-NX-RT FPGAs have a number of high-speed edge clocks that are intended for use with the PIO in the implementation of high-speed interfaces. There are four (4) ECLK networks per bank I/O on the Bottom side of the device. The Edge clock network is powered by a separate power domain (to reduce power noise injection from the core and reduce overall noise induced jitter) while controlled by the same logic that gates the FPGA core and PCLK domains for power management.

Each Edge Clock can be sourced from the following:

- Dedicated PIO Clock input pins (PCLK)
- DLLDEL output (PIO Clock delayed by 90°)
- PLL outputs (CLKOP, CLKOS, CLKOS2, CLKOS3, CLKOS4, and CLKOS5)
- Internal Nodes

Figure 2.7 illustrates the various ECLK sources. Bank 3 is shown in the example. Bank 4 and Bank 5 are similar.

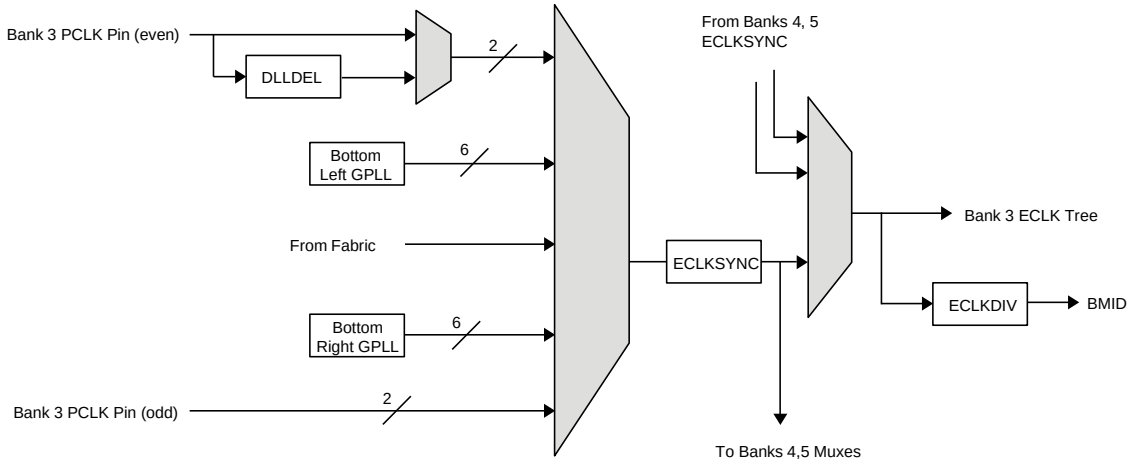


Figure 2.7. Edge Clock Sources per Bank

The edge clocks have low injection delay and low skew. They are typically used for DDR Memory or Generic DDR interfaces. For detailed information on Edge Clock connections, refer to sysCLOCK PLL Design and Usage Guide for Nexus Platform (FPGA-TN-02095).

Clock Dividers

The Certus™-NX-RT devices have two distinct types of clock divider, Primary and Edge. There are from one (1) to eight (8) Primary Clock Divider (PCLKDIV) and which are located in the DCS_CMUX block(s) at the center of the device. There are twelve (12) ECLKDIV dividers per device, locate near the bottom high-speed I/O banks.

The PCLKDIV supports ÷2, ÷4, ÷8, ÷16, ÷32, ÷64, ÷128, and ÷1 (bypass) operation. The PCLKDIV is fed from a DCSMUX within the DCS_CMUX block. The clock divider output drives one input of the DCS Dynamic Clock Select within the DSC_CMUX block. The Reset (RST) control signal is asynchronous and forces all outputs to low. The divider output starts at next cycle after the reset is synchronously released. The PCLKDIV is shown in context in Figure 2.8.

The ECLKDIV is intended to generate a slower-speed system clock from a high-speed edge clock. The block operates in a ÷2, ÷3.5, ÷4, or ÷5 mode and maintains a known phase relationship between the divided down clock and the high-speed clock based on the release of its reset signal. The ECLKDIV can be fed from selected PLL outputs, external primary clock pins (with or without DLLDEL Delay) or from routing. The clock divider outputs feed into the Bottom Mid-mux (BMID). The Reset (RST) control signal is asynchronous and forces all outputs to low. The divider output starts at next cycle after the reset is synchronously released.

The ECLKDIV block is shown in context in Figure 2.7. For further information on clock dividers, refer to sysCLOCK PLL Design and Usage Guide for Nexus Platform (FPGA-TN-02095).

Clock Center Multiplexor Blocks

All clock sources are selected and combined for primary clock routing through the Dynamic Clock Selector Center Multiplexer logic (DCS_CMUX). There are one (1) or two (2) DCS_CMUX blocks per device. Each DCS_CMUX block contains 2 DCSMUX blocks, 1 PCLKDIV, 1 DCS block, and 1 or 2 CMUX blocks. See Figure 2.8 for a representative DCS_CMUX block diagram.

The heart of the DCS_CMUX is the Center Multiplexer (CMUX) block. It can accept up to 64 feed clock sources (Mid-muxes (RMID, LMID, TMIC, BMID) and DCC) and to drive up to 16 primary clock trunk lines.

Up to two (2) clock inputs to the DCS_CMUX can be routed through a Dynamic Clock Select block then routed to the CMUX. One (1) input to the DCS can be optionally divided by the Primary Clock Divider (PCLKDIV). For more information about the DCS_CMUX, refer to sysCLOCK PLL Design and Usage Guide for Nexus Platform (FPGA-TN-02095).

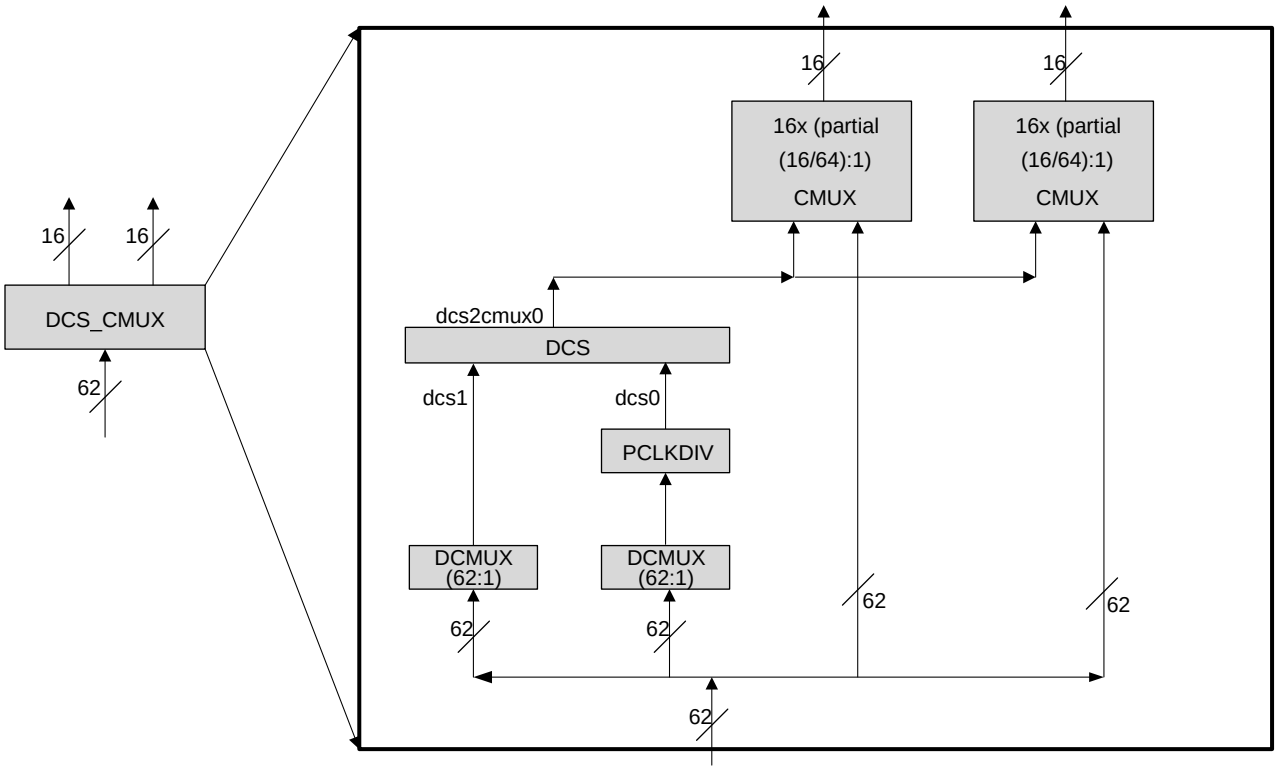


Figure 2.8. DCS_CMUX Diagram

Dynamic Clock Select

The Dynamic Clock Select (DCS) is a smart multiplexer function available in the primary clock routing. It switches between two independent input clock sources. Depending on the operational mode, it switches between two (2) independent input clock sources either with or without any glitches. This is achieved regardless of when the select signal is toggled. Both input clocks must be running to achieve a functioning glitchless DCS output clock, but running clocks are not required when used as a non-glitchless normal clock multiplexer.

There are one (1) or two (2) DCS blocks per device that feed all clock domains. The DCS blocks are located in the DCS_MUX block. The inputs to the DCS blocks come from MIDMUX outputs and user logic clocks via DCC elements. The DCS elements are located at the center of the PLC array core. The output of the DCS is connected to the inputs of Primary Clock Center MUXs (CMUX).

Figure 2.9 shows the timing waveforms of the default DCS operating mode. The DCS block can be programmed to other modes. For more information about the DCS, refer to sysCLOCK PLL Design and Usage Guide for Nexus Platform (FPGA-TN-02095).

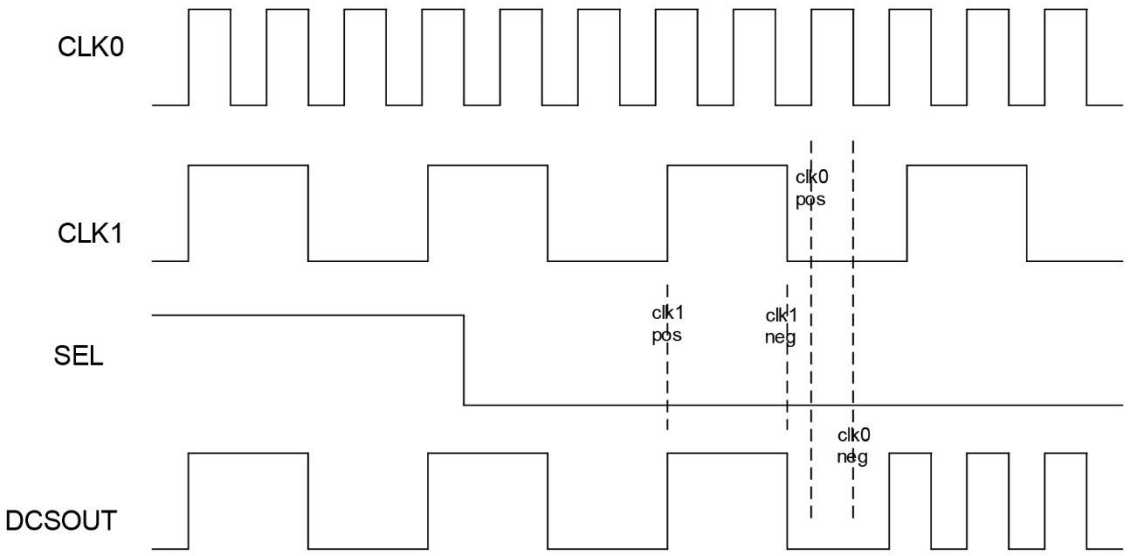


Figure 2.9. DCS Waveforms

Dynamic Clock Control

The Dynamic Clock Control (DCC), Domain Clock enable/disable feature allows internal logic control of the domain primary clock network. When a clock network is disabled, the clock signal is static and does not toggle. All the logic fed by that clock does not toggle, reducing the overall power consumption of the device. The disable function is glitchless and does not increase the clock latency to the primary clock network.

Four additional DCC elements control the clock inputs from the Certus™-NX-RT domain logic to the Center MUX elements (DSC_CMUX).

This DCC controls the clock sources from the Primary CLOCK MIDMUX before they are fed to the Primary Center MUXs that drive the domain clock network. For more information about the DCC, refer to sysCLOCK PLL Design and Usage Guide for Nexus Platform (FPGA-TN-02095).

DDRDLL

Certus™-NX-RT has two identical DDRDLL blocks, located in the lower left and lower right corners of the device. Each DDRDLL (master DLL block) can generate a 9 bit phase shift value corresponding to a 90-degree phase shift of the reference clock input and provide this value to every DQS block and DLLDEL slave delay element. The reference clock can be either from a PLL, or an input pin. The DQSBUF uses this value to control the delay of the DQS inputs from a DDR memory interface to achieve a 90-degree shift in order to clock DQ inputs at the center of the data eye

The value is also sent to another slave DLL, DLLDEL, that takes a primary clock input and generates a 90-degree shifted clock output to drive the clocking structure. This is useful in an edge-aligned Generic DDR interface, where 90-degree clocking needs to be created. Not all primary clock inputs have associated DLLDEL control. Figure 2.10 shows DDRDLL connectivity to a DLLDEL block (connectivity to DQSBUF blocks is similar).

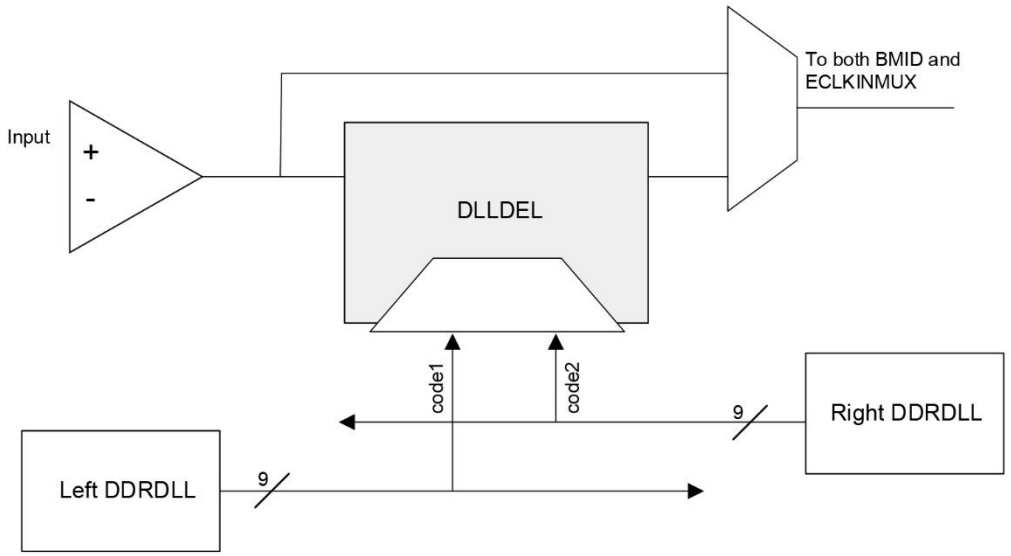


Figure 2.10. DLLDEL Functional Diagram

Each DDRDLL can generate a delay value based on the reference clock frequency. The slave DLL (DQSBUF and DLLDEL) use the value (code) to either create phase shifted inputs from the DDR memory or create a 90-degree shifted clock. Figure 2.11 shows the connections between the DDRDLL and the slave DLLs.

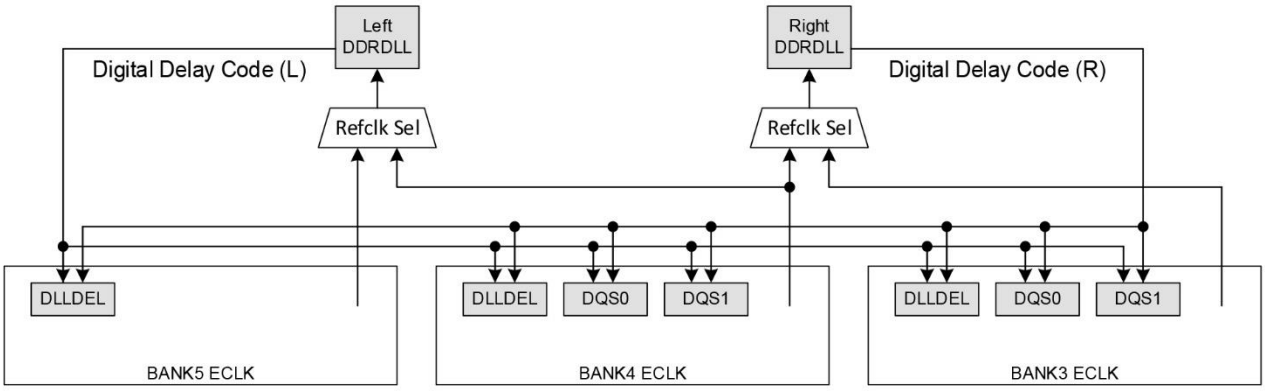


Figure 2.11. Certus™-NX-RT DDRDLL Architecture

SGMII Clock Data Recovery (CDR)

The Certus™-NX-RT device includes two hardened CDR (Clock & Data Recovery) components. The CDRs enable SGMII (Serial Gigabit Media Independent Interface) solutions. There are three main blocks in each CDR: the CDR, deserializer, and FIFO. Each CDR features two loops. The first loop is locked to the reference clock. Once locked, the loop switches to the data path loop where the CDR tracks the data signals to generate the correcting signals needed to achieve and maintain phase lock with the data. The data is then passed through a deserializer which deserialize the data to 10-bit parallel data. The 10-bit parallel data is then sent to the FIFO bridge, which allows the CDR to interface with the rest of the FPGA.

Figure 2.12 shows a block diagram of the SGMII CDR IP.

The two hardened blocks are located at the bottom left of the chip and uses the high speed I/O Bank 5 for the differential pair input. It is recommended that the reference clock should be entered through a GPIO that has connection to the PLL on the lower left corner as well.

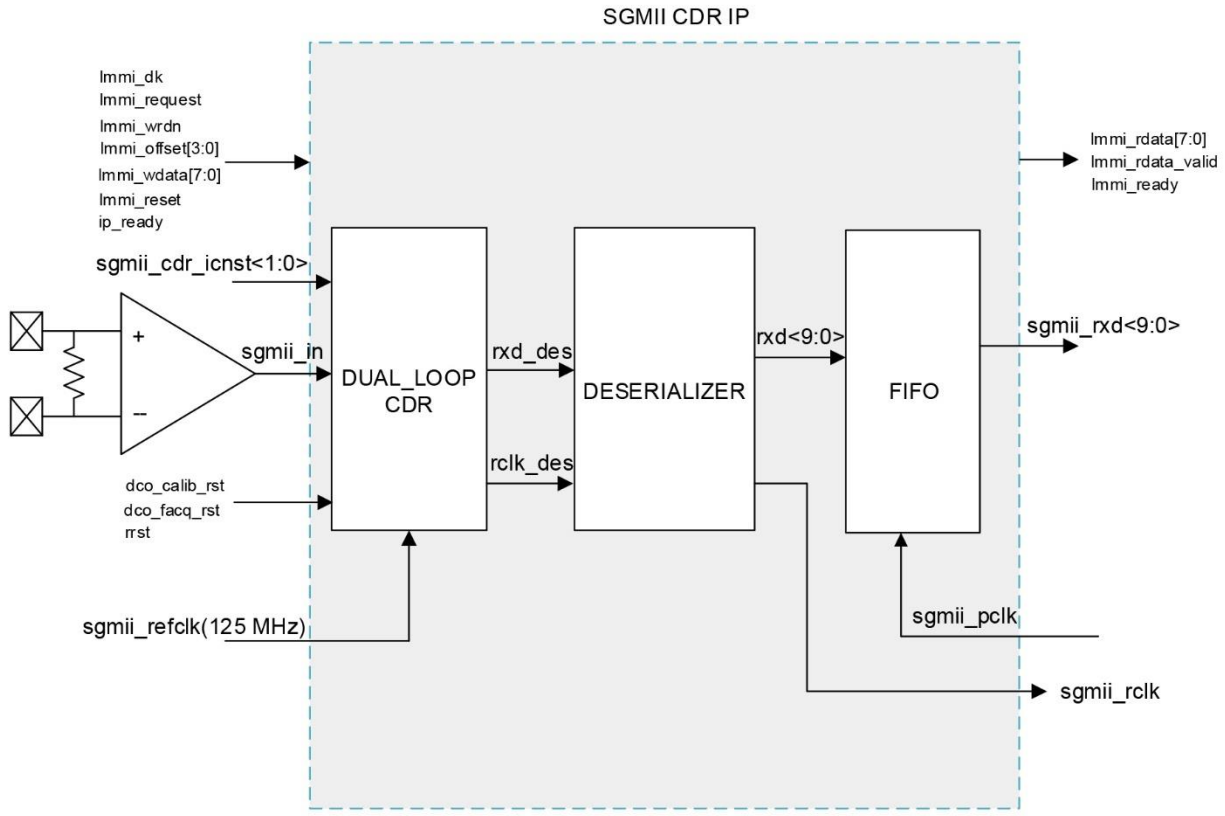


Figure 2.12. SGMII CDR IP

sysMEM Memory

The Certus™-NX-RT devices contain a number of sysMEM Embedded Block RAM (EBR). The EBR consists of an 18 kb RAM with memory core, dedicated input registers and output registers as well as optional pipeline registers at the outputs. Each EBR includes functionality to support true dual-port, pseudo dual-port, single-port RAM, ROM and built in FIFO. In Certus™-NX-RT, unused EBR blocks is powered down to minimize power consumption.

sysMEM Memory Block

The sysMEM block can implement single port, dual port or pseudo dual port memories. Each block can be used in a variety of depths and widths as listed in Table 2.4. FIFOs can be implemented using the built in read and write address counters and programmable full, almost full, empty and almost empty flags. The EBR block facilitates parity checking by supporting an optional parity bit for each data byte. EBR blocks provide byte-enable support for configurations with 18-bit and 36-bit data widths. For more information, refer to Memory Usage Guide for Nexus Platform (FPGA-TN-02094).

EBR also provides a build in ECC engine. The ECC engine supports a write data width of 32 bits and it can be cascaded for larger data widths such as x64. The ECC parity generator creates and stores parity data for each 32-bit word written. When a read operation is performed, it compares the data with its associated parity data and report back if any Single Event Upset (SEU) event has disturbed the data. Any single bit data disturb is automatically corrected at the data output. In addition, two dedicated error flags indicate if a single-bit or two-bit error has occurred.

Table 2.4. sysMEM Block Configurations

Memory Mode	Configurations
Single Port	16,384 × 1
	8,192 × 2
	4,096 × 4
	2,048 × 9
	1,024 × 18
	512 × 36
True Dual Port	16,384 × 1
	8,192 × 2
	4,096 × 4
	2,048 × 9
	1,024 × 18
Pseudo Dual Port	16,384 × 1
	8,192 × 2
	4,096 × 4
	2,048 × 9
	1,024 × 18
	512 × 36

Bus Size Matching

All of the multi-port memory modes support different widths on each of the ports (except ECC mode which only supports a write data width of 32 bits). The RAM bits are mapped LSB word 0 to MSB word 0, LSB word 1 to MSB word 1, and so on. Although the word size and number of words for each port varies, this mapping scheme applies to each port.

RAM Initialization and ROM Operation

If desired, the contents of the RAM can be pre-loaded during device configuration. By preloading the RAM block during the chip configuration cycle and disabling the write controls, the sysMEM block can also be utilized as a ROM.

Memory Cascading

Larger and deeper blocks of RAM can be created using EBR sysMEM Blocks. Typically, the Lattice design tools cascade memory transparently, based on specific design inputs.

Single, Dual and Pseudo-Dual Port Modes

In all the sysMEM RAM modes, the input data and address for the ports are registered at the input of the memory array. The output data of the memory is optionally registered at the output.

Memory Output Reset

The EBR utilizes latches at the A and B output ports. These latches can be reset asynchronously or synchronously. RSTA and RSTB are local signals, which reset the output latches associated with Port A and Port B, respectively. The Global Reset (GSRN) signal can reset both ports. The output data latches and associated resets for both ports are as shown in. Figure 2.13 The optional Pipeline Registers at the outputs of both ports are also reset in the same way.

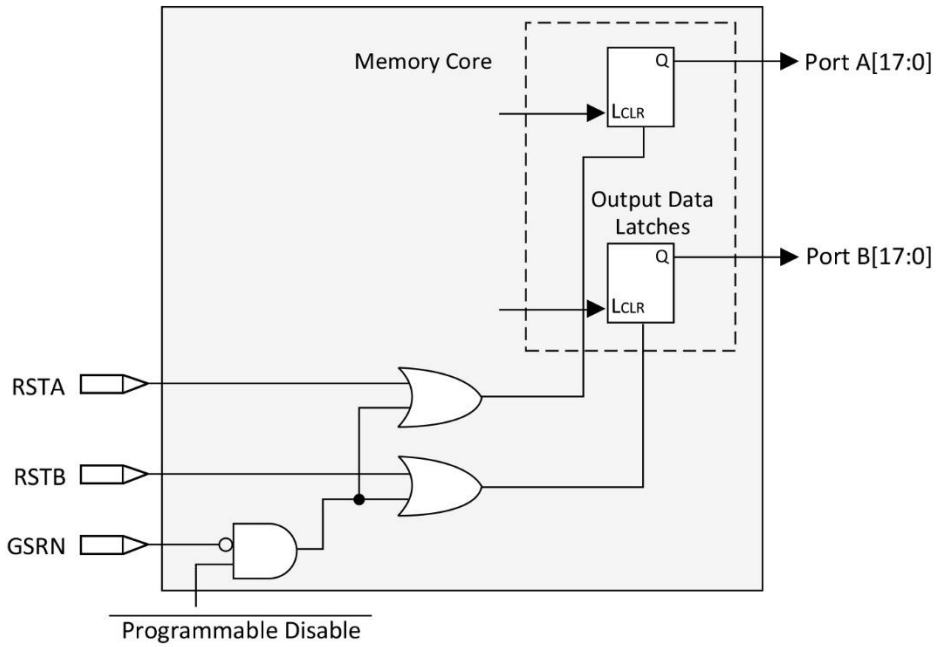


Figure 2.13. Memory Core Reset

For further information on the sysMEM EBR block, see the list of technical documentation in the

References section.

Large RAM

The Certus™-NX-RT device includes additional memory resources in the form of Large Random-Access Memory (LRAM) blocks.

The LRAM is designed to work as Single-Port RAM, Dual-Port RAM, Pseudo Dual-Port RAM, and ROM memories. It is meant to function as additional memory resources for you beyond what is available in the EBR and PFU.

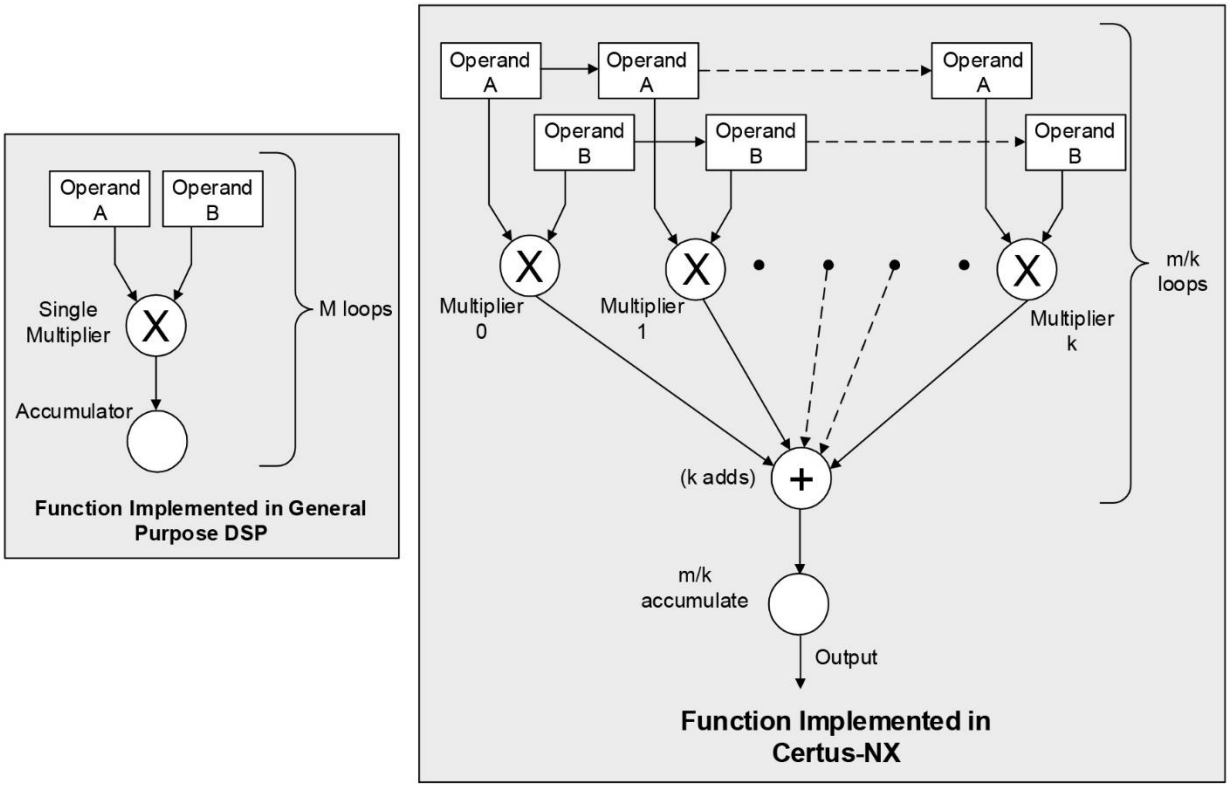
Each individual Large RAM block contains 0.5 Mbits or megabits of memory and has a programmable data width of up to 32 bits. Cascading Large RAM blocks allows data widths of up to 64 bits. Additionally, there is the ability to use either Error Correction Coding (ECC) or byte enable.

sysDSP

The Certus™-NX-RT family provides an enhanced sysDSP architecture, making it ideally suited for low-cost, high-performance Digital Signal Processing (DSP) applications. Typical functions used in these applications are Finite Impulse Response (FIR) filters, Fast Fourier Transforms (FFT) functions, Correlators, Reed-Solomon/Turbo/Convolution encoders and decoders. These complex signal processing functions use similar building blocks such as multiply-adders and multiply-accumulators.

sysDSP Approach Compared to General DSP

Conventional general-purpose DSP chips typically contain one to four (Multiply and Accumulate) MAC units with fixed data-width multipliers; this leads to limited parallelism and limited throughput. Their throughput is increased by higher clock speeds. In the Certus™-NX-RT device family, there are many DSP blocks that can be used to support different data widths.



This allows you to use highly parallel implementations of DSP functions. You can optimize DSP performance versus area by choosing appropriate levels of parallelism. Figure 2.14 compares the fully serial implementation to the mixed parallel and serial implementation.

Figure 2.14. Comparison of General DSP and Certus™-NX-RT Approaches

sysDSP Architecture Features

The Certus™-NX-RT sysDSP block contains two sysDSP slices. The Certus™-NX-RT sysDSP Slice has been significantly enhanced to provide functions needed for advanced processing applications. These enhancements provide improved flexibility and resource utilization.

The Certus™-NX-RT sysDSP block (two sysDSP slices) supports many functions that include the following:

- Symmetry support. The primary target application is wireless. 1D Symmetry is useful for many applications that use FIR filters when their coefficients have symmetry or asymmetry characteristics. The main motivation for using 1D symmetry is cost/size optimization. The expected size reduction is up to 2x.
  - Odd Mode – Filter with Odd number of taps
  - Even Mode – Filter with Even number of taps
  - Two-dimensional (2D) Symmetry Mode – Supports 2D filters for mainly video applications
- Dual-multiplier architecture. Lower accumulator overhead to half and the latency to half compared to single multiplier architecture.
- Fully cascadable DSP across slices. Support for symmetric, asymmetric and non-symmetric filters.
- Multiply (36 × 36, two 18 × 36, four 18 × 18, or eight 9 × 9)
- Multiply Accumulate (supports one 18 × 36 multiplier result accumulation, two 18 × 18 multiplier result accumulation or four 9 × 9 multiplier result accumulation)
- Two Multiplies feeding one Accumulate per cycle for increased processing with lower latency (two 18 × 18 Multiplies feed into an accumulator that can accumulate up to 54 bits)
- Pipeline registers
- 1D Symmetry support. The coefficients of FIR filters have symmetry or negative symmetry characteristics.
  - Odd Mode – Filter with Odd number of taps
  - Even Mode – Filter with Even number of taps
- 2D Symmetry support. The coefficients of 2D FIR filters have symmetry or negative symmetry characteristics.
  - 3 × 3 and 3 × 5 – Internal DSP Slice support
  - 5 × 5 and larger size 2D blocks – Semi internal DSP Slice support
- Flexible saturation and rounding options to satisfy a diverse set of applications situations
- Flexible cascading DSP blocks
  - Minimizes fabric use for common DSP functions
  - Enables implementation of FIR Filter or similar structures using dedicated sysDSP slice resources only
  - Provides matching pipeline registers
  - Can be configured to continue cascading from one row of sysDSP slices to another for longer cascade chains
- RTL Synthesis friendly synchronous reset on all registers, while still supporting asynchronous reset for legacy users
- Dynamic MUX selection to allow Time Division Multiplexing (TDM) of resources for applications that require processor-like flexibility that enables different functions for each clock cycle

For most cases, as shown in Figure 2.15, the Certus™-NX-RT sysDSP block is backwards-compatible with the LatticeECP3™ sysDSP block, such that, legacy applications can be targeted to Certus™-NX-RT sysDSP. Figure 2.15 shows the diagram of sysDSP block.

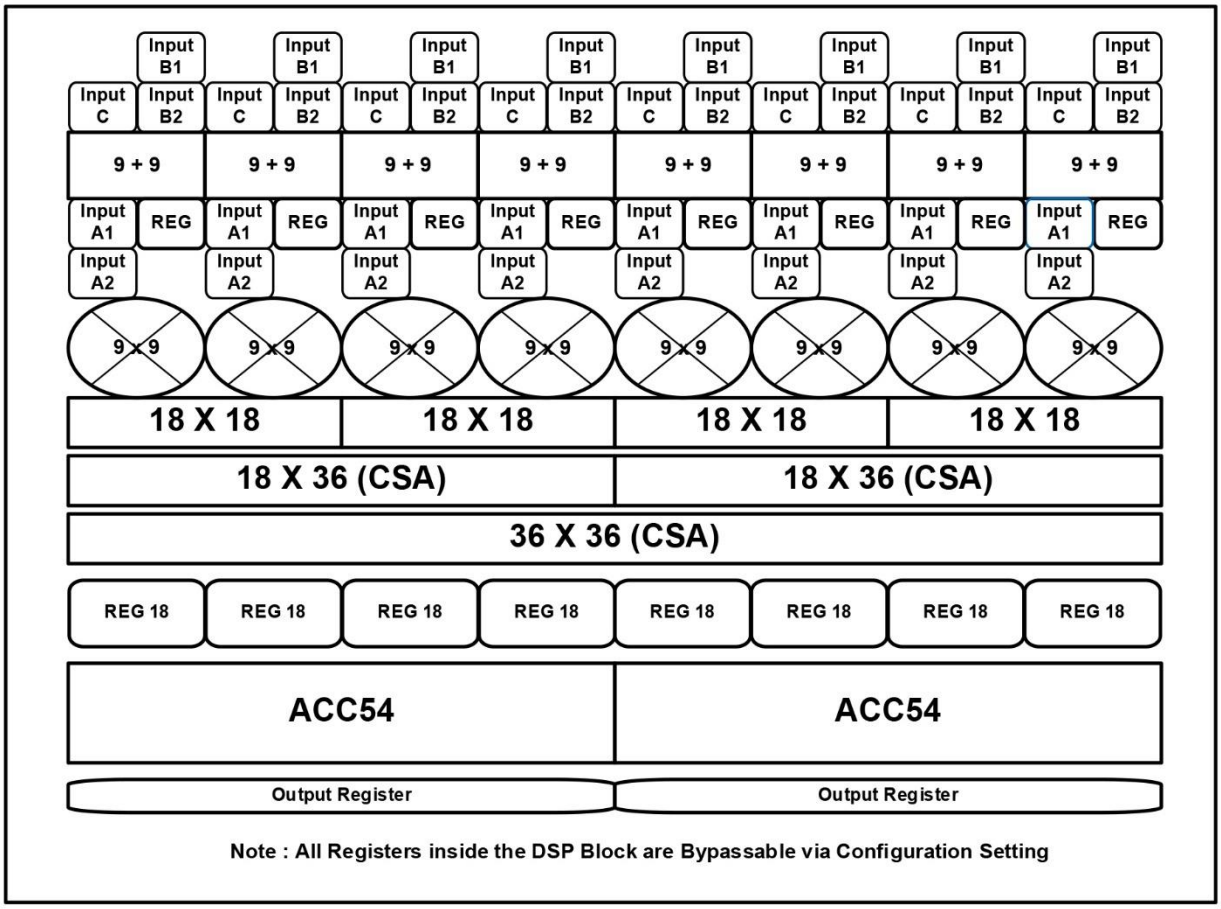


Figure 2.15. Certus™-NX-RT DSP Functional Block Diagram

The Certus™-NX-RT sysDSP block supports the following basic elements.

- MULT (Multiply)
- MAC (Multiply, Accumulate)
- MULTADDSUB (Multiply, Addition/Subtraction)
- MULTADDSUBSUM (Multiply, Addition/Subtraction, Summation)

Table 2.5 shows the capabilities of Certus™-NX-RT sysDSP block versus the above functions.

Table 2.5. Maximum Number of Elements in a sysDSP block

Width of Multiply	x9	x18	x36
MULT	8	4	1
MAC	2	2	—
MULTADDSUB	2	2	—
MULTADDSUBSUM	2	2	—

Some options are available in the four elements. The input register in all the elements can be directly loaded or can be loaded as a shift register from previous operand registers. By selecting *dynamic operation,* the following operations are possible:

- In the Add/Sub option, the Accumulator can be switched between addition and subtraction on every cycle.
- The loading of operands can switch between parallel and serial operations.

For further information, refer to sysDSP Usage Guide for Nexus Platform (FPGA-TN-02096).

Programmable I/O (PIO)

The programmable logic associated with an I/O is called a PIO. The individual PIO are connected to their respective sysI/O buffers and pads. In Certus™-NX-RT devices, the PIO are assembled into groups of two PIO cells called a Programmable I/O Cell or PIC. The PICs are placed on all four sides of the device.

On all the Certus™-NX-RT devices, two adjacent PIO can be combined to provide a complementary output driver pair.

Programmable I/O Cell (PIC)

The programmable I/O cells (PIC) provides I/O function and necessary gearing logic associated with PIO. Certus™-NX-RT consists of base PIC and gearing PIC.

Base PICs contain three blocks: an input register block, output register block, and tri-state register block. These blocks contain registers for operating in a variety of modes along with the necessary clock and selection logic. Base PICs cover the top and left/right bank. Gearing PICs contain gearing logic and edge monitor used for locating the center of data window. Gearing PICs cover the bottom banks to support DDR operation.

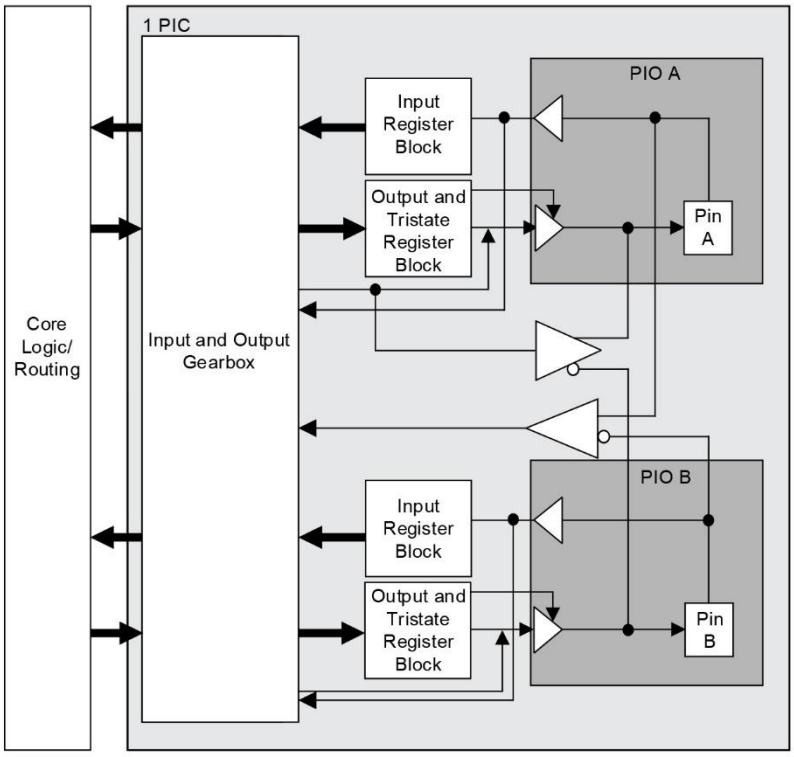


Figure 2.16. Group of Two High Performance Programmable I/O Cells

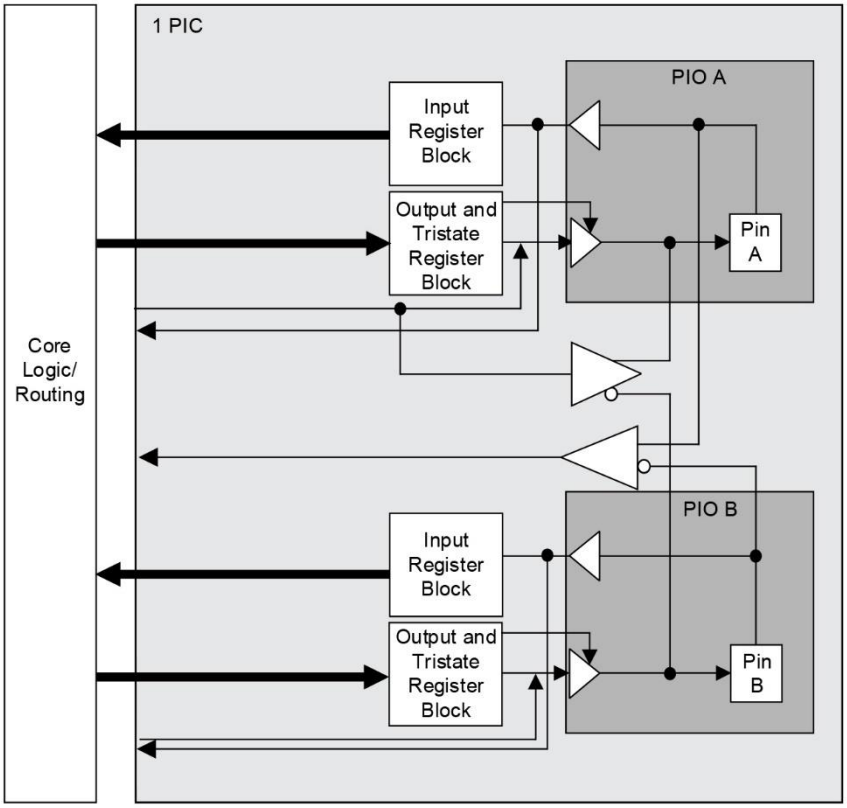


Figure 2.17. Wide Range Programmable I/O Cells

Input Register Block

The input register blocks for the PIO on all edges contain delay elements and registers that can be used to condition high-speed interface signals before they are passed to the device core. In addition, the input register blocks for the PIO on the bottom edges include built-in FIFO logic to interface to DDR and LPDDR memory.

The Input register block on the bottom side includes gearing logic and registers to implement IDDRX1, IDDRX2, IDDRX4, IDDRX5 gearing functions. With two PICs sharing the DDR register path, it can also implement the IDDRX71 function used for 7:1 LVDS interfaces. It uses three sets of registers – shift, update, and transfer to implement gearing and the clock domain transfer. The first stage registers sample the high-speed input data by the high-speed edge clock on its rising and falling edges. The second stage registers perform data alignment based on the control signals. The third stage pipeline registers pass the data to the device core synchronized to the low-speed system clock. For more information on gearing function, refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216).

Input FIFO

The Certus™-NX-RT PIO has dedicated input FIFO per single-ended pin for input data register for DDR Memory interfaces. The FIFO resides before the gearing logic. It transfers data from DQS domain to continuous ECLK domain. On the Write side of the FIFO, it is clocked by DQS clock, which is the delayed version of the DQS Strobe signal from DDR memory. On the Read side of FIFO, it is clocked by ECLK. ECLK may be any high-speed clock with identical frequency as DQS (the frequency of the memory chip). Each DQS group has one FIFO control block. It distributes FIFO read/write pointer to every PIC in same DQS group. DQS Grouping and the DQS Control Block is described in DDR Memory Support section.

Table 2.6. Input Block Port Description

Name	Type	Description
D	Input	High Speed Data Input
Q[1:0]/Q[3:0]/Q[6:0]/Q[7:0]/Q[9:0]	Output	Low Speed Data to the device core
RST	Input	Reset to the Output Block
SCLK	Input	Slow Speed System Clock
ECLK	Input	High Speed Edge Clock
DQS	Input	Clock from DQS control Block used to clock DDR memory data
ALIGNWD	Input	Data Alignment signal from device core.

Figure 2.18 shows the input register block for the PIO on the top, left, and right edges.

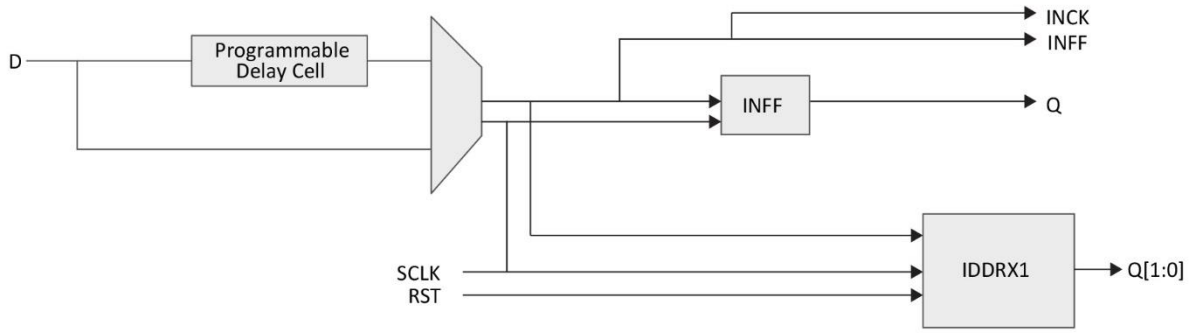


Figure 2.18. Input Register Block for PIO on Top, Left, and Right Sides of the Device

Figure 2.19 shows the input register block for the PIO located on the bottom edge.

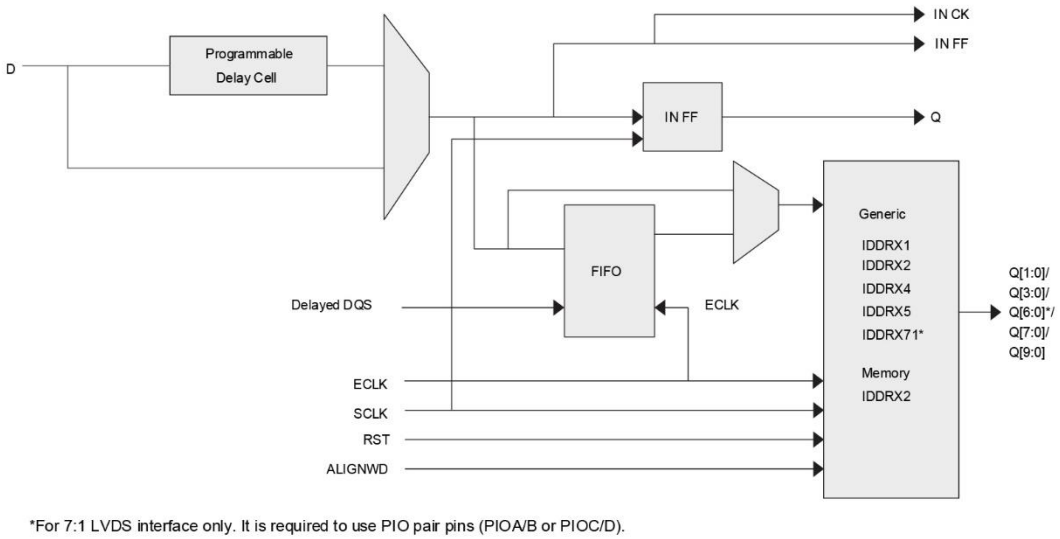


Figure 2.19. Input Register Block for PIO on Bottom Side of the Device

Output Register Block

The output register block registers signal from the core of the device before they are passed to the sysI/O buffers.

The Certus™-NX-RT output data path has programmable registers and output gearing logic. On the bottom side, the output register block can support 1x, 2x, x4, x5, and 7:1 gearing enabling high speed DDR and DDR memory interfaces. On the top, left, and right sides, the banks support 1x gearing. The Certus™-NX-RT output data path diagram is shown in Figure 2.20. The programmable delay cells are also available in the output data path.

For detailed description of the output register block modes and usage, you can refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216).

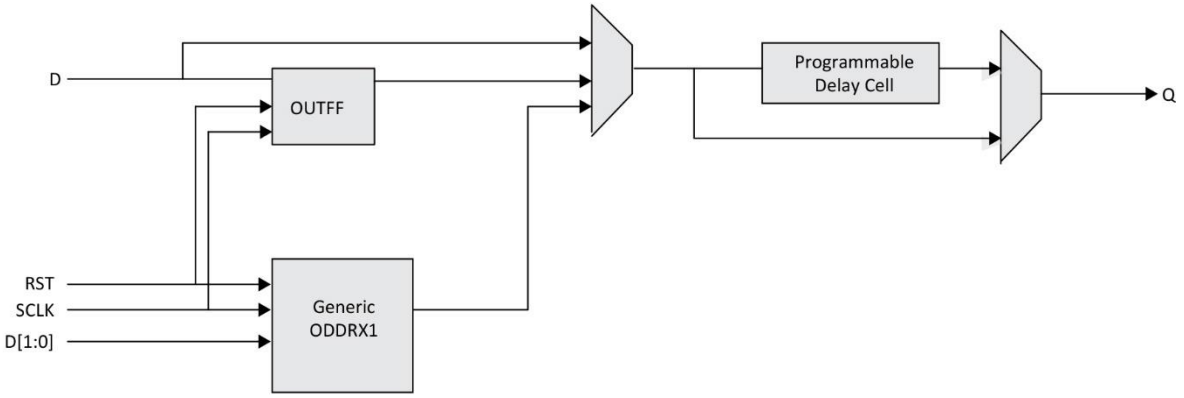
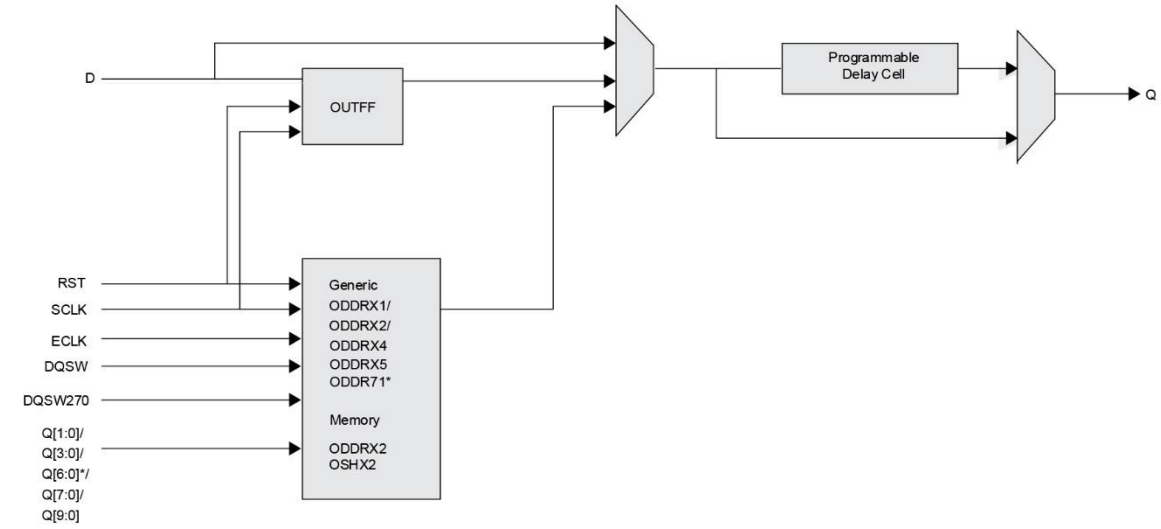


Figure 2.20. Output Register Block on Top, Left, and Right Sides



*For 7:1 LVDS interface only. It is required to use PIO pair pins PIOA/B.

Figure 2.21. Output Register Block on Bottom Side

Table 2.7. Output Block Port Description

Name	Type	Description
Q	Output	High Speed Data Output
D	Input	Data from core to output SDR register
Q[1:0]/Q[3:0]/Q[6:0]/Q[7:0]/Q[9:0]	Input	Low Speed Data from device core to output DDR register
RST	Input	Reset to the Output Block
SCLK	Input	Slow Speed System Clock
ECLK	Input	High Speed Edge Clock
DQSW	Input	Clock from DQS control Block used to generate DDR memory DQS output
DQSW270	Input	Clock from DQS control Block used to generate DDR memory DQ output

Tri-state Register Block

The tri-state register block registers tri-state control signals from the core of the device before they are passed to the sysI/O buffers. The block contains a register for SDR operation. In SDR, the TD input feeds one of the flip-flops that then feeds the output. In DDR, operations used mainly for DDR memory interfaces can be implemented on the bottom side of the device. Here, two inputs feed the tri-state registers clocked by both ECLK and SCLK.

Figure 2.22 and Figure 2.23 show the Tri-state Register Block functions on the device. For detailed description of the tri-state register block modes and usage, you can refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216).

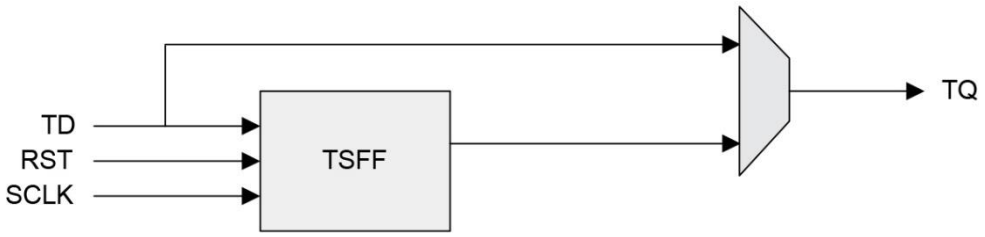


Figure 2.22. Tri-state Register Block on Top, Left, and Right Sides

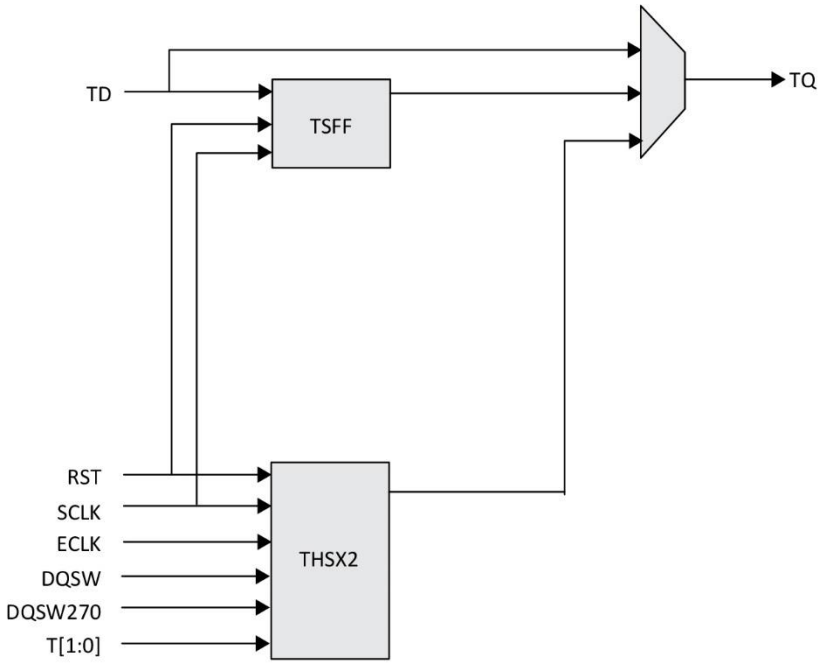


Figure 2.23. Tri-state Register Block on Bottom Side

Table 2.8. Tri-state Block Port Description

Name	Type	Description
TD	Input	Tri-state Input to Tri-state SDR Register
RST	Input	Reset to the Tri-state Block
TD[1:0]	Input	Tri-state input to TSHX2 function
SCLK	Input	Slow Speed System Clock
ECLK	Input	High Speed Edge Clock
DQSW	Input	Clock from DQS control Block used to generate DDR memory DQS output
DQSW270	Input	Clock from DQS control Block used to generate DDR memory DQ output
TQ	Output	Output of the Tri-state block

DDR Memory Support

DQS Grouping for DDR Memory

Certain PICs have additional circuitry to allow the implementation of high-speed source synchronous and DDR3/DDR3L, LPDDR2 or LPDDR3 memory interfaces. The support varies by the edge of the device as detailed below.

The Bottom side of the PIC have fully functional elements supporting DDR3/DDR3L, LPDDR2, or LPDDR3 memory interfaces. Every 16 PIO on the bottom side are grouped into one DQS group, as shown in Figure 2.24. Within each DQS group, there are two pre-placed pins for DQS and DQS# signals. The rest of the pins in the DQS group can be used as DQ signals and DM signal. The number of pins in each DQS group bonded out is package dependent. DQS groups with less than 11 pins bonded out can only be used for LPDDR2/3 Command/ Address busses. In DQS groups with more than 11 pins bonded out, up to two pre-defined pins are assigned to be used as virtual V_{CCIO}, by driving them high to make extra connections to the V_{CCIO} power supply. These soft connections to V_{CCIO} help reduce SSO noise. For details, refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216).

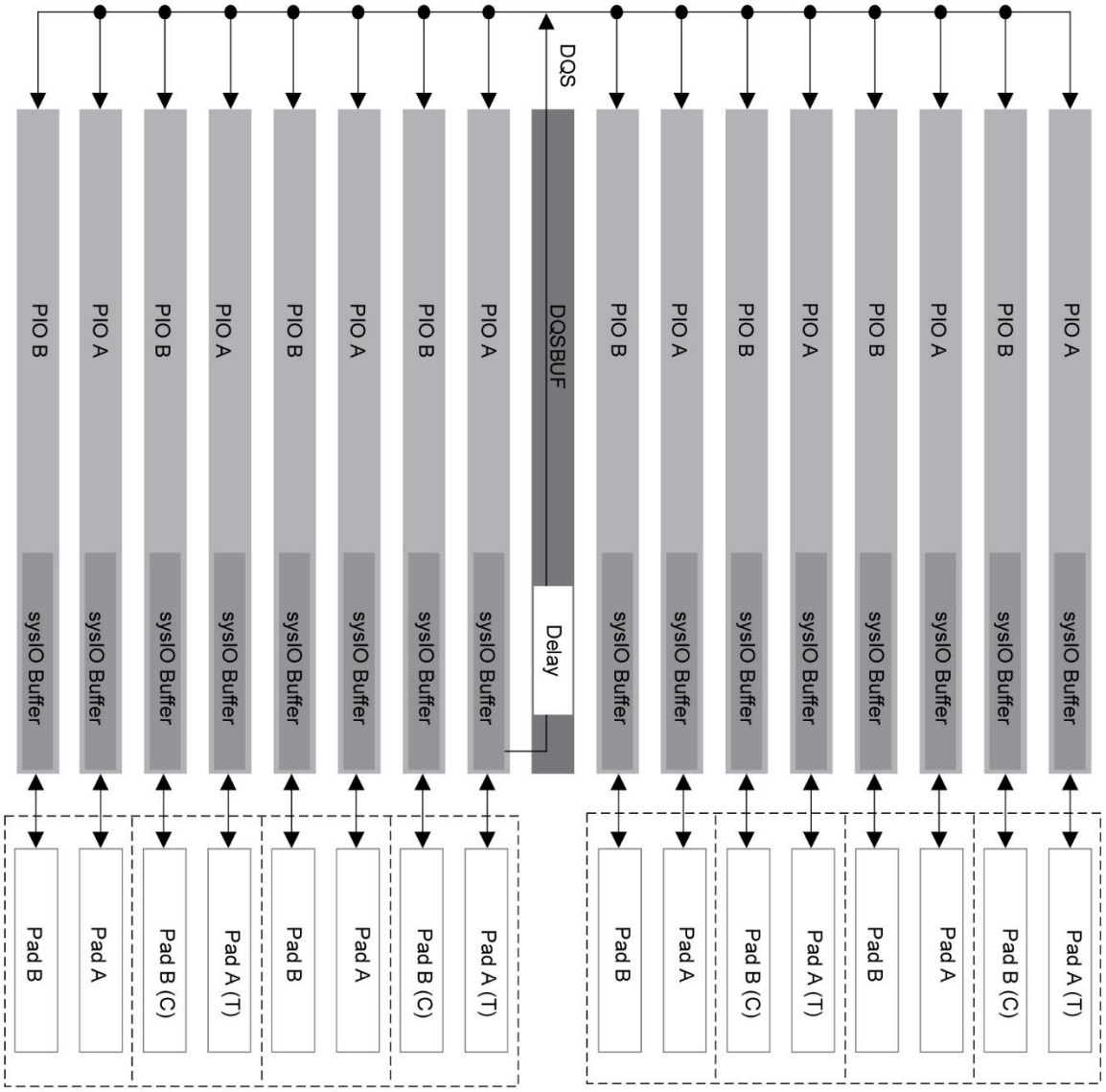


Figure 2.24 DQS Grouping on the Bottom Edge

DLL Calibrated DQS Delay and Control Block (DQSBUF)

To support DDR memory interfaces (DDR3/DDR3L, LPDDR2/3), the DQS strobe signal from the memory must be used to capture the data (DQ) in the PIC registers during memory reads. This signal is output from the DDR memory device aligned to data transitions and must be time shifted before it can be used to capture data in the PIC. This time shift is achieved by using the DQSBUF programmable delay line in the DQS Delay Block (DQS read circuit). The DQSBUFL is implemented as a slave delay line and works in conjunction with a master DDRDLL.

This block also includes slave delay line to generate delayed clocks used during writes to generate DQ and DQS with correct phases within one DQS group. There is a third delay line inside this block used to provide write leveling for DDR write if needed.

Each of the read and write side delays can be dynamically shifted using margin control signals from the core logic.

The FIFO Control Block include here generates the Read and Write Pointers for the FIFO inside the Input Register Block. These pointers are generated to control the DQS to ECLK domain crossing using the FIFO module.

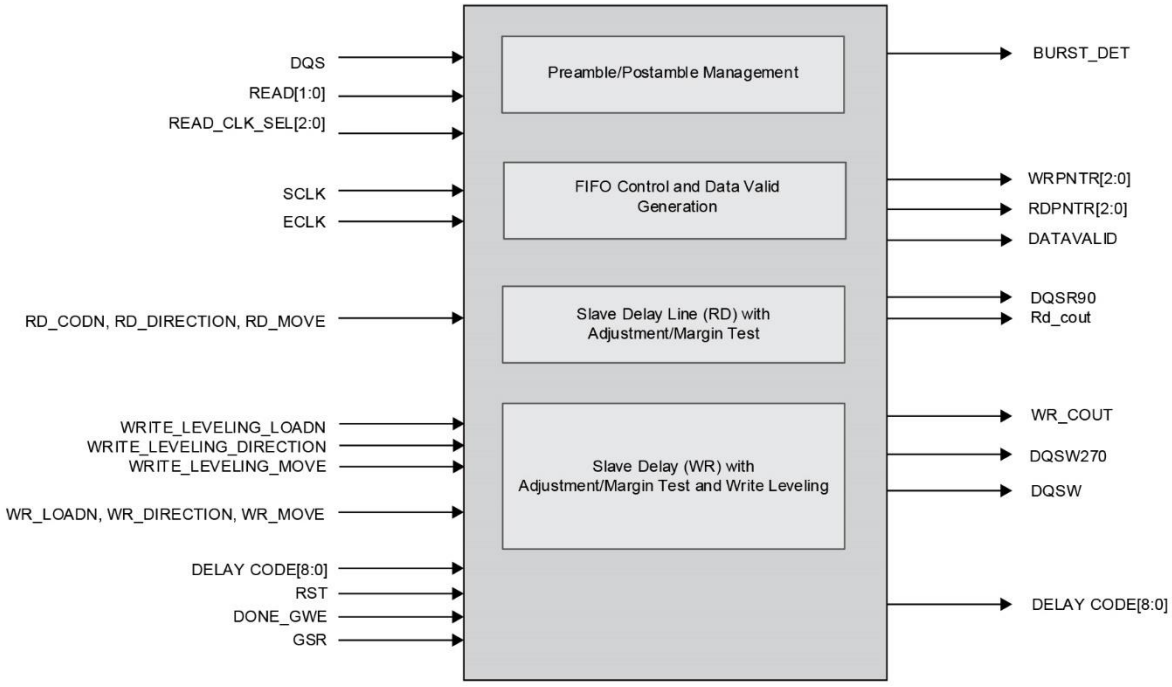


Figure 2.25. DQS Control and Delay Block (DQSBUF)

Table 2.9. DQSBUF Port List Description

Name	Type	Description
DQS	Input	DDR memory DQS strobe
READ[1:0]	Input	Read Input from DDR Controller
READCLKSEL[2:0]	Input	Read pulse selection
SCLK	Input	Slow System Clock
ECLK	Input	High Speed Edge Clock (same frequency as DDR memory)
RDLOADN, RDMOVE, RDDIRECTION	Input	Dynamic Margin Control ports for Read delay
WRLOADN, WRMOVE, WRDIRECTION	Input	Dynamic Margin Control ports for Write delay
DELAYCODE_I[8:0]	Input	Dynamic Delay Control
WRITE_LEVELING_LOADN, WRITE_LEVELING_DIRECTION, WRITE_LEVELING_MOVE	Input	Write Leveling Control
DQSR90	Output	90 delay DQS used for Read
DQSW270	Output	90 delay clock used for DQ Write
DQSW	Output	Clock used for DQS Write
RDPNTR[2:0]	Output	Read Pointer for IFIFO module
WRPNTR[2:0]	Output	Write Pointer for IFIFO module
DATAVALID	Output	Signal indicating start of valid data
BURSTDET	Output	Burst Detect indicator
RD_COUT	Output	Read Count
WR_COUT	Output	Write Count
DELAYCODE_O[8:0]	Output	Dynamic Delay Control

sysI/O Buffer

Each I/O is associated with a flexible buffer referred to as a sysI/O buffer. These buffers are arranged around the periphery of the device in groups referred to as banks. The sysI/O buffers allows you to implement a wide variety of standards that are found in today's systems including LVDS, HSUL, SSTL Class I and II, LVCMOS, and LVTTL.

The Certus™-NX-RT family contains multiple Programmable I/O Cell (PIC) blocks. Each PIC contains two Programmable I/O, PIOA and PIOB. Each PIO includes a sysI/O buffer and I/O logic. Two adjacent PIO can be joined to provide a differential I/O pair referred to as True and Comp, where True Pad is associated with the positive side of the differential I/O, and the complement with the negative.

The top, left, and right side banks support I/O standards from 3.3 V to 1.0 V while the bottom supports I/O standards from 1.8 V to 1.0 V. Every pair of I/O on the bottom bank also have a true LVDS and SLVS Tx Driver. In addition, the bottom bank supports single-ended input termination. Both static and dynamic termination are supported. Dynamic termination is used to support the DDR/LPDDR interface standards. For more information about DDR implementation in I/O Logic and DDR memory interface support, refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216).

Supported sysI/O Standards

Certus™-NX-RT sysI/O buffers supports both single-ended differential and differential standards. Single-ended standards can be further subdivided into internally ratioed standards such as LVCMOS, LVTTL, and externally referenced standards such as HSUL and SSTL. The buffers support the LVTTL, LVCMOS 1.0 V, 1.2 V, 1.5 V, 1.8 V, 2.5 V, and 3.3 V standards. Differential standards supported include LVDS, SLVS, differential LVCMOS, differential SSTL, and differential HSUL. For better support of video standards, subLVDS is also supported. Table 2.10 and Table 2.11 provide a list of sysI/O standards supported in Certus™-NX-RT devices.

Table 2.10. Single-Ended I/O Standards

Standard	Input	Output	Bi-directional
LVTTL33	Yes	Yes	Yes
LVCMOS33	Yes	Yes	Yes
LVCMOS25	Yes	Yes	Yes
LVCMOS18	Yes	Yes	Yes
LVCMOS15	Yes	Yes	Yes
LVCMOS12	Yes	Yes	Yes
LVCMOS10	Yes	No	No
HTSL15 I	Yes	Yes	Yes
SSTL 15 I	Yes	Yes	Yes
SSTL 135 I	Yes	Yes	Yes
HSUL12	Yes	Yes	Yes
LVCMOS18H	Yes	Yes	Yes
LVCMOS15H	Yes	Yes	Yes
LVCMOS12H	Yes	Yes	Yes
LVCMOS10H	Yes	Yes	Yes
LVCMOS10R	Yes	—	Yes*

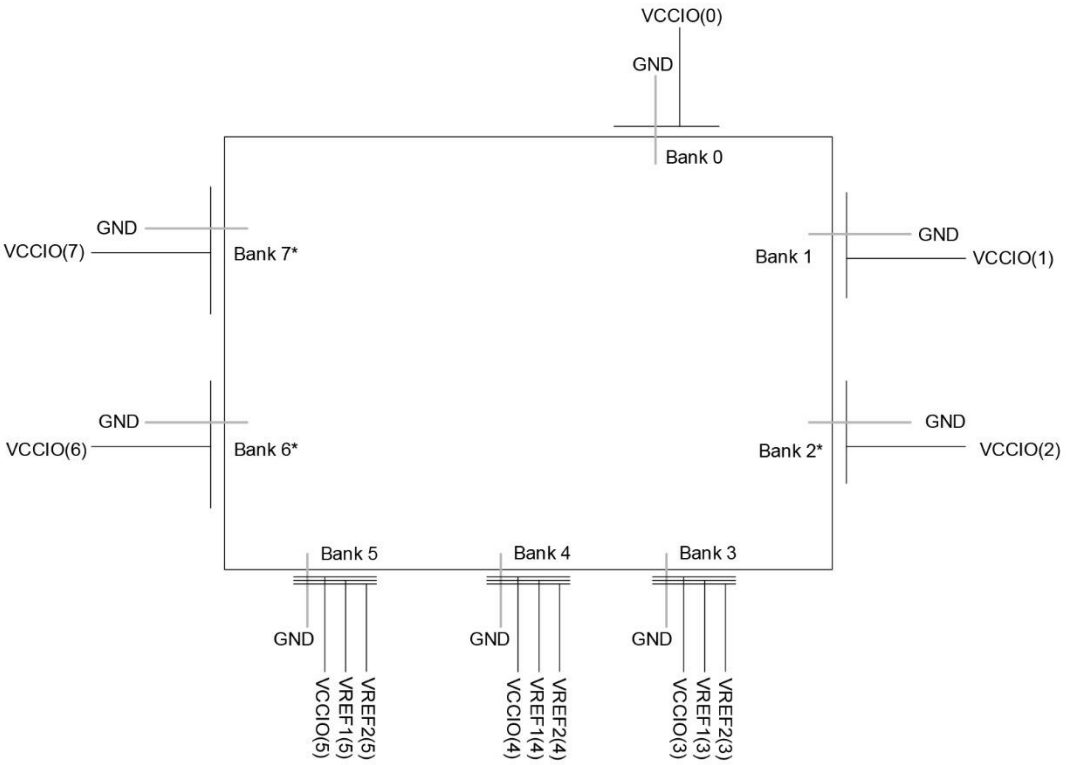
***Note:** Output supported by LVCMOS10H.

Table 2.11. Differential I/O Standards

Standard	Input	Output	Bi-directional
LVDS	Yes	Yes	Yes
SUBLVDS	Yes	No	—
SLVS	Yes	Yes	—
SUBLVDSE	—	Yes	—
SUBLVDSEH	—	Yes	—
LVDSE	—	Yes	—
HSTL15D_I	Yes	Yes	Yes
SSTL15D_I	Yes	Yes	Yes
SSTL15D_II	Yes	Yes	Yes
SSTL135D_I	Yes	Yes	Yes
SSTL135D_II	Yes	Yes	Yes
HSUL12D	Yes	Yes	Yes
LVTTL33D	—	Yes	—
LVCMOS33D	—	Yes	—
LVCMOS25D	—	Yes	—

sysI/O Banking Scheme

Certus™-NX-RT devices have up to eight banks in total. For 40K device, there are one bank on top, two banks each at left and right side of device, and three on the bottom side of device. The higher density Certus™-NX-RT device has more pins in each bank. Bank 0, Bank 1, Bank 2, Bank 6, and Bank 7 support up to V_{CCIO} 3.3 V while Bank 3, Bank 4, and Bank 5 support up to V_{CCIO} 1.8 V. In addition, Bank 3, Bank 4, and Bank 5 support two V_{REF} inputs for flexibility to receive two different referenced input levels on the same bank. Figure 2.26 shows the location of each bank.



***Note:** Bank not available in LFD2NX-17.

Figure 2.26. sysI/O Banking

Typical sysI/O Behavior During Power-up

The internal Power-On-Reset (POR) signal is deactivated when V_{CC} and V_{CCAUX} have reached satisfactory levels. After the POR signal is deactivated, the FPGA core logic becomes active. It is your responsibility to ensure that all other V_{CCIO} banks are active with valid input logic levels to properly control the output logic states of all the I/O banks that are critical to the application. For more information about controlling the output logic state with valid input logic levels during power-up in Certus™-NX-RT devices, see the list of technical documentation in

References section.

V_{CC} and V_{CCAUX} supply the power to the FPGA core fabric, whereas V_{CCIO} supplies power to the I/O buffers. In order to simplify the system design while providing consistent and predictable I/O behavior, it is recommended that the I/O buffers be powered-up prior to the FPGA core fabric. For the different power supply voltage level by the I/O banks, refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216) for detailed information.

VREF1 and VREF2

Bank 3, Bank 4, and Bank 5 can support two separate VREF input voltages, VREF1, and VREF2. To assign a VREF driver, use IO_Type = VREF1_DRIVER or VREF2_DRIVER. To assign VREF to a buffer, use VREF1_LOAD or VREF2_LOAD.

SysI/O Standards Supported by I/O Bank

All banks can support multiple I/O standards under the V_{CCIO} rules discussed above. Table 2.12 and Table 2.13 summarize the I/O standards supported on various sides of the Certus™-NX-RT device.

Table 2.12. Single-Ended I/O Standards Supported on Various Sides

Standard	Top	Left	Right	Bottom
LVTTL33	Yes	Yes	Yes	—
LVCMOS33	Yes	Yes	Yes	—
LVCMOS25	Yes	Yes	Yes	—
LVCMOS18	Yes	Yes	Yes	—
LVCMOS15	Yes	Yes	Yes	—
LVCMOS12	Yes	Yes	Yes	—
LVCMOS10	Yes	Yes	Yes	—
LVCMOS18H	—	—	—	Yes
LVCMOS15H	—	—	—	Yes
LVCMOS12H	—	—	—	Yes
LVCMOS10H	—	—	—	Yes
LVCMOS10R	—	—	—	Yes
HTSL15 I	—	—	—	Yes
SSTL 15 I, II	—	—	—	Yes
SSTL 135 I, II	—	—	—	Yes
HSUL12	—	—	—	Yes

Table 2.13. Differential I/O Standards Supported on Various Sides

Standard	Top	Left	Right	Bottom
LVDS	—	—	—	Yes
SUBLVDS	—	—	—	Yes
SLVS	—	—	—	Yes
SUBLVDSE	Yes	Yes	Yes	—
SUBLVDSEH	—	—	—	Yes
LVDSE	Yes	Yes	Yes	—
HSTL15D_I	—	—	—	Yes
SSTL15D_I	—	—	—	Yes
SSTL15D_II	—	—	—	Yes
SSTL135D_I	—	—	—	Yes
SSTL135D_II	—	—	—	Yes
HSUL12D	—	—	—	Yes
LVTTL33D	Yes	Yes	Yes	—
LVCMOS33D	Yes	Yes	Yes	—
LVCMOS25D	Yes	Yes	Yes	—

Hot Socketing

The Certus™-NX-RT devices have been carefully designed to ensure predictable behavior during power-up and power-down. During power-up and power-down sequences, the I/O remain in tri-state until the power supply voltage is high enough to ensure reliable operation. In addition, leakage into I/O pins is controlled within specified limits. Bank 0, Bank 1, Bank 2, Bank 6, and Bank 7 are fully hot socketable while Bank 3, Bank 4, and Bank 5 do not support hot socketing.

sysI/O Buffer Configurations

This section describes the various sysI/O features available on the Certus™-NX-RT device. Refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216) for detailed information.

Analog Interface

The Certus™-NX-RT family provides an analog interface, consisting of two Analog to Digital Convertors (ADC), three continuous time comparators and an internal junction temperature monitoring diode. The two ADCs can sample the input sequentially or simultaneously.

Analog to Digital Converters

The Analog to Digital Convertor is a 12-bit, 1 MSPS SAR (Successive Approximation Register) architecture converter. The ADC supports both continuous and single shot conversion modes.

The ADC input is selected among pre-selected GPIO input pairs, dedicated analog input pair, the internal junction temperature sensing diode and internal voltage rails. The input signal can be converted in either uni-polar or bi-polar mode.

The reference voltage is selectable between the 1.2 V internal reference generator and an external reference. The ADC can convert up to a 1.8 V input signal with a 1.8 V external reference voltage. The ADC has an auto-calibration function which calibrates the gain and offset

Continuous Time Comparators

The continuous-time comparator can be used to monitor a dedicated input pair or a GPIO input pair. The output of the comparator is provided as continuous and latched outputs.

Internal Junction Temperature Monitoring Diode

On-die junction temperature can be monitored using the internal junction temperature monitoring diode. The PTAT (proportional to absolute temperature) diode voltage can be monitored by the ADC to provide a digital temperature readout. Refer to ADC Usage Guide for Nexus Platform (FPGA-TN-02129) for more details.

IEEE 1149.1-Compliant Boundary Scan Testability

All Certus™-NX-RT devices have boundary scan cells that are accessed through an IEEE 1149.1 compliant Test Access Port (TAP). This allows functional testing of the circuit board on which the device is mounted through a serial scan path that can access all critical logic nodes. Internal registers are linked internally, allowing test data to be shifted in and loaded directly onto test nodes, or test data to be captured and shifted out for verification. The test access port consists of dedicated I/O: TDI, TDO, TCK, and TMS. The test access port uses V_{CCIO1} for power supply. The test access port is supported for V_{CCIO1} = 1.8 V - 3.3 V.

For more information, refer to sysCONFIG Usage Guide for Nexus Platform (FPGA-TN-02099).

Device Configuration

All Certus™-NX-RT devices contain various ports that can be used for configuration, including a Test Access Port (TAP). The TAP supports both the IEEE Standard 1149.1 Boundary Scan specification and the IEEE Standard 1532 In-System Configuration specification. JTAG_EN is the only dedicated configuration pin. PPROGRAMN/INITN/DONE are enabled by default but can be turned into GPIO. The remaining sysCONFIG pins are used as dual function pins. Refer to sysCONFIG Usage Guide for Nexus Platform (FPGA-TN-02099) for more information about using the dual-use pins as general purpose I/O.

There are various ways to configure a Certus™-NX-RT device:

- JTAG (TAP)
- Master Serial Peripheral Interface (SPI) – to load from external SPI flash using x1, x2, and x4 QSPI interfaces.
- Inter-Integrated Circuit Bus (I^2C)
- Improved Inter-Integrated Circuit Bus (I3C)
- Slave SPI from a system host
- Lattice Memory Mapped Interface (LMMI), refer to sysI/O Usage Guide for Nexus Platform (FPGA-TN-02067) for details.
- JTAG, SSPI, MSPI, I^2C, and I3C are supported for V_{CCIO} = 1.8 V - 3.3 V

On power-up, based on the voltage level (high or low) of the PROGRAMN pin the FPGA SRAM is configured by the appropriate sysCONFIG port. If PROGRAMN pin is *low,* the FPGA is in the Slave configuration ports (Slave SPI, Slave I^2C or Slave I3C) and is waiting for the correct Slave Configuration port activation key. PROGRAMN must be driven high within 50 ns of the end of transmission of the Slave Configuration port activation key, that is, the de-assertion of SCSN. If no slave port is declared active before the PROGRAMN pin is sensed HIGH, the FPGA is in Master SPI booting mode. In Master SPI booting mode, the FPGA boots from an external SPI flash. Once a configuration port is activated, it remains active throughout that configuration cycle. The IEEE 1149.1 port can be activated any time after power-up by enabling the JTAG_EN pin and sending the appropriate command through the TAP port.

Enhanced Configuration Options

Certus™-NX-RT devices have enhanced configuration features such as:

- Early I/O release
- Bitstream decryption
- Decompression support
- Watchdog Timer support
- Dual and Multi-boot image support

Early I/O Release is a new configuration feature in which certain I/O banks are released earlier so that customer systems have minimal disruption. For more details, refer to sysCONFIG Usage Guide for Nexus Platform (FPGA-TN-02099).

Watchdog Timer is a new configuration feature that helps you add a programmable timer option for timeout applications.

Dual-Boot and Multi-Boot Image Support

Dual-boot and multi-boot images are supported for applications requiring reliable remote updates of configuration data for the system FPGA. After the system is running with a basic configuration, a new boot image can be downloaded remotely and stored in a separate location in the configuration storage device. Any time after the update the Certus™-NX-RT devices can be re-booted from this new configuration file. If there is a problem, such as corrupt data during download or incorrect version number with this new boot image, the Certus™-NX-RT device can revert back to the original backup golden configuration and try again. This all can be done without power cycling the system. For more information, refer to sysCONFIG Usage Guide for Nexus Platform (FPGA-TN-02099).

Single Event Upset (SEU) Handling

Certus™-NX-RT devices are unique because the underlying technology used to build these devices is much more robust and less prone to soft errors.

Certus™-NX-RT devices have an improved, hardware implemented, Soft Error Detection (SED) circuit which can be used to detect SRAM errors so they can be corrected. There are two layers of SED implemented in Certus™-NX-RT making it more robust and reliable.

The SED hardware in Certus™-NX-RT devices is part of the Configuration block. The SED module in Certus™-NX-RT is an enhanced version as compared to the SED modules implemented in other Lattice devices. The configuration data is divided into frames so that the entire FPGA can be programmed precisely with ease. The SED hardware reads data from the FPGAs configuration memory and performs Error Correcting Code (ECC) calculation on every frame of configuration data (see Figure 2.1). Once an error is detected, a notification is generated and SED resumes operation. For single bit errors, the corrected value is rewritten to the particular frame using ECC information. If more than one-bit error is detected within one frame of configuration data, an error message is generated. Certus™-NX-RT devices also have a dedicated logic to perform Cycle Redundancy Code (CRC) checks the entire bitstream, which runs in parallel along with ECC.

After the ECC is calculated on all frames of configuration data, CRC is calculated and checked for the entire bitstream. ECC and CRC checks do not include the contents of the RAMs (EBR, Large RAM, and distributed RAM memory).

For further information on SED support, refer to Soft Error Detection (SED)/Correction (SEC) Usage Guide for Nexus Platform (FPGA-TN-02076).

On-Chip Oscillator

The Certus™-NX-RT device features two on board oscillators. Both Oscillators are controlled with internally generated current.

The low frequency oscillator (LFOSC) is tailored for low power operation and runs at a nominal frequency of 128 kHz. The LFOSC always runs and can be used to perform always on functions with the lowest possible power. The high frequency oscillator (HFOSC) runs at a nominal frequency of 450 MHz but can be divided down to a range of 256 MHz to 2 MHz by user attributes.

User I^2C IP

The Certus™-NX-RT device has one I^2C hard interface, which can be configured either as a master (controller) or as a slave (responder). The pins for the I^2C interface are pre-assigned.

The interface core has the option to delay the either the input or the output data (SDA), or both, by 50 ns nominal, using dedicated on-chip delay elements. This provides an easier interface to any external I^2C components. In addition, 50 ns glitch filters are available for both SDA and SCL.

When the interface is configured as a master (controller), it is able to control other devices on the I^2C bus through the pre-assigned pins. When the core is configured as a slave (responder), the device is able to provide, for example, I/O expansion to an I^2C master (controller).

The I²C core supports the following functionality:

- Master (controller) and Slave (responder) operation
- 7-bit and 10-bit addressing
- Multi-master (controller) arbitration support
- Clock stretching
- Up to 1 MHz data transfer speed (Standard-Mode, Fast-Mode, Fast-Mode Plus)
- General Call support
- Optional receive and transmit data FIFOs with programmable sizes
- Optional 50 ns delay on input or output data (SDA), or both
- Hard-Connection and Programmable I/O Connection Support
- Programmable to a mode compliant with I3C requirements on legacy I²C Slave Devices.
- Fast-Mode and Fast-Mode Plus Support
- Disabled Clock Stretching
- 50 ns SCL and SDA Glitch Filters
- Programmable 7-bit Address

For further information on the User I²C, refer to I2C Hardened IP Usage Guide for Nexus Platform (FPGA-TN-02142).

Trace ID

Each Certus™-NX-RT device contains a unique (per device) TraceID that can be used for tracking purposes or for IP security applications. The TraceID is 64 bits long. Eight out of 64 bits are user-programmable, the remaining 56 bits are factory-programmed. The TraceID is accessible through the SPI, I²C, or JTAG interfaces. For further information on TraceID, refer to Using TraceID (FPGA-TN-02084).

Density Shifting

The Certus™-NX-RT family is designed to ensure that different density devices in the same family and in the same package have the same pinout. Furthermore, the architecture ensures a high success rate when performing design migration from lower density devices to higher density devices. In many cases, it is also possible to shift a lower utilization design targeted for a high-density device to a lower density device. However, the exact details of the final resource utilization impact the likelihood of success in each case. An example is that some user I/O may become No Connects in smaller devices in the same package. Refer to the Certus™-NX-RT Pin Migration Tables and Lattice Radiant software for specific restrictions and limitations.

Peripheral Component Interconnect Express (PCIe)

The Certus™-NX-RT -40 Device features one lane of hardened PCIe block on the top side of the device. The PCIe block implements all three layers defined by the PCI Express Specification: Physical, Data Link, and Transaction as shown in Figure 2.27. Below is a summary of the features supported by the PCIe block:

- Gen 1 (2.5 Gbps) and Gen 2 (5.0 Gbps) speed
- PCIe Express Base Specification 3.0 compliant including compliance with earlier PCI Express Specifications
- Multi-function support with up to four physical functions
- Endpoint and Root Complex support
- Type 0 Configuration Registers in Endpoint Mode
- Complete Error-Handling Support
- 32-bit Core Data Width
- Many power management features including power budgeting

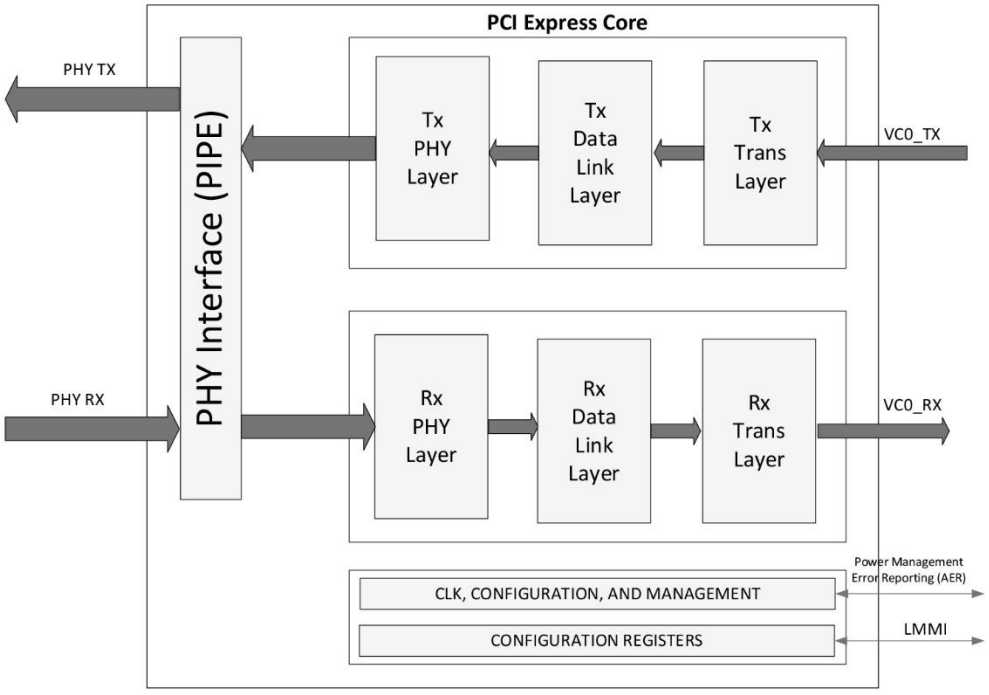


Figure 2.27. PCIe Core

The hardened PCIe block can be instantiated with the primitive PCIe through Lattice Radiant software however, it is not recommended to directly instantiate the PCIe primitive itself. It is highly recommended to generate the PCIe Endpoint Soft IP through the Radiant IP Catalog & IP Block Wizard instead. In Figure 2.28, the PCIe core is configured as Endpoint using a soft IP wrapper that provides useful functions such as bridging support for bus interfaces and DMA applications. In addition to the standard Transaction Layer Packet (TLP) interface, the data interface can also be configured to be AXI4 or AHB-Lite as well. The PCIe hardened block also features a register interface for LMMI and User Configuration Space Register Interface (UCFG). The PCIe block has many registers which contain information about the current status of the PCIe block as well as the capability to dynamically switch PCIe settings. One easy way to access these registers is through the Reveal Controller Tool.

For more information about the PCIe soft IP, refer to the PCIe Endpoint IP Core document.

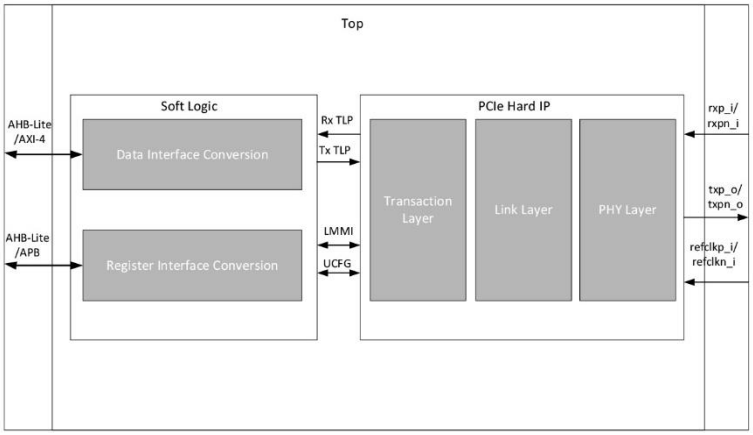


Figure 2.28. PCIe Soft IP Wrapper

Cryptographic Engine

The Certus™-NX-RT family of devices support several cryptographic features that helps customer secure their design. Some of the key cryptographic features include Advanced Encryption Standard (AES) encryption, Hashing Algorithms and true random number generation (TRNG). The Certus™-NX-RT device also features bitstream encryption (using AES-256), used for protecting confidential FPGA bitstream data, and bitstream authentication (using ECDSA), which maintains bitstream integrity and protects the FPGA design bitstream from copying and tampering.

The Cryptographic Engine (CRE) is the main block, which is responsible for bitstream encryption as well as authentication of the Certus™-NX-RT device. Once the bitstream is authenticated and the device is ready for user functions, the CRE is available to implement various cryptographic functions in your FPGA design. To enable specific cryptographic functions, the CRE has to be configured by setting a few registers.

The Cryptographic Engine supports the below user-mode features:

- True Random Number generator (TRNG)
- Secure Hashing Algorithm (SHA)-256 bit
- Message authentication codes (MACs) – HMAC
- Lattice Memory Mapped Interface (LMMI) to user logic
- High Speed Port (HSP) for FIFO-based streaming data transfer

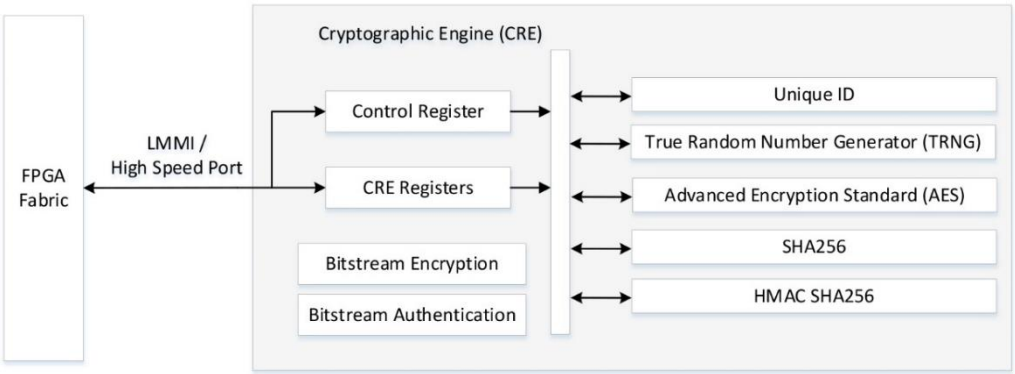


Figure 2.29. Cryptographic Engine Block Diagram

DC and Switching Characteristics

Absolute Maximum Ratings

Table 3.1 Absolute Maximum Ratings

Symbol	Parameter	Min	Max	Unit
V_{CC}, V_{CCECLK}	Supply Voltage	–0.5	1.10	V
V_{CCAUX}, V_{CCAUXA}, $V_{CCAUXH3}$, $V_{CCAUXH4}$, $V_{CCAUXH5}$	Supply Voltage	–0.5	1.98	V
$V_{CCIO0,1,2,6,7}$	I/O Supply Voltage	–0.5	3.63	V
$V_{CCIO3,4,5}$	I/O Supply Voltage	–0.5	1.98	V
$V_{CCPLL\_DPHY0,1}$	Hardened D-PHY PLL Supply Voltage	–0.5	1.10	V
$V_{CCPLLSD0}$	SerDes Block PLL Supply Voltage	–0.5	1.98	V
$V_{CCA\_DPHY0,1}$	Analog Supply Voltage for Hardened D-PHY	–0.5	1.98	V
$V_{CC\_DPHY0,1}$	Digital Supply Voltage for Hardened D-PHY	–0.5	1.10	V
V_{CCSD0}	SerDes Supply Voltage	–0.5	1.10	V
$V_{CCADC18}$	ADC Block 1.8 V Supply Voltage	–0.5	1.98	V
$V_{CCAUXSD}$	SerDes and AUX Supply Voltage	–0.5	1.98	V
—	Input or I/O Voltage Applied, Bank 0, Bank 1, Bank 2, Bank 6, Bank 7	–0.5	3.63	V
—	Input or I/O Voltage Applied, Bank 3, Bank 4, Bank 5	–0.5	1.98	V
—	Voltage Applied on SerDes Pins	–0.5	1.98	V
T_A	Storage Temperature (Ambient)	–65	+150	°C
T_J	Junction Temperature	—	+125	°C

Notes:

1. Stress above those listed under the *Absolute Maximum Ratings* may cause permanent damage to the device. Functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.
2. Compliance with the Lattice Thermal Management document is required.
3. All voltages referenced to GND.
4. All V_{CCAUX} should be connected on PCB.

Operational Environment

Table 3.2 Operational Environment

Symbol	Parameter		Units
TID	Total Ionizing Dose[1]	100	kRad (Si)
SEL	Single Event Latchup Immunity	≤ 80	MeV-cm^2/mg

Notes:

1. Per MIL-STD-883, method 1019, condition A. Device tested at 25C post-irradiation, see report for details
2. Recommended Operating Conditions 1, 2, 3
3. SEL tested at 125 ^{0}C

Table 3.3. Recommended Operating Conditions

Symbol	Parameter	Conditions	Min	Typ.	Max	Unit
V_{CC}, V_{CCECLK}	Core Supply Voltage	V_{CC} = 1.0	0.95	1.00	1.05	V
V_{CCAUX}	Auxiliary Supply Voltage	Bank 0, Bank 1, Bank 2, Bank 6, Bank 7	1.746	1.80	1.89	V
$V_{CCAUXH3/4/5}$	Auxiliary Supply Voltage	Bank 3, Bank 4, Bank 5	1.746	1.80	1.89	V
V_{CCAUXA}	Auxiliary Supply Voltage for core logic	—	1.746	1.80	1.89	V
V_{CCIO}	I/O Driver Supply Voltage	V_{CCIO} = 3.3 V, Bank 0, Bank 1, Bank 2, Bank 6, Bank 7	3.135	3.30	3.465	V
		V_{CCIO} = 2.5 V, Bank 0, Bank 1, Bank 2, Bank 6, Bank 7	2.375	2.50	2.625	V
		V_{CCIO} = 1.8 V, All Banks	1.71	1.80	1.89	V
		V_{CCIO} = 1.5 V, All Banks4	1.425	1.50	1.575	V
		V_{CCIO} = 1.35 V, All Banks (For DDR3L Only)	1.2825	1.35	1.4175	V
		V_{CCIO} = 1.2 V, All Banks4	1.14	1.20	1.26	V
		V_{CCIO} = 1.0 V, Bank 3, Bank 4, Bank 5	0.95	1.00	1.05	V
ADC External Power Supplies						
$V_{CCADC18}$	ADC 1.8 V Power Supply	—	1.71	1.80	1.89	V
SerDes Block External Power Supplies						
V_{CCSD0}	Supply Voltage for SerDes Block and SerDes I/O	—	0.95	1.00	1.05	V
$V_{CCPLLSD0}$	SerDes Block PLL Supply Voltage	—	1.71	1.80	1.89	V
$V_{CCAUXSD}$	SerDes Block Auxiliary Supply Voltage	—	1.71	1.80	1.89	V
Operating Temperature						
t_J	Junction Temperature	—	–40	—	+125	°C

Notes:

1. For correct operation, all supplies must be held in their valid operation voltage range.
2. All supplies with same voltage should be from the same voltage source. Proper isolation filters are needed to properly isolate noise from each other.
3. Common supply rails must be tied together except SerDes.
4. MSPI (Bank 0) and JTAG, SSPI, I^2C, and I3C (Bank 1) ports are supported for V_{CCIO} = 1.8 V to 3.3 V.
5. Data in this section is in preliminary state, subject to change.

Power Supply Ramp Rates

Table 3.4. Power Supply Ramp Rates

Symbol	Parameter	Min	Typ.	Max	Unit
t_{RAMP}	Power Supply ramp rates for all supplies[1]	0.1	—	50	V/ms

Notes:

1. Assumes monotonic ramp rates.
2. All supplies need to be in the operating range as defined in the Recommended Operating Conditions, when the device has completed configuration and entering into User Mode. Supplies that are not in the operating range needs to be adjusted to faster ramp rate, or you have to delay configuration or wake up.

Power Up Sequence

Power-On-Reset (POR) puts the Certus™-NX-RT device into a reset state. There is no power up sequence required for the Certus™-NX-RT device.

Table 3.5. Power-On Reset

Symbol	Parameter		Min	Typ.	Max	Unit
V_{PORUP}	Power-On-Reset ramp-up trip point (Monitoring V_{CC}, V_{CCAUX}, V_{CCIO0}, and V_{CCIO1})	V_{CC}	0.72	—	0.84	V
		V_{CCAUX}	1.30	—	1.71	V
		V_{CCIO0}, V_{CCIO1}	0.87	—	1.07	V
V_{PORDN}	Power-On-Reset ramp-up trip point (Monitoring V_{CC} and V_{CCAUX}	V_{CC}	0.48	—	0.85	V
		V_{CCAUX}	1.36	—	1.57	V

Notes:

1. Assumes monotonic ramp rates.
2. All supplies need to be in the operating range as defined in the Recommended Operating Conditions, when the device has completed configuration and entering into User Mode. Supplies that are not in the operating range needs to be adjusted to faster ramp rate, or you have to delay configuration or wake up.

On-Chip Programmable Termination

The Certus™-NX-RT devices support a variety of programmable of on-chip terminations options, including:

- Dynamically switchable Single-Ended Termination with programmable resistor values of 40 Ω, 50 Ω, 60 Ω, or 75 Ω.
- Common mode termination of 100 Ω for differential inputs.

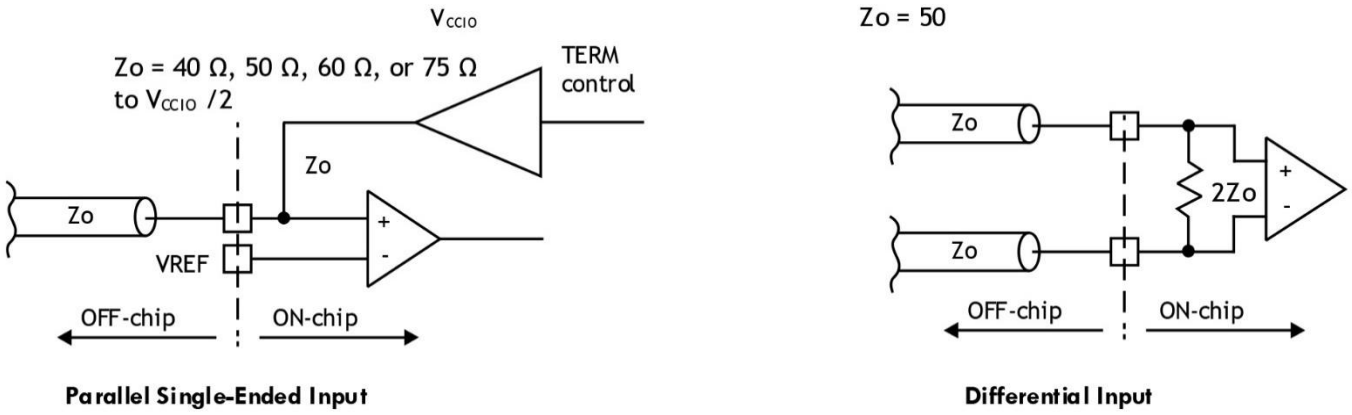


Figure 3.1. On-Chip Termination

See Table 3.6 for termination options for input modes.

Table 3.6. On-Chip Termination Options for Input Modes

IO_TYPE	Differential Termination Resistor*	Terminate to $V_{CCIO}/2$*
subLVDS	100, OFF	OFF
SLVS	100, OFF	OFF
HSTL15D_I	100, OFF	OFF
SSTL15D_I	100, OFF	OFF
SSTL135D_I	100, OFF	OFF
HSUL12D	100, OFF	OFF
LVCMOS15H	OFF	OFF
LVCMOS12H	OFF	OFF
LVCMOS10H	OFF	OFF
LVCMOS12H	OFF	OFF
LVCMOS10H	OFF	OFF
LVCMOS18H	OFF	OFF, 40, 50, 60, 75
HSTL15_I	OFF	50
SSTL15_I	OFF	OFF, 40, 50, 60, 75
SSTL135_I	OFF	OFF, 40, 50, 60, 75
HSUL12	OFF	OFF, 40, 50, 60, 75

Notes:

1. TERMINATE to $V_{CCIO}/2$ (Single-Ended) and DIFFERENTIAL TERMINATION RESISTOR when turned on can only have one setting per bank. Only left and right banks have this feature.
2. Use of TERMINATE to $V_{CCIO}/2$ and DIFFERENTIAL TERMINATION RESISTOR are mutually exclusive in an I/O bank. On-chip termination tolerance -10%/+60%.

Refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216) for on-chip termination usage and value ranges.

Hot Socketing Specifications

Table 3.7. Hot Socketing Specifications for GPIO

Symbol	Parameter	Condition	Min	Typ.	Max	Unit
I_{DK}	Input or I/O Leakage Current for Wide Range I/O (excluding MCLK/MCSN/MOSI/INITN/DONE)	$0 < V_{IN} < V_{IH}$(max) $0 < V_{CC} < V_{CC}$(max) $0 < V_{CCIO} < V_{CCIO}$(max) $0 < V_{CCAUX} < V_{CCAUX}$(max)	-1.5	—	1.5	mA

Notes:

1. IDK is additive to I_{PU}, I_{PD}, or IBH.
2. Hot socket specification defines when the hot socketed device's junction temperature is at 85°C or below. When the hot socketed device's junction temperature is above 85°C, the IDK current can exceed the above spec.
3. Going beyond the hot socketing ranges specified here will cause exponentially higher Leakage currents and potential reliability issues. A total of 64mA per 8 I/O should not be exceeded.

ESD Performance

Refer to the Certus™-NX-RT Product Family Qualification Summary for complete qualification data, including ESD performance.

DC Electrical Characteristics

Table 3.8. DC Electrical Characteristics – Wide Range (Over Recommended Operating Conditions)

Symbol	Parameter	Condition	Min	Typ.	Max	Unit
I_{IL}, I_{IH}[1]	Input or I/O Leakage current (Commercial/Industrial)	$0 \le V_{IN} \le V_{CCIO}$	—	—	10	µA
I_{IH}[2]	Input or I/O Leakage current	$V_{CCIO} \le V_{IN} \le V_{IH}$ (max)	—	—	100	µA
I_{PU}	I/O Weak Pull-up Resistor Current	$0 \le V_{IN} \le 0.7 * V_{CCIO}$	–30	—	–150	µA
I_{PD}	I/O Weak Pull-down Resistor Current	V_{IL} (max) $\le V_{IN} \le V_{CCIO}$	30	—	150	µA
I_{BHLS}	Bus Hold Low Sustaining Current	$V_{IN} = V_{IL}$ (max)	30	—		µA
I_{BHHS}	Bus Hold High Sustaining Current	$V_{IN} = 0.7 * V_{CCIO}$	–30	—		µA
I_{BHLO}	Bus hold low Overdrive Current	$0 \le V_{IN} \le V_{CCIO}$	—	—	150	µA
I_{BHHO}	Bus hold high Overdrive Current	$0 \le V_{IN} \le V_{CCIO}$	—	—	–150	µA
V_{BHT}	Bus Hold Trip Points	—	V_{IL} (max)	—	V_{IH} (min)	V

Notes:

1. Input or I/O leakage current is measured with the pin configured as an input or as an I/O with the output tri-stated. Bus Maintenance circuits are disabled.
2. The input leakage current I_{IH} is the worst-case input leakage per GPIO when the pad signal is high and also higher than the bank V_{CCIO}. This is considered a mixed mode input.

Table 3.9. DC Electrical Characteristics – High Speed (Over Recommended Operating Conditions)

Symbol	Parameter	Condition	Min	Typ.	Max	Unit
I_{IL}, I_{IH}[1]	Input or I/O Leakage current (Commercial/Industrial)	$0 \le V_{IN} \le V_{CCIO}$	—	—	10	µA
I_{PU}	I/O Weak Pull-up Resistor Current	$0 \le V_{IN} \le 0.7 * V_{CCIO}$	–30	—	–150	µA
I_{PD}	I/O Weak Pull-down Resistor Current	V_{IL} (max) $\le V_{IN} \le V_{CCIO}$	30	—	150	µA
I_{BHLS}	Bus Hold Low Sustaining Current	$V_{IN} = V_{IL}$ (max)	30	—		µA
I_{BHHS}	Bus Hold High Sustaining Current	$V_{IN} = 0.7 * V_{CCIO}$	–30	—		µA
I_{BHLO}	Bus hold low Overdrive Current	$0 \le V_{IN} \le V_{CCIO}$	—	—	150	µA
I_{BHHO}	Bus hold high Overdrive Current	$0 \le V_{IN} \le V_{CCIO}$	—	—	–150	µA
V_{BHT}	Bus Hold Trip Points	—	V_{IL} (max)	—	V_{IH} (min)	V

Notes:

1. Input or I/O leakage current is measured with the pin configured as an input or as an I/O with the output tri-stated. Bus Maintenance circuits are disabled.

Table 3.10. Capacitors – Wide Range (Over Recommended Operating Conditions)

Symbol	Parameter	Condition	Min	Typ.	Max	Unit
C_1*	I/O Capacitance*	V_{CCIO} = 3.3 V, 2.5 V, 1.8 V, 1.5 V, 1.2 V, V_{CC} = typ., V_{IO} = 0 to V_{CCIO} + 0.2V	—	6	—	pf
C_2*	Dedicated Input Capacitance*	V_{CCIO} = 3.3 V, 2.5 V, 1.8 V, 1.5 V, 1.2 V, V_{CC} = typ., V_{IO} = 0 to V_{CCIO} + 0.2V	—	6	—	pf

Notes:

1. T_A 25°C, ƒ = 1.0MHz

Table 3.11. Capacitors – High Performance (Over Recommended Operating Conditions)

Symbol	Parameter	Condition	Min	Typ.	Max	Unit
C1*	I/O Capacitance*	V_{CCIO} = 1.8 V, 1.5 V, 1.2 V, V_{CC} = typ., V_{IO} = 0 to V_{CCIO} + 0.2V	—	6	—	pf
C2*	Dedicated Input Capacitance*	V_{CCIO} = 1.8 V, 1.5 V, 1.2 V, V_{CC} = typ., V_{IO} = 0 to V_{CCIO} + 0.2V	—	6	—	pf
C3*	D-PHY I/O Capacitance	$V_{CCA\_D\text{-}PHY}$ = 1.8 V, V_{CC} = typ., V_{IO} = 0 to $V_{CCA\_D\text{-}PHY}$ + 0.2V	—	5	—	pf
C4*	SerDes I/O Capacitance	V_{CCSD0} = 1.0 V, V_{CC} = typ., V_{IO} = 0 to V_{CCSD0} + 0.2 V	—	5	—	pf

Notes:

1. T_A 25°C, ƒ = 1.0MHz

Table 3.12. Single Ended Input Hysteresis – Wide Range (Over Recommended Operating Conditions)

IO_TYPE	V_{CCIO}	TYP Hysteresis
LVCMOS33	3.3 V	250 mV
LVCMOS25	3.3 V	200 mV
	2.5 V	250 mV
LVCMOS18	1.8 V	180 mV
LVCMOS15	1.5 V	50 mV
LVCMOS12	1.2 V	0
LVCMOS10	1.2 V	0

Table 3.13. Single Ended Input Hysteresis – High Performance (Over Recommended Operating Conditions)

IO_TYPE	V_{CCIO}	TYP Hysteresis
LVCMOS18H	1.8 V	180 mV
LVCMOS15H	1.5 V	50 mV
	1.5 V	150 mV
LVCMOS12H	1.2 V	0
LVCMOS10H	1.2 V	0

Supply Currents

For estimating and calculating current, use Power Calculator in Lattice Design software.

This operating and peak current is design dependent and can be calculated in Lattice Design Software. Some blocks can be placed into low current standby modes. Refer to Power Management and Calculation for Certus™-NX-RT (FPGA-TN-02214).

sysI/O Recommended Operating Conditions

Table 3.14. sysI/O Recommended Operating Conditions

Standard	Support Banks	V_{CCIO} (Input)	V_{CCIO} (Output)
		Typ.	Typ.
Single-Ended			
LVCMOS33	0, 1, 2, 6, 7	3.3	3.3
LVTTL33	0, 1, 2, 6, 7	3.3	3.3
LVCMOS25[1,2]	0, 1, 2, 6, 7	2.5, 3.3	2.5
LVCMOS18[1,2]	0, 1, 2, 6, 7	1.2, 1.5, 1.8, 2.5, 3.3	1.8
LVCMOS18H	3, 4, 5	1.8	1.8
LVCMOS15[1,2]	0, 1, 2, 6, 7	1.2, 1.5, 1.8, 2.5, 3.3	1.5
LVCMOS15H[1]	3, 4, 5	1.5, 1.8	1.5
LVCMOS12[1, 2]	0, 1, 2, 6, 7	1.2, 1.5, 1.8, 2.5, 3.3	1.2
LVCMOS12H[1]	3, 4, 5	1.2, 1.356, 1.5, 1.8	1.2
LVCMOS10[1]	0, 1, 2, 6, 7	1.2, 1.5, 1.8, 2.5, 3.3	—
LVCMOS10H[1]	3, 4, 5	1.0, 1.2, 1.35[6], 1.5, 1.8	1.0
LVCMOS10R[1]	3, 4, 5	1.0, 1.2, 1.35[6], 1.5, 1.8	—
SSTL135_I, SSTL135_II[3]	3, 4, 5	1.356	1.35
SSTL15_I, SSTL15_II[3]	3, 4, 5	1.57	1.57
HSTL15_I[3]	3, 4, 5	1.57	1.57
HSUL12[3]	3, 4, 5	1.2	1.2
Differential			
LVDS	3, 4, 5	1.8	1.8
LVDSE5	0, 1, 2, 6, 7	—	2.5
subLVDS	3, 4, 5	1.8	—
subLVDSE[5]	0, 1, 2, 6, 7	—	1.8
subLVDSEH[5]	3, 4, 5	—	1.8
SLVS	3, 4, 5	1.0, 1.2, 1.35[6], 1.5, 1.8[4]	1.2, 1.35[6], 1.5, 1.8[4]
LVCMOS33D[5]	0, 1, 2, 6, 7	—	3.3
LVTTL33D[5]	0, 1, 2, 6, 7	—	3.3
LVCMOS25D[5]	0, 1, 2, 6, 7	—	2.5
SSTL135D_I, SSTL135D_II[5]	3, 4, 5	—	1.356
SSTL15D_I, SSTL15D_II[5]	3, 4, 5	—	1.5
HSTL15D_I[5]	3, 4, 5	—	1.5
HSUL12D[5]	3, 4, 5	—	1.2

Notes:

1. Single-ended input can mix into I/O Banks with V_{CCIO} different from the standard requires due to some of these input standards use internal supply voltage source (V_{CC}, V_{CCAUX}) to power the input buffer, which makes them to be independent of V_{CCIO} voltage. For more details, please refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216). The following is a brief guideline to follow:
   a) Weak pull-up on the I/O must be set to OFF.
   b) Bank 3, Bank 4, and Bank 5 I/O can only mix into banks with V_{CCIO} higher than the pin standard, due to clamping diode on the pin in these banks. Bank 0, Bank 1, Bank 2, Bank 6, and Bank 7 does not have this restriction.
   c) LVCMOS25 uses V_{CCIO} supply on input buffer in Bank 0, Bank 1, Bank 2, Bank 6, and Bank 7. It can be supported with V_{CCIO} = 3.3 V to meet the V_{IH} and V_{IL} requirements, but there is additional current drawn on V_{CCIO}. Hysteresis has to be disabled when using 3.3 V supply voltage.
   d) LVCMOS15 uses V_{CCIO} supply on input buffer in Bank 3, Bank 4, and Bank 5. It can be supported with V_{CCIO} = 1.8 V to meet the V_{IH} and V_{IL} requirements, but there is additional current drawn on V_{CCIO}.
2. Single-ended LVCMOS inputs can mixed into I/O Banks with different V_{CCIO}, providing weak pull-up is not used. For additional information on Mixed I/O in Bank V_{CCIO}, refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216).
3. These inputs use differential input comparator in Bank 3, Bank 4, and Bank 5. The differential input comparator uses V_{CCAUX}H power supply. These inputs require the V_{REF} pin to provide the reference voltage in the Bank. Refer to Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216) for details.
4. All differential inputs use differential input comparator in Bank 3, Bank 4, and Bank 5. The differential input comparator uses V_{CCAUX}H power supply. There is no differential input signaling supported in Bank 0, Bank 1, Bank 2, Bank 6, and Bank 7.
5. These outputs are emulating differential output pair with single-ended output drivers with true and complement outputs driving on each of the corresponding true and complement output pair pins. The common mode voltage, V_{CM}, is ½ * V_{CCIO}. Refer Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216) for details.
6. V_{CCIO} = 1.35 V is only supported in Bank 3, Bank 4, and Bank 5, for use with DDR3L interface in the bank. These Input and Output standards can fit into the same bank with the V_{CCIO} = 1.35 V.
7. LVCMOS15 input uses V_{CCIO} supply voltage. If V_{CCIO} is 1.8 V, the DC levels for LVCMOS15 are still met, but there could be increase in input buffer current.

sysI/O Single-Ended DC Electrical Characteristics[3]

Table 3.15. sysI/O DC Electrical Characteristics – Wide Range I/O (Over Recommended Operating Conditions)

Input/Output Standard	V_{IL}[1]		V_{IH}[1]		V_{OL} Max (V)	V_{OH} Min[2] (V)	I_{OL} (mA)	I_{OH} (mA)
	Min (V)	Max (V)	Min (V)	Max (V)				
LVTTL33 LVCMOS33	—	0.8	2.0	3.465	0.4	V_{CCIO} – 0.4	2, 4, 8, 12, 16	-2, -4, -8, -12, -16
LVCMOS25	—	0.7	1.7	2.625	0.4	V_{CCIO} – 0.45	2, 4, 8, 10	-2, -4, -8, -10
LVCMOS18	—	0.35 * V_{CCIO}	0.65 * V_{CCIO}	1.9	0.4	V_{CCIO} – 0.45	2, 4, 8	-2, -4, -8
LVCMOS15	—	0.35 * V_{CCIO}	0.65 * V_{CCIO}	1.575	0.4	V_{CCIO} – 0.4	2, 4	-2, -4
LVCMOS12	—	0.35 * V_{CCIO}	0.65 * V_{CCIO}	1.26	0.4	V_{CCIO} – 0.4	2, 4	-2, -4, -8, -12
LVCMOS10	—	0.3 * V_{CCIO}	0.7 * V_{CCIO}	1.05	No O/P Support			

Notes:

1. V_{CCIO} for input level refers to the supply rail level associated with a given input standard or the upstream driver V_{CCIO} rail levels.
2. V_{CCIO} for the output levels refer to the V_{CCIO} of the Certus™-NX-RT device.
3. For electro-migration, the combined DC current sourced or sinked by I/O pads between two consecutive V_{CCIO} or GND pad connections, or between the last V_{CCIO} or GND in an I/O bank and the end of an I/O bank, as shown in the Logic Signal Connections table (also shown as I/O grouping) shall not exceed a maximum of n × 8 mA. n is the number of I/O pads between the two consecutive bank V_{CCIO} or GND connections or between the last V_{CCIO} and GND in a bank and the end of a bank. I/O Grouping can be found in the Data Sheet Pin Summary Tables, which can also be generated from the Lattice Radiant software.

Table 3.16. sysI/O DC Electrical Characteristics – High Performance I/O (Over Recommended Operating Conditions)

Input/Output Standard	V_{IL}[1]		V_{IH}[1]		V_{OL} Max (V)	V_{OH} Min[2] (V)	I_{OL} (mA)	I_{OH} (mA)
	Min (V)	Max (V)	Min (V)	Max (V)				
LVCMOS18H		0.35 * V_{CCIO}	0.65 * V_{CCIO}	V_{CCIO} + 0.3	0.4	V_{CCIO} – 0.45	2, 4, 8, 12	-2, -4, -8, -12
LVCMOS15H		0.35 * V_{CCIO}	0.65 * V_{CCIO}	V_{CCIO} + 0.3	0.4	V_{CCIO} – 0.4	2, 4, 8	-2, -4, -8
LVCMOS12H		0.35 * V_{CCIO}	0.65 * V_{CCIO}	1.26	0.4	V_{CCIO} – 0.4	2, 4, 8	-2, -4, -8
LVCMOS10H		0.3 * V_{CCIO}	0.7 * V_{CCIO}	1.05	0.27 * V_{CCIO}	0.75 * V_{CCIO}	2, 4	-2, -4
SSTL15_I		V_{REF} – 0.10	V_{REF} + 0.1	1.575	0.30	V_{CCIO} – 0.30	7.5	–7.5
SSTL15_II		V_{REF} – 0.10	V_{REF} + 0.1	1.575	0.30	V_{CCIO} – 0.30	8.8	–8.8
HSTL15_I		V_{REF} – 0.10	V_{REF} + 0.1	1.575	0.40	V_{CCIO} – 0.40	8	–8
SSTL135_I		V_{REF} – 0.09	V_{REF} + 0.09	1.418	0.27	V_{CCIO} – 0.27	6.75	–6.75
SSTL135_II		V_{REF} – 0.09	V_{REF} + 0.09	1.418	0.27	V_{CCIO} – 0.27	8	–8
LVCMOS10R		V_{REF} – 0.10	V_{REF} + 0.10	1.05	—	—	—	—
HSUL12		V_{REF} – 0.10	V_{REF} + 0.10	1.26	0.3	V_{CCIO} – 0.3	8.0, 7.5, 6.25, 5	-8.0, -7.5, -6.25, -5

Notes:

1. V_{CCIO} for input level refers to the supply rail level associated with a given input standard or the upstream driver V_{CCIO} rail levels.
2. V_{CCIO} for the output levels refer to the V_{CCIO} of the Certus™-NX-RT device.
3. For electro-migration, the combined DC current sourced or sinked by I/O pads between two consecutive V_{CCIO} or GND pad connections, or between the last V_{CCIO} or GND in an I/O bank and the end of an I/O bank, as shown in the Logic Signal Connections table (also shown as I/O grouping) shall not exceed a maximum of n × 8 mA. n is the number of I/O pads between the two consecutive bank V_{CCIO} or GND connections or between the last V_{CCIO} and GND in a bank and the end of a bank. I/O Grouping can be found in the Data Sheet Pin Summary Tables, which can also be generated from the Lattice Radiant software.

Table 3.17. I/O Resistance Characteristics (Over Recommended Operating Conditions)

Parameter	Description	Test Conditions	Min	Typ	Max	Unit
50RS	Output Drive Resistance when 50RS Drive Strength Selected	V_{CCIO} = 1.8 V, 2.5 V, or 3.3 V	—	50	—	Ω
R_{DIFF}	Input Differential Termination Resistance	Bank 3, Bank 4, and Bank 5 for I/O selected to be differential		100		Ω
SE Input Termination	Input Single Ended Termination Resistance	Bank 3, Bank 4, and Bank 5 for I/O selected to be Single Ended	36	40	64	Ω
			46	50	80	
			56	60	96	
			71	75	120	

sysI/O Differential DC Electrical Characteristics

LVDS

LVDS input buffer on Certus™-NX-RT is operating with V_{CCAUX} = 1.8 V and independent of Bank V_{CCIO} voltage. LVDS output buffer is powered by the Bank V_{CCIO} at 1.8 V.

LVDS can only be supported in Bank 3, Bank 4, and Bank 5. LVDS25 output can be emulated with LVDS25E in Bank 0, Bank 1, Bank 2, Bank 6, and Bank 7. This is described in LVDS25E (Output Only) section.

Table 3.18. LVDS DC Electrical Characteristics (Over Recommended Operating Conditions)

Parameter	Description	Test Conditions	Min	Typ	Max	Unit
V_{INP}, V_{INM}	Input Voltage	—	0	—	1.60	V
V_{ICM}	Input Common Mode Voltage	Half the sum of the two Inputs	0.05	—	1.55[2]	V
V_{THD}	Differential Input Threshold	Difference between the two Inputs	±100	—	—	mV
I_{IN}	Input Current	Power On or Power Off	—	—	±10	µA
V_{OH}	Output High Voltage for VOP or VOM	R_T = 100 Ω	—	1.425	1.60	V
V_{OL}	Output Low Voltage for VOP or VOM	R_T = 100 Ω	0.9	1.075	—	V
V_{OD}	Output Voltage Differential	(V_{OP} - V_{OM}), R_T = 100 Ω	250	350	450	mV
ΔV_{OD}	Change in VOD Between High and Low	—	—	—	50	mV
V_{OCM}	Output Common Mode Voltage	(V_{OP} + V_{OM})/2, R_T = 100 Ω	1.125	1.25	1.375	V
ΔV_{OCM}	Change in VOCM, VOCM(MAX) - VOCM(MIN)	—	—	—	50	mV
I_{SAB}	Output Short Circuit Current	V_{OD} = 0V Driver outputs shorted to each other	—	—	12	mA
ΔV_{OS}	Change in VOS between H and L	—	—	—	50	mV

Notes:

1. LVDS input or output are supported in Bank 3, Bank 4, and Bank 5. LVDS input uses V_{CCAUX} on the differential input comparator and can be located in any V_{CCIO} voltage bank. LVDS output uses V_{CCIO} on the differential output driver, and can only be located in bank with V_{CCIO} = 1.8 V.
2. V_{ICM} is depending on VID, input differential voltage, so the voltage on pin cannot exceed $V_{INP/INN}$(min/max) requirements. V_{ICM}(min) = $V_{INP/INN}$(min) + ½ VID, V_{ICM}(max) = $V_{INP/INN}$(max) – ½ VID. Values in the table is based on minimum VID of +/- 100 mV.

LVDS25E (Output Only)

Three sides of the Certus™-NX-RT devices, Top, Left and Right, support LVDS25 outputs with emulated complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The scheme shown in Figure 3.2 is one possible solution for point-to-point signals.

Table 3.19. LVDS25E DC Conditions

Parameter	Description	Typical	Unit
V_{CCIO}	Output Driver Supply (±5%)	2.50	V
Z_{OUT}	Driver Impedance	20	Ω
R_S	Driver Series Resistor (±1%)	158	Ω
R_P	Driver Parallel Resistor (±1%)	140	Ω
R_T	Receiver Termination (±1%)	100	Ω
V_{OH}	Output High Voltage	1.43	V
V_{OL}	Output Low Voltage	1.07	V
V_{OD}	Output Differential Voltage	0.35	V
V_{CM}	Output Common Mode Voltage	1.25	V
Z_{BACK}	Back Impedance	100.5	Ω
I_{DC}	DC Output Current	6.03	mA

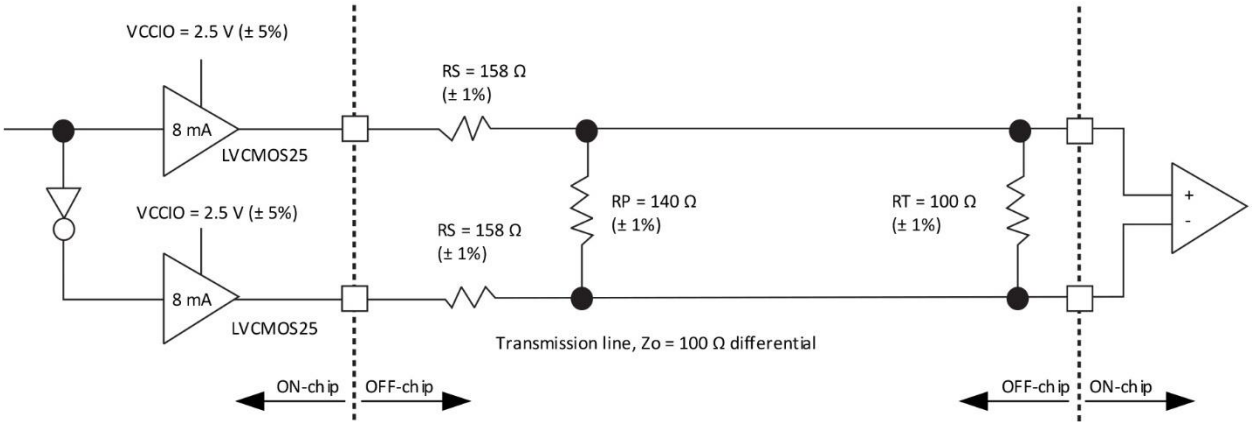


Figure 3.2. LVDS25E Output Termination Example

SubLVDS (Input Only)

SubLVDS is a reduced-voltage form of LVDS signaling, very similar to LVDS. It is a standard used in many camera types of applications, and follow the SMIA 1.0, Part 2: CCP2 Specification. Being similar to LVDS, the Certus™-NX-RT devices can support the subLVDS input signaling with the same LVDS input buffer. The output for subLVDS is implemented in subLVDSE/subLVDSEH with a pair of LVCMOS18 output drivers (see SubLVDSE/SubLVDSEH (Output Only) section).

Table 3.20. SubLVDS Input DC Electrical Characteristics (Over Recommended Operating Conditions)

Parameter	Description	Test Conditions	Min	Typ	Max	Unit
V_{ID}	Input Differential Threshold Voltage	Over V_{ICM} range	70	150	200	mV
V_{ICM}	Input Common Mode Voltage	Half the sum of the two Inputs	0.4	0.9	1.4	V

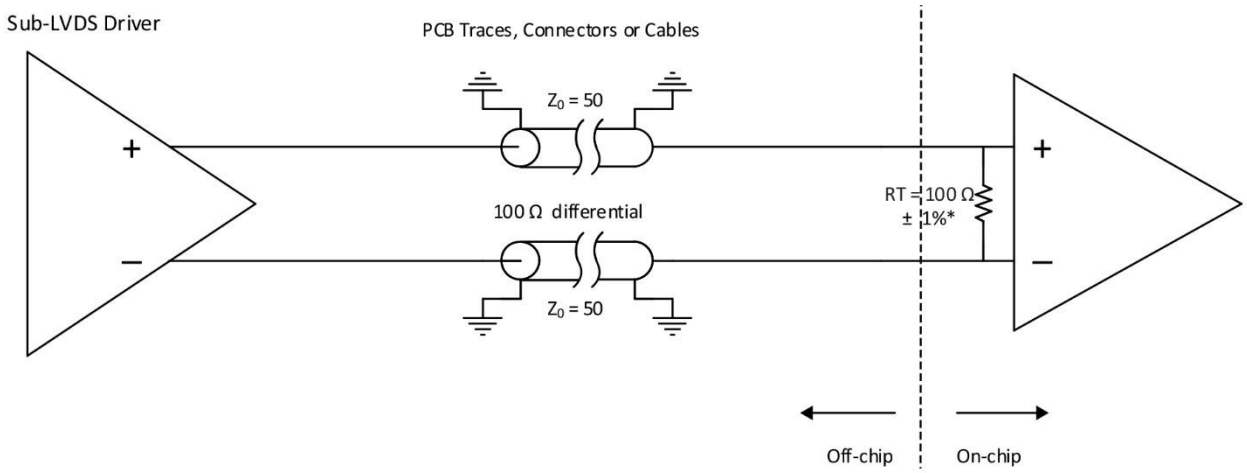


Figure 3.3. SubLVDS Input Interface

SubLVDSE/SubLVDSEH (Output Only)

SubLVDS output uses a pair of LVCMOS18 drivers with True and Complement outputs. The V_{CCIO} of the bank used for subLVDSE or subLVDSEH needs to be powered by 1.8V. SubLVDSE is for Bank 0, Bank 1, Bank 2, Bank 5, and Bank 6; and subLVDSEH is for Bank 3, Bank 4, and Bank 5.

Performance of the subLVDSE/subLVDSEH driver is limited to the performance of LVCMOS18.

Table 3.21. SubLVDS Output DC Electrical Characteristics (Over Recommended Operating Conditions)

Parameter	Description	Test Conditions	Min	Typ	Max	Unit
V_{OD}	Output Differential Voltage Swing	—	—	150	—	mV
V_{OCM}	Output Common Mode Voltage	Half the sum of the two Outputs	—	0.9	—	V

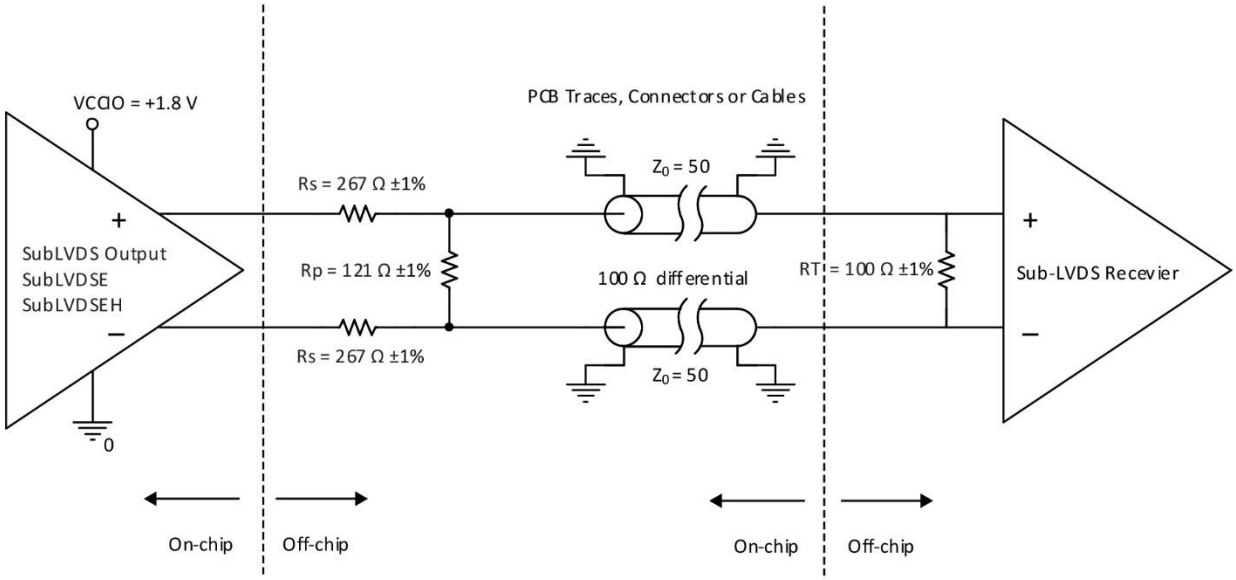


Figure 3.4. SubLVDS Output Interface

SLVS

Scalable Low-Voltage Signaling (SLVS) is based on a point-to-point signaling method defined in the JEDEC JESD8-13 (SLVS-400) standard. This standard evolved from the traditional LVDS standard with smaller voltage swings and a lower common-mode voltage. The 200 mV (400 mV p-p) SLVS swing contributes to a reduction in power.

The Certus™-NX-RT devices receive SLVS differential input with the LVDS input buffer. This LVDS input buffer is design to cover wide input common mode range that can meet the SLVS input standard specified by the JEDEC standard.

Table 3.22. SLVS Input DC Characteristics (Over Recommended Operating Conditions)

Parameter	Description	Test Conditions	Min	Typ	Max	Unit
V_{ID}	Input Differential Threshold Voltage	Over V_{ICM} range	70	—	—	mV
V_{ICM}	Input Common Mode Voltage	Half the sum of the two Inputs	70	200	330	mV

The SLVS output on Certus™-NX-RT is supported with the LVDS drivers found in Bank 3, Bank 4, and Bank 5. The LVDS driver on Certus™-NX-RT is a current controlled driver. It can be configured as LVDS driver or configured with the 100 Ω differential termination with center-tap set to VOCM at 200 mV. This means the differential output driver can be placed into bank with V_{CCIO} = 1.2 V, 1.5 V, or 1.8 V, even if it is powered by V_{CCIO}.

Table 3.23. SLVS Output DC Characteristics (Over Recommended Operating Conditions)

Parameter	Description	Test Conditions	Min	Typ	Max	Unit
V_{CCIO}	Bank V_{CCIO}	—	–5%	1.2, 1.5, 1.8	+ 5%	V
V_{OD}	Output Differential Voltage Swing	—	140	200	270	mV
V_{OCM}	Output Common Mode Voltage	Half the sum of the two Outputs	150	200	250	mV
Z_{OS}	Single-Ended Output Impedance	—	37.5	50	62.5	Ω

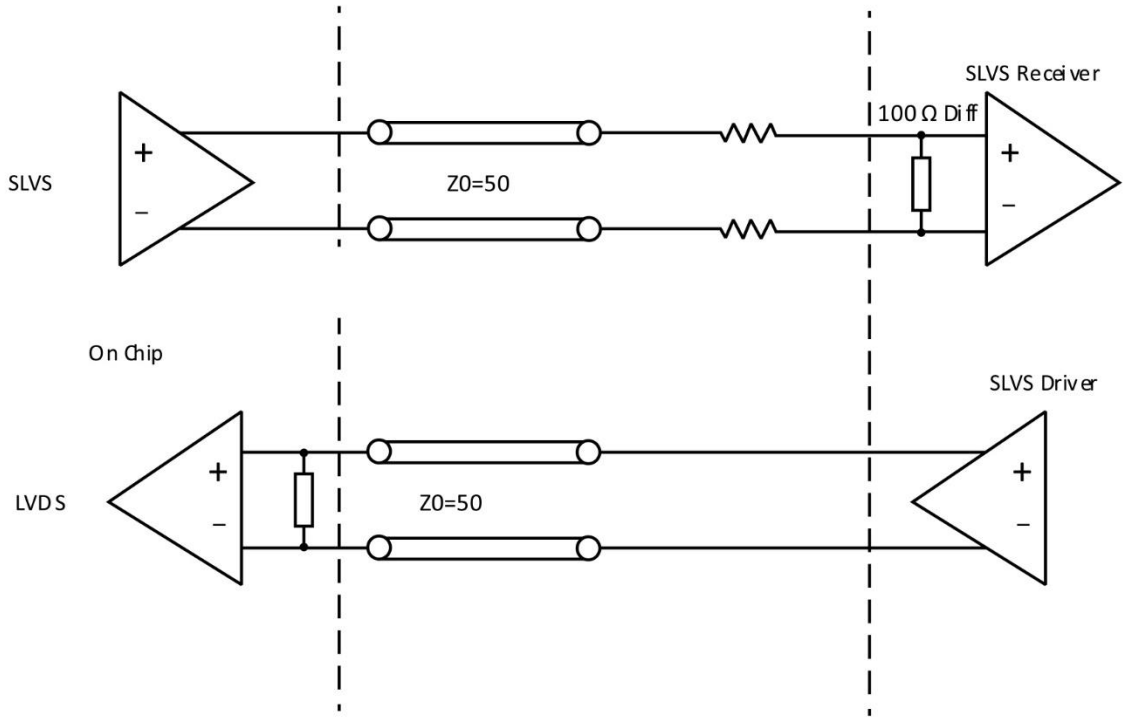


Figure 3.5. SLVS Interface

Differential HSTL15D (Output Only)

Differential HSTL outputs are implemented as a pair of complementary single-ended HSTL outputs.

Differential SSTL135D, SSTL15D (Output Only)

Differential SSTL is used for differential clock in DDR3/DDR3L memory interface. All differential SSTL outputs are implemented as a pair of complementary single-ended SSTL outputs. All allowable single-ended output classes (class I and class II) are supported.

Differential HSUL12D (Output Only)

Differential HSUL is used for differential clock in LPDDR2/LPDDR3 memory interface. All differential HSUL outputs are implemented as a pair of complementary single-ended HSUL12 outputs. All allowable single-ended drive strengths are supported.

Differential LVCMOS25D, LVCMOS33D, LVTTL33D (Output Only)

Differential LVCMOS and LVTTL outputs are implemented as a pair of complementary single-ended outputs. All allowable single-ended output drive strengths are supported.

Certus™-NX-RT Maximum sysI/O Buffer Speed

Over recommended operating conditions.

Table 3.24. Certus™-NX-RT Maximum I/O Buffer Speed[1,2,3,4,7]

Parameter	Description	Banks	Max	Unit
Maximum sysI/O Input Frequency				
Single-Ended				
LVCMOS33	LVCMOS33, V_{CCIO} = 3.3 V	0, 1, 2, 6, 7	200	MHz
LVTTL33	LVTTL33, V_{CCIO} = 3.3 V	0, 1, 2, 6, 7	200	MHz
LVCMOS25	LVCMOS25, V_{CCIO} = 2.5 V	0, 1, 2, 6, 7	200	MHz
LVCMOS18[5]	LVCMOS18, V_{CCIO} = 1.8 V	0, 1, 2, 6, 7	200	MHz
LVCMOS18H	LVCMOS18, V_{CCIO} = 1.8 V	3, 4, 5	200	MHz
LVCMOS15[5]	LVCMOS15, V_{CCIO} = 1.5 V	0, 1, 2, 6, 7	100	MHz
LVCMOS15H[5]	LVCMOS15, V_{CCIO} = 1.5 V	3, 4, 5	150	MHz
LVCMOS12[5]	LVCMOS12, V_{CCIO} = 1.2 V	0, 1, 2, 6, 7	50	MHz
LVCMOS12H[5]	LVCMOS12, V_{CCIO} = 1.2 V	3, 4, 5	100	MHz
LVCMOS10[5]	LVCMOS 1.0, V_{CCIO} = 1.2 V	0, 1, 2, 6, 7	50	MHz
LVCMOS10H[5]	LVCMOS 1.0, V_{CCIO} = 1.0 V	3, 4, 5	50	MHz
LVCMOS10R	LVCMOS 1.0, V_{CCIO} independent	3, 4, 5	50	MHz
SSTL15_I, SSTL15_II	SSTL_15, V_{CCIO} = 1.5 V	3, 4, 5	1066	Mbps
SSTL135_I, SSTL135_II	SSTL_135, V_{CCIO} = 1.35 V	3, 4, 5	1066	Mbps
HSUL12	HSUL_12, V_{CCIO} = 1.2 V	3, 4, 5	1066	Mbps
HSTL15	HSTL15, V_{CCIO} = 1.5 V	3, 4, 5	250	Mbps
Differential				
LVDS	LVDS, V_{CCIO} independent	3, 4, 5	1250	Mbps
subLVDS	subLVDS, V_{CCIO} independent	3, 4, 5	1250	Mbps
SLVS	SLVS similar to MIPI HS, V_{CCIO} independent	3, 4, 5	1250	Mbps
SSTL15D	Differential SSTL15, V_{CCIO} independent	3, 4, 5	1066	Mbps
SSTL135D	Differential SSTL135, V_{CCIO} independent	3, 4, 5	1066	Mbps
HUSL12D	Differential HSUL12, V_{CCIO} independent	3, 4, 5	1066	Mbps
HSTL15D	Differential HSTL15, V_{CCIO} independent	3, 4, 5	250	Mbps
Maximum sysI/O Output Frequency				
Single-Ended				
LVCMOS33 (all drive strengths)	LVCMOS33, V_{CCIO} = 3.3 V	0, 1, 2, 6, 7	200	MHz
LVCMOS33 (RS50)	LVCMOS33, V_{CCIO} = 3.3 V, RSERIES = 50 Ω	0, 1, 2, 6, 7	200	MHz
LVTTL33 (all drive strengths)	LVTTL33, V_{CCIO} = 3.3 V	0, 1, 2, 6, 7	200	MHz
LVTTL33 (RS50)	LVTTL33, V_{CCIO} = 3.3 V, RSERIES = 50 Ω	0, 1, 2, 6, 7	200	MHz
LVCMOS25 (all drive strengths)	LVCMOS25, V_{CCIO} = 2.5 V	0, 1, 2, 6, 7	200	MHz
LVCMOS25 (RS50)	LVCMOS25, V_{CCIO} = 2.5 V, RSERIES = 50 Ω	0, 1, 2, 6, 7	200	MHz
LVCMOS18 (all drive strengths)	LVCMOS18, V_{CCIO} = 1.8 V	0, 1, 2, 6, 7	200	MHz
LVCMOS18 (RS50)	LVCMOS18, V_{CCIO} = 1.8 V, RSERIES = 50 Ω	0, 1, 2, 6, 7	200	MHz

Parameter	Description	Banks	Max	Unit
LVCMOS18H (all drive strengths)	LVCMOS18, V_{CCIO} = 1.8 V	3, 4, 5	200	MHz
LVCMOS18H (RS50)	LVCMOS18, V_{CCIO} = 1.8 V, RSERIES = 50 Ω	3, 4, 5	200	MHz
LVCMOS15 (all drive strengths)	LVCMOS15, V_{CCIO} = 1.5 V	0, 1, 2, 6, 7	100	MHz
LVCMOS15H (all drive strengths)	LVCMOS15, V_{CCIO} = 1.5 V	3, 4, 5	150	MHz
LVCMOS12 (all drive strengths)	LVCMOS12, V_{CCIO} = 1.2 V	0, 1, 2, 6, 7	50	MHz
LVCMOS12H (all drive strengths)	LVCMOS12, V_{CCIO} = 1.2 V	3, 4, 5	100	MHz
LVCMOS10H (all drive strengths)	LVCMOS12, V_{CCIO} = 1.2 V	3, 4, 5	50	MHz
SSTL15_I, SSTL15_II	SSTL_15, V_{CCIO} = 1.5 V	3, 4, 5	1066	Mbps
SSTL135_I, SSTL135_II	SSTL_135, V_{CCIO} = 1.35 V	3, 4, 5	1066	Mbps
HSUL12 (all drive strengths)	HSUL_12, V_{CCIO} = 1.2 V	3, 4, 5	1066	Mbps
HSTL15	HSTL15, V_{CCIO} = 1.5 V	3, 4, 5	250	Mbps
Differential				
LVDS	LVDS, V_{CCIO} = 1.8 V	3, 4, 5	1250	Mbps
LVDS25E[6]	LVDS25, Emulated V_{CCIO} = 2.5 V	0, 1, 2, 6, 7	400	Mbps
SubLVDSE[6]	subLVDS, Emulated, V_{CCIO} = 1.8 V	0, 1, 2, 6, 7	400	Mbps
SubLVDSEH[6]	subLVDS, Emulated, V_{CCIO} = 1.8 V	3, 4, 5	800	Mbps
SLVS	SLVS similar to MIPI, V_{CCIO} = 1.2 V	3, 4, 5	1250	Mbps
SSTL15D	Differential SSTL15, V_{CCIO} = 1.5 V	3, 4, 5	1066	Mbps
SSTL135D	Differential SSTL135, V_{CCIO} = 1.35 V	3, 4, 5	1066	Mbps
HUSL12D	Differential HSUL12, V_{CCIO} = 1.2 V	3, 4, 5	1066	Mbps
HSTL15D	Differential HSTL15, V_{CCIO} = 1.5 V	3, 4, 5	250	Mbps

Notes:

1. Maximum I/O speed is the maximum switching rate of the I/O operating within the guidelines of the defining standard. The actual interface speed performance using the I/O also depends on other factors, such as internal and external timing.
2. These numbers are characterized but not test on every device.
3. Performance is specified in MHz, as defined in clock rate when the sysI/O is used as pin. For data rate performance, this can be converted to Mbps, which equals to 2 times the clock rate.
4. LVCMOS and LVTTL are measured with load specified in Table 3.40.
5. These LVCMOS inputs can be placed in different V_{CCIO} voltage. Performance may vary. Please refer to Lattice Design Software
6. These emulated outputs performance is based on externally properly terminated as described in LVDS25E (Output Only) and SubLVDSE/SubLVDSEH (Output Only).
7. All speeds are measured with fast slew.
8. For maximum differential I/O performance only Differential I/O should be placed in the bottom I/O banks. If this is not possible, the following will impact on maximum performance:
   a. If Fast Slew Rate LVCMOS I/O are used, they should be limited to no more than nine I/O (adjacent), four I/O (same bank), 55 I/O (left/right banks) to keep degradation below 50%.
   b. If non-Differential I/O (SLOW SLEW) are placed on the bottom but not within the same bank as differential I/O, then the maximum Differential performance is degraded to 70% of original when 21 aggressors are toggling.
   c. If non-Differential I/O (SLOW SLEW) are placed within the same bank as Differential I/O then the maximum performance is degraded to 50% of original when 16 aggressor are toggling.
   d. No performance impact if MIPI LP and MIPI HS are in the same bank.
   e. If Differential RX/TX I/O are both placed within the same bank then the maximum performance is degraded to 90%.
   f. For DDR3/3L, LPDDR2/3 separate DQ/DQS groups from Address/Commands/CLK groups into separate banks.

Typical Building Block Function Performance

These building block functions can be generated using Lattice Design Software Tool. Exact performance may vary with the device and the design software tool version. The design software tool uses internal parameters that have been characterized but are not tested on every device.

Table 3.25. Pin-to-Pin Performance

Function	Typ. @ V_{CC} = 1.0 V	Unit
16-bit Decoder (I/O configured with LVCMOS18, Left and Right Banks)	5.5	ns
16-bit Decoder (I/O configured with HSTL15_I, Bottom Banks)	5.1	ns
16:1 Mux (I/O configured with LVCMOS18, Left and Right Banks)	6	ns
16:1 Mux (I/O configured with HSTL15_I, Bottom Banks)	6.1	ns

Notes:

1. These functions are generated using Lattice Radiant Design Software tool. Exact performance may vary with the device and the design software tool version. The design software tool uses internal parameters that have been characterized but are not tested on every device.

Table 3.26. Register-to-Register Performance

Function	Typ. @ V_{CC} = 1.0 V	Unit
Basic Functions		
16-bit Adder	5002	MHz
32-bit Adder	496	MHz
16-bit Counter	402	MHz
32-bit Counter	371	MHz
Embedded Memory Functions		
512 × 36 Single Port RAM, with Output Register	5002	MHz
1024 × 18 True-Dual Port RAM using same clock, with EBR Output Registers	5002	MHz
1024 × 18 True-Dual Port RAM using asynchronous clocks, with EBR Output Registers	5002	MHz
Large Memory Functions		
32 k × 32 Single Port RAM, with Output Register	1652	MHz
32 k × 32 Single Port RAM with ECC, with Output Register	1302	MHz
32 k × 32 True-Dual Port RAM using same clock, with Output Registers	340	MHz
Distributed Memory Functions		
16 × 4 Single Port RAM (One PFU)	5002	MHz
16 × 2 Pseudo-Dual Port RAM (One PFU)	5002	MHz
16 × 4 Pseudo-Dual Port (Two PFUs)	5002	MHz
DSP Functions		
9 × 9 Multiplier with Input Output Registers	351	MHz
9 × 9 Multiplier with Input/Pipelined/Output Registers	240	MHz
18 × 18 Multiplier with Input/Output Registers	214	MHz
18 × 18 Multiplier with Input/Pipelined/Output Registers	191	MHz
36 × 36 Multiplier with Input/Output Registers	201	MHz
36 × 36 Multiplier with Input/Pipelined/Output Registers	129	MHz

Function	Typ. @ V_{CC} = 1.0 V	Unit
MAC 9 × 9 with Input/Output Registers	218	MHz
MAC 9 × 9 with Input/Pipelined/Output Registers	238	MHz

Notes:

1. The Clock port is configured with LVDS I/O type. Performance Grade: 9_High-Performance_1.0V.
2. Limited by the Minimum Pulse Width of the component
3. These functions are generated using Lattice Radiant Design Software tool. Exact performance may vary with the device and the design software tool version. The design software tool uses internal parameters that have been characterized but are not tested on every device.
4. For the Pipelined designs, the number of pipeline stages used are 2.

LMMI

Table 3.27 summarizes the performance of the LMMI interface with supported IPs. Additional timing requirement and constraint can be identified through the Lattice Radiance design tools.

Table 3.27. LMMI F_{MAX} Summary

IP	F_{MAX} (MHz)
CDR0	73
CDR1	70
DPHY0	67
DPHY1	55
CRE	54
I^2C	38
PCIe	57
PLL_ULC	59
PLL_LLC	55
PLL_LRC	37

Derating Timing Tables

Logic timing provided in the following sections of this data sheet and the Lattice Radiant design tools are worst case numbers in the operating range. Actual delays at nominal temperature and voltage for best case process, can be much better than the values given in the tables. The Lattice Radiant design tool can provide logic timing numbers at a particular temperature and voltage.

Certus™-NX-RT External Switching Characteristics

Over recommended commercials operating conditions.

Table 3.28 Certus™-NX-RT External Switching Characteristics (V_{CC} = 1.0 V)

Parameter	Description	Min	Max	Unit
Clocks				
Primary Clock				
f_{MAX_PRI}	Frequency for Primary Clock	—	325.2	MHz
t_{W_PRI}	Clock Pulse Width for Primary Clock	1.35	—	ns
t_{SKEW_PRI}	Primary Clock Skew Within a Device	—	554	ps
Edge Clock				
f_{MAX_EDGE}	Frequency for Edge Clock Tree	—	650.4	MHz
t_{W_EDGE}	Clock Pulse Width for Edge Clock	0.646	—	ns
t_{SKEW_EDGE}	Edge Clock Skew Within a Device	—	148	ps
Generic SDR Input				
General I/O Pin Parameters Using Dedicated Primary Clock Input without PLL				
t_{CO}	Clock to Output – PIO Output Register	—	6.91	ns
t_{SU}	Clock to Data Setup – PIO Input Register	0	—	ns
t_{H}	Clock to Data Hold – PIO Input Register	3.35	—	ns
t_{SU_DEL}	Clock to Data Setup – PIO Input Register with Data Input Delay	1.86	—	ns
t_{H_DEL}	Clock to Data Hold – PIO Input Register with Data Input Delay	0.26	—	ns
General I/O Pin Parameters Using Dedicated Primary Clock Input with PLL				
t_{COPLL}	Clock to Output – PIO Output Register	—	4.72	ns
t_{SUPLL}	Clock to Data Setup – PIO Input Register	1.31	—	ns
t_{HPLL}	Clock to Data Hold - PIO Input Register	1.22	—	ns
t_{SU_DELPLL}	Clock to Data Setup - PIO Input Register with Data Input Delay	4.99	—	ns
t_{H_DELPLL}	Clock to Data Hold - PIO Input Register with Data Input Delay	0	—	ns
Generic DDR Input/Output				
Generic DDRX1 Inputs/Outputs with Clock and Data Centered at Pin (GDDRX1_RX/TX.SCLK.Centered) using PCLK Clock Input – Figure3.6 and Figure 3.8				
t_{SU_GDDR1}	Input Data Setup Before CLK	0.917	—	ns
		0.275	—	UI
t_{HO_GDDR1}	Input Data Hold After CLK	0.917	—	ns
		0.275	—	UI
t_{DVB_GDDR1}	Output Data Valid After CLK Output	1.108	—	ns
		–0.559	—	ns + ½ UI
t_{DQVA_GDDR1}	Output Data Valid After CLK Output	1.108	—	ns
		–0.559	—	ns + ½ UI

Parameter	Description	Min	Max	Unit
f_{DATA_GDDRX1}	Input/Output Data Rate	—	300	Mbps
f_{MAX_GDDRX1}	Frequency of PCLK	—	150	MHz
½ UI	Half of Data Bit Time, or 90-degree	1.667	—	ns
Output TX to Input RX Margin per Edge		0.191	—	ns
Generic DDRX1 Inputs/Outputs with Clock and Data Aligned at Pin (GDDRX1_RX/TX.SCLK.Aligned) using PCLK Clock Input – Figure 3.7 and Figure 3.9				
t_{DVA_GDDR1}	Input Data Valid After CLK	—	–0.917	ns + 1/2 UI
		—	0.75	ns
		—	0.225	UI
t_{DVE_GDDR1}	Input Data Hold After CLK	0.917	—	ns + 1/2 UI
		2.583	—	ns
		0.775	—	UI
t_{DIA_GDDR1}	Output Data Invalid After CLK Output	—	0.559	ns
t_{DIB_GDDR1}	Output Data Invalid Before CLK Output	—	0.559	ns
f_{DATA_GDDRX1}	Input/Output Data Rate	—	300	Mbps
f_{MAX_GDDRX1}	Frequency for PCLK	—	150	MHz
½ UI	Half of Data Bit Time, or 90 degree	1.667	—	ns
Output TX to Input RX Margin per Edge		0.191	—	ns
Generic DDRX1 Inputs/Outputs with Clock and Data Aligned at Pin (GDDRX1_RX/TX.SCLK.Aligned) using PCLK Clock Input – Figure 3.6 and Figure 3.8				
t_{SU_GDDRX2}	Data Setup before CLK Input	0.270	ns	ns
		0.162	UI	UI
t_{HO_GDDRX2}	Data Hold after CLK Input	0.270	ns	ns
t_{DVB_GDDRX2}	Output Data Valid Before CLK Output	0.684	ns	ns
		–0.149	ns + 1/2 UI	ns + 1/2 UI
t_{DQVA_GDDRX2}	Output Data Valid After CLK Output	0.684	ns	ns
		–0.149	ns + 1/2 UI	ns + 1/2 UI
f_{DATA_GDDRX2}	Input/Output Data Rate	—	Mbps	Mbps
f_{MAX_GDDRX2}	Frequency for ECLK	—	MHz	MHz
½ UI	Half of Data Bit Time, or 90 degree	0.833	ns	ns
fPCLK	PCLK frequency	—	MHz	MHz
Output TX to Input RX Margin per Edge		0.434	—	ns
Generic DDRX2 Inputs/Outputs with Clock and Data Aligned at Pin (GDDRX2_RX/TX.ECLK.Aligned) using PCLK Clock Input – Figure 3.7 and Figure 3.9				
t_{DVA_GDDRX2}	Input Data Valid After CLK	—	–0.458	ns + 1/2 UI
		—	0.375	ns
		—	0.225	UI
t_{DVE_GDDRX2}	Input Data Hold After CLK	0.458	—	ns + 1/2 UI
		1.292	—	ns
		0.775	—	UI
t_{DIA_GDDRX2}	Output Data Invalid After CLK Output	—	0.149	ns
t_{DIB_GDDRX2}	Output Data Invalid Before CLK Output	—	0.149	ns

Parameter	Description	Min	Max	Unit
f_{DATA_GDDRX2}	Input/Output Data Rate	—	600	Mbps
f_{MAX_GDDRX2}	Frequency for ECLK	—	300	MHz
½ UI	Half of Data Bit Time, or 90 degree	0.833	—	ns
f_{PCLK}	PCLK frequency	—	247.5	MHz
Output TX to Input RX Margin per Edge		0.266	—	ns
Generic DDRX4 Inputs/Outputs with Clock and Data Centered at Pin (GDDRX4_RX/TX.ECLK.Centered) using PCLK Clock Input – Figure 3.6 and Figure 3.8				
t_{SU_GDDRX4}	Input Data Set-Up Before CLK	0.220	—	ns
		0.220	—	UI
t_{HO_GDDRX4}	Input Data Hold After CLK	0.220	—	ns
t_{DVB_GDDRX4}	Output Data Valid Before CLK Output	0.351	—	ns
		–0.148	—	UI
t_{DQVA_GDDRX4}	Input/Output Data Rate	0.351	—	ns
		–0.148	—	UI
f_{DATA_GDDRX4}	Frequency for ECLK	—	1000	Mbps
f_{MAX_GDDRX4}	PCLK frequency	—	500	MHz
½ UI	Half of Data Bit Time, or 90-degree	0.5	—	ns
f_{PCLK}	Input Data Set-Up Before CLK	—	125	MHz
Output TX to Input RX Margin per Edge		0.151	—	ns
Generic DDRX4 Inputs/Outputs with Clock and Data Aligned at Pin (GDDRX4_RX/TX.ECLK.Aligned) using PCLK Clock Input, Left and Right sides Only – Figure 3.7 and Figure 3.9				
t_{DVA_GDDRX4}	Input Data Valid After CLK	—	–0.275	ns + 1/2 UI
		—	0.225	ns
		—	0.225	UI
t_{DVE_GDDRX4}	Input Data Hold After CLK	0.275	—	ns + 1/2 UI
		0.775	—	ns
		0.775	—	UI
t_{DIA_GDDRX4}	Output Data Invalid After CLK Output	—	0.149	ns
t_{DIB_GDDRX4}	Output Data Invalid Before CLK Output	—	0.149	ns
f_{DATA_GDDRX4}	Input/Output Data Rate	—	1000	Mbps
f_{MAX_GDDRX4}	Frequency for ECLK	—	500	MHz
½ UI	Half of Data Bit Time, or 90 degree	0.5	—	ns
f_{PCLK}	PCLK frequency	—	125	MHz
Output TX to Input RX Margin per Edge		0.076	—	ns
Generic DDRX5 Inputs/Outputs with Clock and Data Centered at Pin (GDDRX5_RX/TX.ECLK.Centered) using PCLK Clock Input – Figure 3.6 and Figure 3.8				
t_{SU_GDDRX5}	Input Data Set-Up Before CLK	0.22	—	ns
		0.22	—	UI

Parameter	Description	Min	Max	Unit
t_{HO_GDDRX5}	Input Data Hold After CLK	0.22	—	ns
$t_{WINDOW\_GDDRX5C}$	Input Data Valid Window	0.44	—	ns
t_{DVB_GDDRX5}	Output Data Valid Before CLK Output	0.351	—	ns
		–0.149	—	ns+1/2UI
t_{DQVA_GDDRX5}	Output Data Valid After CLK Output	0.351	—	ns
		–0.149	—	ns+1/2UI
f_{DATA_GDDRX5}	Input/Output Data Rate	—	1000	Mbps
f_{MAX_GDDRX5}	Frequency for ECLK	—	500	MHz
½ UI	Half of Data Bit Time, or 90 degree	0.5	—	ns
f_{PCLK}	PCLK frequency	—	100	MHz
Output TX to Input RX Margin per Edge		0.151	—	ns
Generic DDRX5 Inputs/Outputs with Clock and Data Aligned at Pin (GDDRX5_RX/TX.ECLK.Aligned) using PCLK Clock Input, Left and Right sides Only – Figure 3.7 and Figure 3.9				
t_{DVA_GDDRX5}	Input Data Valid After CLK	—	–0.275	ns + 1/2 UI
		—	0.225	ns
		—	0.225	UI
t_{DVE_GDDRX5}	Input Data Hold After CLK	0.275	—	ns + 1/2 UI
		0.775	—	ns
		0.775	—	UI
$t_{WINDOW\_GDDRX5A}$	Input Data Valid Window	0.55	—	ns
t_{DIA_GDDRX5}	Output Data Invalid After CLK Output	—	0.149	ns
t_{DIB_GDDRX5}	Output Data Invalid Before CLK Output	—	0.149	ns
f_{DATA_GDDRX5}	Input/Output Data Rate	—	1000	Mbps
f_{MAX_GDDRX5}	Frequency for ECLK	—	500	MHz
½ UI	Half of Data Bit Time, or 90 degree	0.5	—	ns
f_{PCLK}	PCLK frequency	—	100	MHz
Output TX to Input RX Margin per Edge		0.076	—	ns
Soft D-PHY DDRX4 Inputs/Outputs with Clock and Data Centered at Pin, using PCLK Clock Input				
$t_{SU\_GDDRX4\_MP}$	Input Data Set-Up Before CLK	0.21	—	ns
		0.21	—	UI
$t_{HO\_GDDRX4\_MP}$	Input Data Hold After CLK	0.2	—	ns
		0.2	—	UI
$t_{DVB\_GDDRX4\_MP}$	Output Data Valid Before CLK Output	0.3	—	ns
		0.3	—	UI
$t_{DQVA\_GDDRX4\_MP}$	Output Data Valid After CLK Output	0.3	—	ns
		0.3	—	UI
$f_{DATA\_GDDRX4\_MP}$	Input Data Bit Rate for MIPI PHY	—	1000	Mbps
½ UI	Half of Data Bit Time, or 90 degree	0.5	—	ns
f_{PCLK}	PCLK frequency	—	125	MHz
Output TX to Input RX Margin per Edge		0.1	—	ns

Parameter	Description	Min	Max	Unit
Video DDRX71 Inputs/Outputs with Clock and Data Aligned at Pin (GDDRX71_RX.ECLK) using PLL Clock Input – Figure 3.11 and Figure 3.12				
t_{RPBi_DVA}	Input Valid Bit "i" switch from CLK Rising Edge ("i" = 0 to 6, 0 aligns with CLK)	—	0.237	UI
		—	–0.278	ns+(1/2+i)*UI
t_{RPBi_DVE}	Input Hold Bit "i" switch from CLK Rising Edge ("i" = 0 to 6, 0 aligns with CLK)	0.748	—	UI
		0.263	—	ns+(1/2+i)*UI
t_{TPBi_DOV}	Data Output Valid Bit "i" switch from CLK Rising Edge ("i" = 0 to 6, 0 aligns with CLK)	—	0.159	ns+i*UI
t_{TPBi_DOI}	Data Output Invalid Bit "i" switch from CLK Rising Edge ("i" = 0 to 6, 0 aligns with CLK)	-0.159	—	ns+(i+ 1)*UI
$t_{TPBi\_skew\_UI}$	TX skew in UI	—	0.150	UI
t_B	Serial Data Bit Time, = 1UI	1.058	—	ns
f_{DATA_TX71}	DDR71 Serial Data Rate	—	945	Mbps
f_{MAX_TX71}	DDR71 ECLK Frequency	—	472.5	MHz
f_{CLKIN}	7:1 Clock (PCLK) Frequency	—	133.7	MHz
Output TX to Input RX Margin per Edge		0.159	—	ns
Memory Interface				
DDR3/DDR3L/LPDDR2/LPDDR3 READ (DQ Input Data are Aligned to DQS) – Figure 3.7				
tDVBDQ_DDR3 tDVBDQ_DDR3L tDVBDQ_LPDDR2 tDVBDQ_LPDDR3	Data Input Valid before DQS Input	—	–0.235	ns + 1/2 UI
tDVADQ_DDR3 tDVADQ_DDR3L tDVADQ_LPDDR2 tDVADQ_LPDDR3	Data Input Valid after DQS Input	0.235	—	ns + 1/2 UI
fDATA_DDR3 fDATA_DDR3L fDATA_LPDDR2 fDATA_LPDDR3	DDR Memory Data Rate	—	1066	Mb/s
fMAX_ECLK_DDR3 fMAX_ECLK_DDR3L fMAX_ECLK_LPDDR2 fMAX_ECLK_LPDDR3	DDR Memory ECLK Frequency	—	533	MHz
fMAX_SCLK_DDR3 fMAX_SCLK_DDR3L fMAX_SCLK_LPDDR2 fMAX_SCLK_LPDDR3	DDR Memory SCLK Frequency	—	133.3	MHz
DDR3/DDR3L/LPDDR2/LPDDR3 WRITE (DQ Output Data are Centered to DQS) – Figure 3.10				
tDQVBS_DDR3 tDQVBS_DDR3L tDQVBS_LPDDR2 tDQVBS_LPDDR3	Data Output Valid before DQS Output	—	–0.235	ns + 1/2 UI
tDQVAS_DDR3 tDQVAS_DDR3L tDQVAS_LPDDR2 tDQVAS_LPDDR3	Data Output Valid after DQS Output	0.235	—	ns + 1/2 UI

Parameter	Description	Min	Max	Unit
fDATA_DDR3 fDATA_DDR3L fDATA_LPDDR2 fDATA_LPDDR3	DDR Memory Data Rate	—	1066	Mb/s
fMAX_ECLK_DDR3 fMAX_ECLK_DDR3L fMAX_ECLK_LPDDR2 fMAX_ECLK_LPDDR3	DDR Memory ECLK Frequency	—	533	MHz
fMAX_SCLK_DDR3 fMAX_SCLK_DDR3L fMAX_SCLK_LPDDR2 fMAX_SCLK_LPDDR3	DDR Memory SCLK Frequency	—	133.3	MHz

Notes:

1. Commercial timing numbers are shown. Industrial numbers are typically slower and can be extracted from the Lattice Radiant software.
2. General I/O timing numbers are based on LVCMOS 1.8, 8 mA, Fast Slew Rate, 0 pf load.
3. Generic DDR timing are numbers based on LVDS I/O.
4. DDR3 timing numbers are based on SSTL15.
5. LPDDR2 and LPDDR3 timing numbers are based on HSUL12.
6. Uses LVDS I/O standard for measurements.
7. Maximum clock frequencies are tested under best case conditions. System performance may vary upon the user environment.
8. All numbers are generated with the Lattice Radiant software.

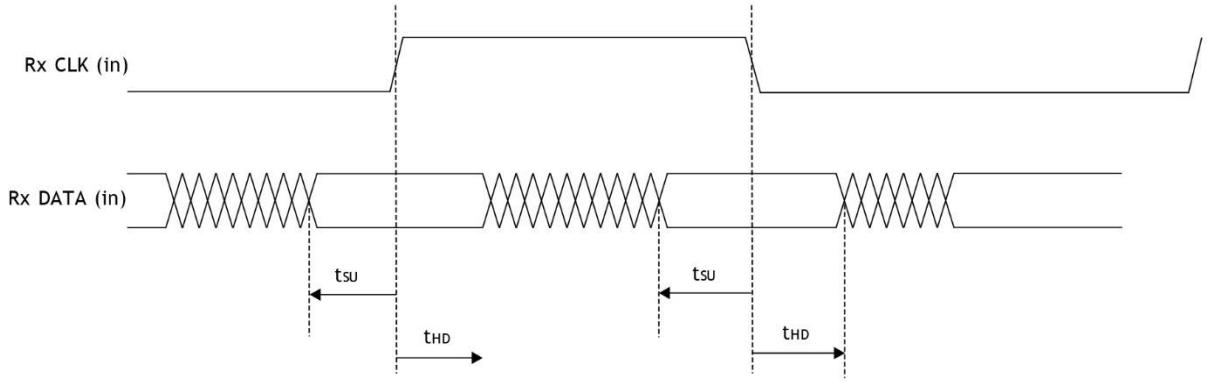


Figure 3.6. Receiver RX.CLK.Centered Waveforms

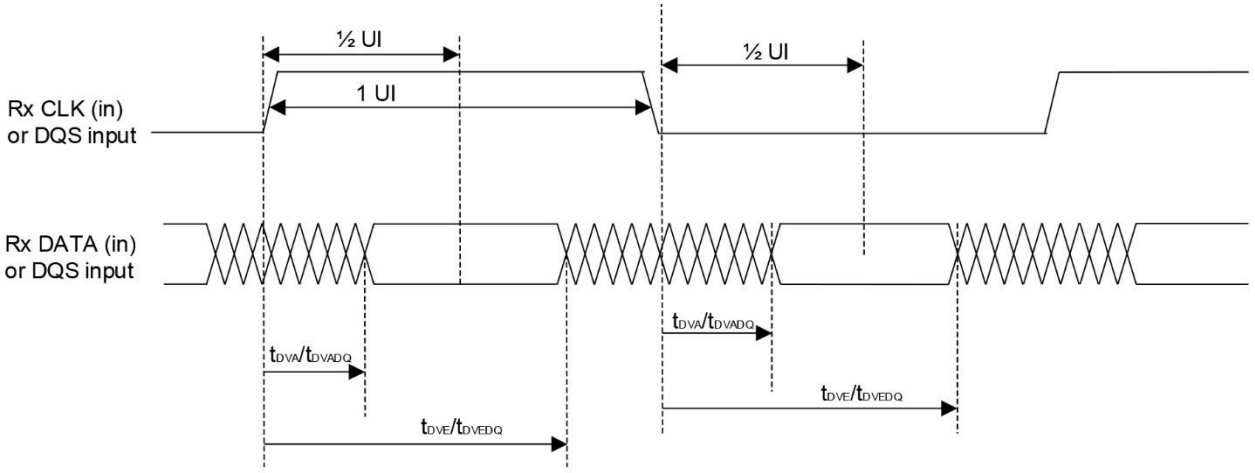


Figure 3.7. Receiver RX.CLK. Aligned and DDR Memory Input Waveforms

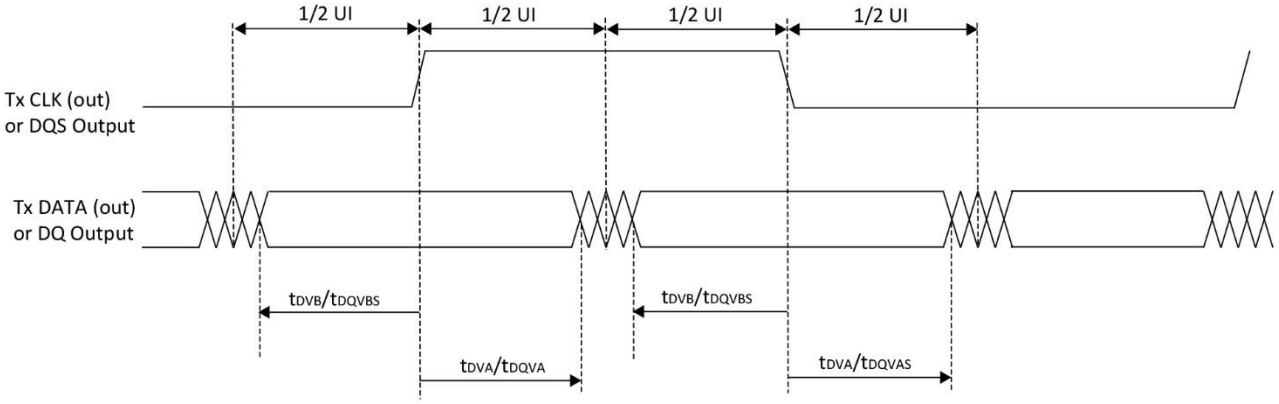


Figure 3.8. Transmit TX.CLK. Centered and DDR Memory Output Waveforms

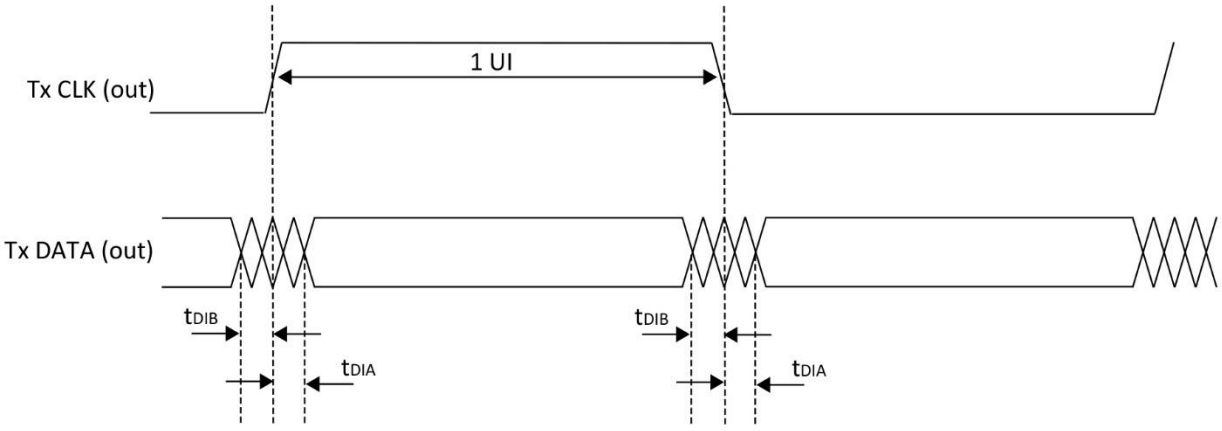
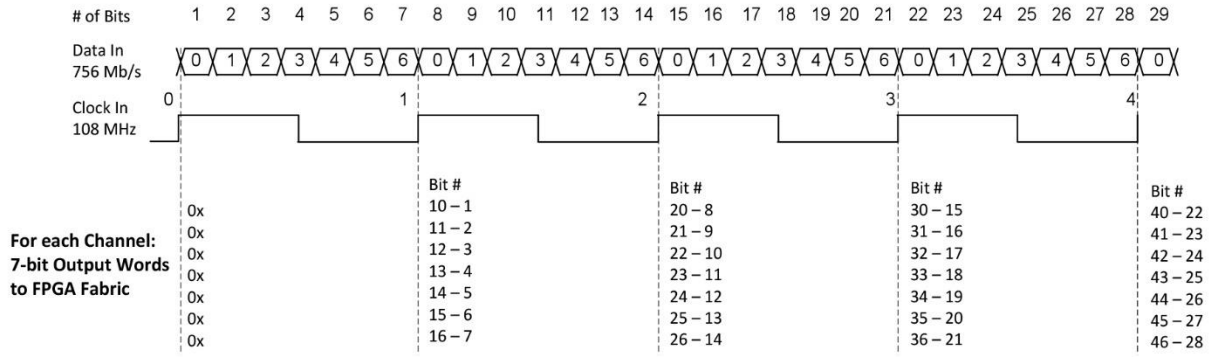


Figure 3.9. Transmit TX.CLK. Aligned Waveforms

Receiver – Shown for one LVDS Channel



Transmitter – Shown for one LVDS Channel

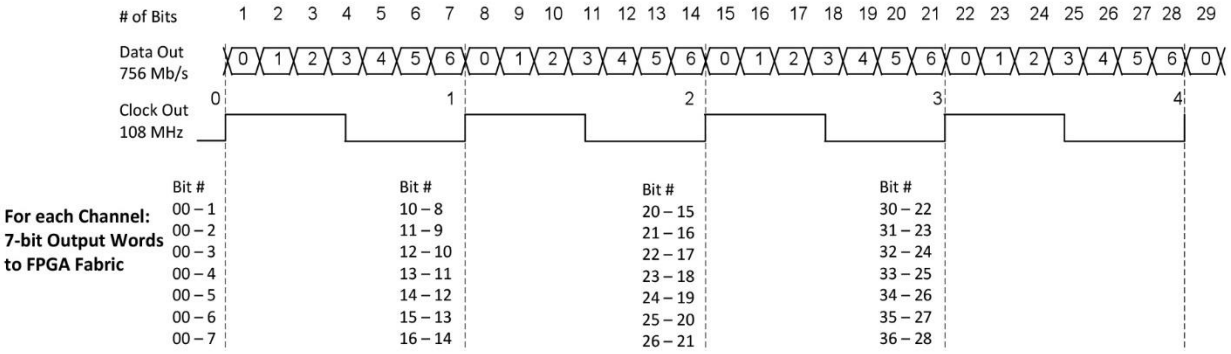


Figure 3.10. DDRX71 Video Timing Waveforms

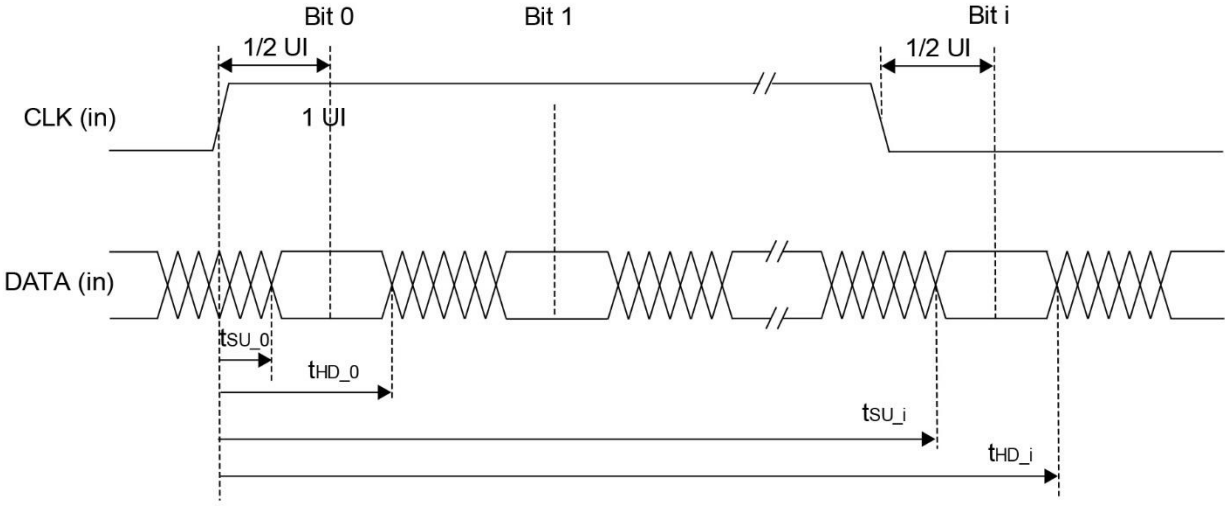


Figure 3.11. Receiver DDRX71_RX Waveforms

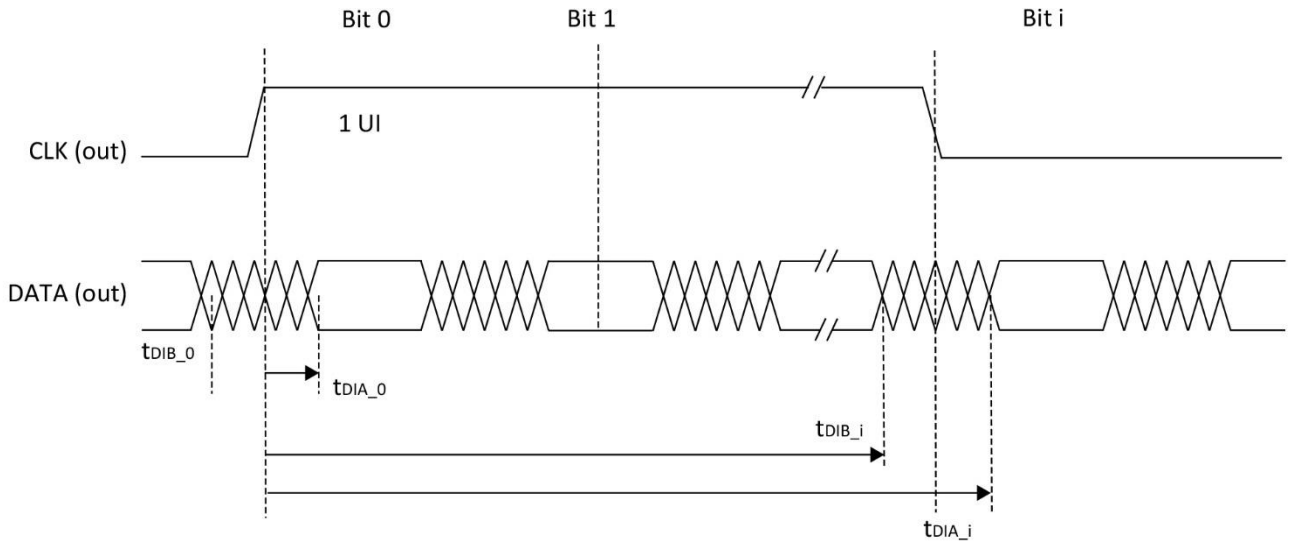


Figure 3.12. Transmitter DDRX71_TX Waveforms

Certus™-NX-RT sysCLOCK PLL Timing (V_{CC} = 1.0 V)

Over recommended operating conditions.

Table 3.29. sysCLOCK PLL Timing (V_{CC} = 1.0V)

Parameter	Description	Conditions	Min	Typ	Max	Unit
f_{IN}	Input Clock Frequency (CLKI, CLKFB)	—	10	—	500	MHz
f_{OUT}	Output Clock Frequency	—	6.25	—	800	MHz
f_{VCO}	PLL VCO Frequency	—	800	—	1600	MHz
f_{PFD}[3]	Phase Detector Input Frequency	Without Fractional-N Enabled	10	—	500	MHz
		With Fractional-N Enabled	10	—	100	MHz
AC Characteristics						
t_{DT}	Output Clock Duty Cycle	—	45	—	55	%
t_{PH4}	Output Phase Accuracy	—	–5	—	5	%
t_{OPJIT}[1]	Output Clock Period Jitter	f_{OUT} ≥ 200 MHz	—	—	250	ps p-p
		f_{OUT} < 200 MHz	—	—	0.05	UIPP
	Output Clock Cycle-to-Cycle Jitter	f_{OUT} ≥ 200 MHz	—	—	250	ps p-p
		f_{OUT} < 200 MHz	—	—	0.05	UIPP
	Output Clock Phase Jitter	f_{PFD} ≥ 200 MHz	—	—	250	ps p-p
		f_{PFD} < 200 MHz	—	—	0.05	UIPP
	Output Clock Period Jitter (Fractional-N)	f_{OUT} ≥ 200 MHz	—	—	350	ps p-p
		f_{OUT} < 200 MHz	—	—	0.07	UIPP
	Output Clock Cycle-to-Cycle Jitter (Fractional-N)	f_{OUT} ≥ 200 MHz	—	—	400	ps p-p
		f_{OUT} < 200 MHz	—	—	0.08	UIPP
f_{BW}[4]	PLL Loop Bandwidth	—	0.45		13	MHz

Parameter	Description	Conditions	Min	Typ	Max	Unit
t_{LOCK}[2]	PLL Lock-in Time	—	—	—	10	ms
t_{UNLOCK}	PLL Unlock Time (from RESET goes HIGH)	—	—	—	50	ns
t_{IPJIT}	Input Clock Period Jitter	f_{PFD} ≥ 20 MHz	—	—	500	ps p-p
		f_{PFD} < 20 MHz	—	—	0.01	UIPP
t_{HI}	Input Clock High Time	90% to 90%	0.5	—	—	ns
t_{LO}	Input Clock Low Time	10% to 10%	0.5	—	—	ns
t_{RST}	RST/ Pulse Width	—	1	—	—	ms
f_{SSC_MOD}	Spread Spectrum Clock Modulation Frequency	—	20	—	200	kHz
$f_{SSC\_MOD\_AMP}$	Spread Spectrum Clock Modulation Amplitude Range	—	0.25	—	2.00	%
$f_{SSC\_MOD\_STEP}$	Spread Spectrum Clock Modulation Amplitude Step Size	—	—	0.25	—	%

Notes:

1. Jitter sample is taken over 10,000 samples for Period jitter, and 1,000 samples for Cycle-to-Cycle jitter of the primary PLL output with clean reference clock with no additional I/O toggling.
2. Output clock is valid after t_{LOCK} for PLL reset and dynamic delay adjustment.
3. Period jitter and cycle-to-cycle jitter numbers are guaranteed for f_{PFD} > 10 MHz. For f_{PFD} < 10 MHz, the jitter numbers may not be met in certain conditions.
4. Result from Lattice Radiant software.

Certus™-NX-RT Internal Oscillators Characteristics

Table 3.30. Internal Oscillators (V_{CC} = 1.0V)

Symbol	Parameter Description	Min	Typ	Max	Unit
f_{CLKHF}	HFOSC CLKK Clock Frequency	418.5	450	481.5	MHz
f_{CLKLF}	LFOSC CLKK Clock Frequency	18.2	32	45.8	kHz
DCH_{CLKHF}	HFOSC Duty Cycle (Clock High Period)	43	50	57	%
DCH_{CLKLF}	LFOSC Duty Cycle (Clock High Period)	45	50	55	%

Certus™-NX-RT User I²C Characteristics

Table 3.31. User I²C Specifications (V_{CC} = 1.0 V)

Symbol	Parameter Description	STD Mode			Fast Mode			Fast Mode Plus[2]			Units
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
f_{SCL}	SCL Clock Frequency	—	—	100	—	—	400	—	—	1000	kHz
T_{DELAY}[1]	Optional delay through delay block	—	—	62	—	—	62	—	—	62	ns

Notes:

1. Refer to the I²C Specification for timing requirements. User design should set constraints in Lattice Design Software to meet this industrial I²C Specification.
2. Fast Mode Plus maximum speed may be achieved by using external pull up resistor on I²C bus. Internal pull up may not be sufficient to support the maximum speed.

Certus™-NX-RT Analog-Digital Converter (ADC) Block Characteristics

Table 3.32. ADC Specifications

Symbol	Description	Condition	Min	Typ	Max	Unit
V_{REFINT_ADC}	ADC Internal Reference Voltage	—	—	1.2	—	V
V_{REFEXT_ADC}	ADC External Reference Voltage	—	1.0	—	1.8	V
N_{RES_ADC}	ADC Resolution	—	—	12	—	bits
$ENOB_{ADC}$	Effective Number of Bits	—	9.9	10.8	—	bits
V_{SR_ADC}	ADC Input Range	Bipolar Mode, Internal V_{REF}	V_{CM_ADC} — $V_{REFINT\_ADC/4}$	V_{CM_ADC}	V_{CM_ADC} + $V_{REFINT\_ADC/4}$	V
		Bipolar Mode, External V_{REF}	V_{CM_ADC} — $V_{REFEXT\_ADC/4}$	V_{REFEXT_ADC}	V_{CM_ADC} + $V_{REFEXT\_ADC/4}$	V
		Uni-polar Mode, Internal V_{REF}	0	—	V_{REFINT_ADC}	V
		Uni-polar Mode, External V_{REF}	0	—	V_{REFEXT_ADC}	V
V_{CM_ADC}	ADC Input Common Mode Voltage (for fully differential signals)	Internal V_{REF}	—	$V_{REFINT\_ADC/2}$	—	V
		External V_{REF}	—	$V_{REFEXT\_ADC/2}$	—	V
f_{CLK_ADC}	ADC Clock Frequency	—	—	25	40	MHz
DC_{CLK_ADC}	ADC Clock Duty Cycle	—	48	50	52	%
f_{INPUT_ADC}	ADC Input Frequency	—	—	—	500	kHz
FS_{ADC}	ADC Sampling Rate	—	—	1	—	MS/s
N_{TRACK_ADC}	ADC Input Tracking Time	—	2	—	—	cycles
R_{IN_ADC}	ADC Input Equivalent Resistance	1 MS/s, Sampled @ 2 clock cycles	—	116	—	KΩ
t_{CAL_ADC}	ADC Calibration Time	—	—	—	6500	cycles
L_{OUTput_ADC}	ADC Conversion Time	—	25	—	—	cycles

Symbol	Description	Condition	Min	Typ	Max	Unit
DNL_{ADC}	ADC Differential Nonlinearity	—	–1	—	1	LSB
INL_{ADC}	ADC Integral Nonlinearity	—	–2.18	—	2.29	LSB
$SFDR_{ADC}$	ADC Spurious Free Dynamic Range	—	65.9	77	—	dBc
THD_{ADC}	ADC Total Harmonic Distortion	—	—	–76	–66.4	dB
SNR_{ADC}	ADC Signal to Noise Ratio	—	61.66	67.5	—	dB
$SNDR_{ADC}$	ADC Signal to Noise Plus Distortion Ratio	—	61.55	67	—	dB
ERR_{GAIN_ADC}	ADC Gain Error	—	–0.5	—	0.5	% FSADC
ERR_{OFFSET_ADC}	ADC Offset Error	—	–4	—	4	LSB
C_{IN_ADC}	ADC Input Equivalent Capacitance	—	—	2	—	pF

Certus™-NX-RT Comparator Block Characteristics

Table 3.33. Comparator Specifications

Symbol	Description	Min	Typ	Max	Unit
f_{IN_COMP}	Comparator Input Frequency	—	—	10	MHz
V_{IN_COMP}	Comparator Input Voltage	0	—	V_{CC_ADC18}	V
V_{OFFSET_COMP}	Comparator Input Offset	–34.3	—	36.44	mV
V_{HYST_COMP}	Comparator Input Hysteresis	10	—	31.62	mV
$V_{LATENCY\_COMP}$	Comparator Latency	—	—	31.24	ns

Certus™-NX-RT Digital Temperature Readout Characteristics

Digital Temperature Readout (DTR) is implemented in one of the channels of the ADC1.

Table 3.34. Digital Temperature Readout Specifications

Symbol	Description	Condition	Min	Typ	Max	Unit
DTR_{RANGE}	DTR Detect Temperature Range	—	–40	—	125	°C
$DTR_{ACCURACY}$	DTR Accuracy	with external voltage reference range of 1.0 V to 1.8 V	–16	±6	16	°C
$DTR_{RESOLUTION}$	DTR Resolution	with external voltage reference	–0.3	—	0.3	°C

Certus™-NX-RT Hardened PCIe Characteristics

PCIe (2.5 Gb/s)

Over recommended operating conditions

Table 3.35. PCIe (2.5 Gb/s)

Symbol	Description	Condition	Min	Typ	Max	Unit
Transmitter[1]						
UI	Unit Interval	—	399.88	400	400.12	ps
BW_{TX}	Tx PLL bandwidth	—	1.5	—	22	MHz
$V_{TX-DIFF-PP}$	Differential p-p Tx voltage swing	—	0.8	—	1.2	Vp-p
$V_{TX-DIFF-PP-LOW}$	Low power differential p-p Tx voltage swing	—	0.4	—	1.2	Vp-p
$V_{TX-DE-RATIO-3.5dB}$	Tx de-emphasis level ratio at 3.5dB	—	3	—	4	dB
$T_{TX-RISE-FALL}$	Transmitter rise and fall time	—	0.125	—	—	UI
T_{TX-EYE}	Transmitter Eye, including all jitter sources	—	0.75	—	—	UI
$T_{TX-EYE-MEDIAN-to-MAX-JITTER}$	Max. time between jitter median and max deviation from the median	—	—	—	0.125	UI
$RL_{TX-DIFF}$	Tx Differential Return Loss, including pkg and silicon	—	10	—	—	dB
RL_{TX-CM}	Tx Common Mode Return Loss, including pkg and silicon	50 MHz < freq < 2.5 GHz	6	—	—	dB
$Z_{TX-DIFF-DC}$	DC differential Impedance	—	80	—	120	Ω
$V_{TX-CM-AC-P}$	Tx AC peak common mode voltage, RMS	—	—	—	20	mV, RMS
$I_{TX-SHORT}$	Transmitter short-circuit current	—	—	—	90	mA
$V_{TX-DC-CM}$	Transmitter DC common-mode voltage	—	0	—	1.2	V
$V_{TX-IDLE-DIFF-AC-p}$	Electrical Idle Output peak voltage	—	—	—	20	mV
$V_{TX-RCV-DETECT}$	Voltage change allowed during Receiver Detect	—	—	—	600	mV
$T_{TX-IDLE-MIN}$	Min. time in Electrical Idle	—	20	—	—	ns
$T_{TX-IDLE-SET-TO-IDLE}$	Max. time from EI Order Set to valid Electrical Idle	—	—	—	8	ns
$T_{TX-IDLE-TO-DIFF-DATA}$	Max. time from Electrical Idle to valid differential output	—	—	—	8	ns
$L_{TX-SKEW}$	Lane-to-Lane output skew		—	—	500 ps + 2 UI	ps
Receiver[2]						
UI	Unit Interval	—	399.9	400	400.12	ps
$V_{RX-DIFF-PP}$	Differential Rx peak-peak voltage	—	0.175	—	1.2	Vp-p
T_{RX-EYE}[3]	Receiver eye opening time	—	0.4	—	—	UI
$T_{RX-EYE-MEDIAN-to-MAX-JITTER}$[3]	Max time delta between median and deviation from median	—	—	—	0.3	UI
$RL_{RX-DIFF}$	Receiver differential Return Loss, package plus silicon	—	10	—	—	dB

Symbol	Description	Condition	Min	Typ	Max	Unit
RL_{RX-CM}	Receiver common mode Return Loss, package plus silicon	—	6	—	—	dB
Z_{RX-DC}	Receiver DC single ended impedance	—	40	—	60	Ω
$Z_{RX-DIFF-DC}$	Receiver DC differential impedance	—	80	—	120	Ω
$Z_{RX-HIGH-IMP-DC}$	Receiver DC single ended impedance when powered down	—	200	—	—	kΩ
$V_{RX-CM-AC-P}$[3]	Rx AC peak common mode voltage	—		—	150	mV, peak
$V_{RX-IDLE-DET-DIFF-PP}$	Electrical Idle Detect Threshold	—	65	—	175	mVp-p
$L_{RX-SKEW}$	Receiver –lane-lane skew			—	20	ps

Notes:

1. Refer to PCI Express Base Specification Revision 3.0 Table 4.18 test condition and requirement for respective parameters.
2. Refer to PCI Express Base Specification Revision 3.0 Table 4.24 test condition and requirement for respective parameters.
3. Spec compliant requirement

PCIe (5 Gb/s)

Over recommended operating conditions

Table 3.36. PCIe (5 Gb/s)

Symbol	Description	Condition	Min	Typ	Max	Unit
Transmitter[1]						
UI	Unit Interval	—	199.94	200	200.06	ps
$B_{WTX-PKG-PLL1}$	Tx PLL bandwidth corresponding to $PKG_{TX-PLL1}$	—	8	—	16	MHz
$B_{WTX-PKG-PLL2}$	Tx PLL bandwidth corresponding to $PKG_{TX-PLL2}$	—	5	—	16	MHz
$P_{KGTX-PLL1}$	Tx PLL Peaking corresponding to $PK_{GTX-PLL1}$	—	—	—	3	dB
$P_{KGTX-PLL2}$	Tx PLL Peaking corresponding to $PKG_{TX-PLL2}$	—	—	—	1	dB
$V_{TX-DIFF-PP}$	Differential p-p Tx voltage swing	—	0.8	—	1.2	V, p-p
$V_{TX-DIFF-PP-LOW}$	Low power differential p-p Tx voltage swing	—	0.4	—	1.2	V, p-p
$V_{TX-DE-RATIO-3.5dB}$	Tx de-emphasis level ratio at 3.5dB	—	3	—	4	dB
$V_{TX-DE-RATIO-6dB}$	Tx de-emphasis level ratio at 6dB	—	5.5	—	6.5	dB
$T_{MIN-PULSE}$	Instantaneous lone pulse width	—	0.9	—	—	UI
$T_{TX-RISE-FALL}$	Transmitter rise and fall time	—	0.15	—	—	UI
T_{TX-EYE}	Transmitter Eye, including all jitter sources	—	0.75	—	—	UI
T_{TX-DJ}	Tx deterministic jitter > 1.5 MHz	—	—	—	0.15	UI
T_{TX-RJ}	Tx RMS jitter < 1.5 MHz	—	—	—	3	ps, RMS
$T_{RF-MISMATCH}$	Tx rise/fall time mismatch	—	—	—	0.1	UI
$R_{LTX-DIFF}$	Tx Differential Return Loss, including package and silicon	50 MHz < freq < 1.25 GHz	10	—	—	dB
$T_{TX-IDLE-MIN}$	Min. time in Electrical Idle	1.25 GHz < freq < 2.5 GHz	8	—	—	dB
R_{LTX-CM}	Tx Common Mode Return Loss, including package and silicon	50 MHz < freq < 2.5 GHz	6	—	—	dB

Symbol	Description	Condition	Min	Typ	Max	Unit
$Z_{TX-DIFF-DC}$	DC differential Impedance	—	—	—	120	Ω
$V_{TX-CM-AC-PP}$	Tx AC peak common mode voltage, peak-peak	—	—	—	150	mV, p-p
$I_{TX-SHORT}$	Transmitter short-circuit current	—	—	—	90	mA
$V_{TX-DC-CM}$	Transmitter DC common-mode voltage	—	0	—	1.2	V
$V_{TX-IDLE-DIFF-DC}$	Electrical Idle Output DC voltage	—	0	—	5	mV
$V_{TX-IDLE-DIFF-AC-p}$	Electrical Idle Differential Output peak voltage	—	—	—	20	mV
$V_{TX-RCV-DETECT}$	Voltage change allowed during Receiver Detect	—	—	—	600	mV
$T_{TX-IDLE-MIN}$	Min. time in Electrical Idle	—	20	—	—	ns
$T_{TX-IDLE-SET-TO-IDLE}$	Max. time from EI Order Set to valid Electrical Idle	—	—	—	8	ns
$T_{TX-IDLE-TO-DIFF-DATA}$	Max. time from Electrical Idle to valid differential output	—	—	—	8	ns
$L_{TX-SKEW}$	Lane-to-Lane output skew	—	—	—	500 + 4 UI	ps
Receiver[2]						
UI	Unit Interval	—	199.94	200	200.06	ps
$V_{RX-DIFF-PP}$	Differential Rx peak-peak voltage	—	0.343	—	1.2	V, p-p
$T_{RX-RJ-RMS}$	Receiver random jitter tolerance (RMS)	1.5 MHz – 100 MHz Random noise	—	—	4.2	ps, RMS
T_{RX-DJ}	Receiver deterministic jitter tolerance	—	—	—	88	ps
$R_{LRX-DIFF}$	Receiver differential Return Loss, package plus silicon	50 MHz < freq < 1.25 GHz	10	—	—	dB
R_{LRX-CM}	Receiver common mode Return Loss, package plus silicon	1.25 GHz < freq < 2.5 GHz	8	—	—	dB
R_{LRX-CM}	Receiver common mode Return Loss, package plus silicon	—	6	—	—	dB
Z_{RX-DC}	Receiver DC single ended impedance	—	40	—	60	Ω
$Z_{RX-HIGH-IMP-DC}$	Receiver DC single ended impedance when powered down	—	200	—	—	kΩ
$V_{RX-CM-AC-P}$[3]	Rx AC peak common mode voltage	—	—	—	150	mV, peak
$V_{RX-IDLE-DET-DIFF-PP}$	Electrical Idle Detect Threshold	—	65	—	1753	mv, pp
$L_{RX-SKEW}$	Receiver –lane-lane skew	—	—	—	8	ns

Notes:

1. Refer to PCI Express Base Specification Revision 3.0 Table 4.18 test condition and requirement for respective parameters.
2. Refer to PCI Express Base Specification Revision 3.0 Table 4.24 test condition and requirement for respective parameters.
3. Spec compliant requirement

Certus™-NX-RT Hardened SGMII Receiver Characteristics

SGMII Rx Specifications

Over recommended operating conditions

Table 3.37. SGMII Rx

Symbol	Description	Condition	Min	Typ	Max	Unit
f_{DATA}	SGMII Data Rate	—	—	1.25	—	Gb/s
f_{REFCLK}	SGMII Reference Clock Frequency (Data Rate / 10)	—	—	125	—	MHz
J_{TOL_Rj}	Jitter Tolerance, Random (RMS)	—	—	—	15	mUIrms
J_{TOL_Dj}	Jitter Tolerance, Deterministic	Periodic jitter 1 kHz–22 MHz		—	0.05*	UI
J_{TOL_Tj}	Jitter Tolerance, Total	Periodic jitter 1 kHz–22 MHz		—	0.26*	UI
Δf/f	Data Rate and Reference Clock Accuracy	—	–300	—	300	ppm

***Note:** J_{TOT} can meet the following deterministic jitter mask specification: 0 to 3.5 kHz: 10 UI; 3.5 to 700 kHz: log-log slope 10 UI to 0.05 UI; above 700 kHz: 0.05 UI.

Certus™-NX-RT sysCONFIG Port Timing Characteristics

Over recommended operating conditions

Table 3.38. Certus NX sysCONFIG Port Timing Specifications

Symbol	Description	Device	Min	Typ	Max	Unit
Master SPI POR / REFRESH Timing						
t_{ICFG}	Time during POR, from V_{CC}, V_{CCAUX}, V_{CCIO0} or V_{CCIO1} (whichever is the last) pass POR trip voltage, or REFRESH command executed, to the last rising edge of INITN	—	—	—	5	ms
t_{VMC}	Time from last rising edge of INITN to the valid Master MCLK	—	—	—	5	µs
f_{MCLK_DEF}	Default MCLK frequency (Before MCLK frequency selection in bitstream)	—	—	3.5	—	MHz
Slave SPI/I²C/I3C POR / REFRESH Timing						
t_{MSPI_INH}	Time during POR, from V_{CC}, V_{CCAUX}, V_{CCIO0} or V_{CCIO1} (whichever is the last) pass POR trip voltage, or REFRESH command executed, to pull PROGRAMN LOW to prevent entering MSPI mode	—	—	—	1	µs
$t_{ACT\_PROGRAMN\_H}$	Minimum time driving PROGRAMN HIGH after last activation clock	—	50	—	—	ns
t_{CONFIG_CCLK}	Minimum time to start driving CCLK (SSPI) after PROGRAMN HIGH	—	50	—	—	ns
t_{CONFIG_SCL}	Minimum time to start driving SCL (I²C/I3C) after PROGRAMN HIGH	—	50	—	—	ns
PROGRAMN Configuration Timing						
$t_{PROGRAMN}$	PROGRAMN LOW pulse accepted	—	50	—	—	ns
$t_{PROGRAMN\_RJ}$	PROGRAMN LOW pulse rejected	—	—	—	25	ns
t_{INIT_LOW}	PROGRAMN LOW to INITN LOW	—	—	—	100	ns
t_{INIT_HIGH}	PROGRAMN LOW to INITN HIGH	LIFCL-40	—	30	—	µs
		LIFCL-17	—	30	—	µs
t_{DONE_LOW}	PROGRAMN LOW to DONE LOW	—	—	—	55	µs
t_{DONE_HIGH}	PROGRAMN HIGH to DONE HIGH	—		—	2	s
t_{IODISS}	PROGRAMN LOW to I/O Disabled	—	—	—	125	ns
Master SPI						
f_{MCLK}*	Max selected MCLK output frequency	—	—	112.5	124	MHz
f_{MCLK_DC}	MCLK output clock duty cycle	—	40	—	60	%
t_{MCLKH}	MCLK output clock pulse width HIGH	—	3.5	—	—	ns
t_{MCLKL}	MCLK output clock pulse width LOW	—	3.5	—	—	ns
t_{SU_MSI}	MSI to MCLK setup time	—	3	—	—	ns
t_{HD_MSI}	MSI to MCLK hold time	—	0.5	—	—	ns
t_{CO_MSO}	MCLK to MSO delay	—	—	—	12	ns

Symbol	Description	Device	Min	Typ	Max	Unit
Slave SPI						
f_{CCLK}	CCLK input clock frequency	—	—	—	120	MHz
t_{CCLKH}	CCLK input clock pulse width HIGH	—	3.5	—	—	ns
t_{CCLKL}	CCLK input clock pulse width LOW	—	3.5	—	—	ns
t_{VMC_SLAVE}	Time from rising edge of INITN to Slave CCLK driven	—	50	—	—	ns
t_{VMC_MASTER}	CCLK input clock duty cycle	—	40	—	60	%
t_{SU_SSI}	SSI to CCLK setup time	—	3.2	—	—	ns
t_{HD_SSI}	SSI to CCLK hold time	—	1.9	—	—	ns
t_{CO_SSO}	CCLK falling edge to valid SSO output	—	—	—	30	ns
t_{EN_SSO}	CCLK falling edge to SSO output enabled	—	—	—	30	ns
t_{DIS_SSO}	CCLK falling edge to SSO output disabled	—	—	—	30	ns
t_{HIGH_SCSN}	SCSN HIGH time	—	74	—	—	ns
t_{SU_SCSN}	SCSN to CCLK setup time	—	3.5	—	—	ns
t_{HD_SCSN}	SCSN to CCLK hold time	—	1.6	—	—	ns
I²C/I3C						
f_{SCL_I2C}	SCL input clock frequency for I²C	—	—	—	1	MHz
f_{SCL_I3C}	SCL input clock frequency for I3C	—	—	—	12	MHz
t_{SCLH_I2C}	SCL input clock pulse width HIGH for I²C	—	400	—	—	ns
t_{SCLL_I2C}	SCL input clock pulse width LOW for I²C	—	400	—	—	ns
$t_{SU\_SDA\_I2C}$	SDA to SCL setup time for I²C	—	250	—	—	ns
$t_{HD\_SDA\_I2C}$	SDA to SCL hold time for I²C	—	50	—	—	ns
$t_{SU\_SDA\_I3C}$	SDA to SCL setup time for I3C	—	30	—	—	ns
$t_{HD\_SDA\_I3C}$	SDA to SCL hold time for I3C	—	30	—	—	ns
t_{CO_SDA}	SCL falling edge to valid SDA output	—	—	—	200	ns
t_{EN_SDA}	SCL falling edge to SDA output enabled	—	—	—	200	ns
t_{DIS_SDA}	SCL falling edge to SDA output disabled	—	—	—	200	ns
Wake-Up Timing						
t_{DONE_HIGH}	Last configuration clock cycle to DONE going HIGH	—	—	—	60	µs
t_{FIO_EN}	User I/O enabled in Fast I/O Mode	LIFCL-40	—		6.245	M cycles
		LIFCL-17	—		2.321	M cycles
t_{IOEN}	Config clock to user I/O enabled	—	150	—	—	ns
t_{MCLKZ}	Master MCLK to Hi-Z	—	—	—	2.5	µs

***Note:** fMCLK has a dependency on HFOSC and is 1/3 of fCLKHF.

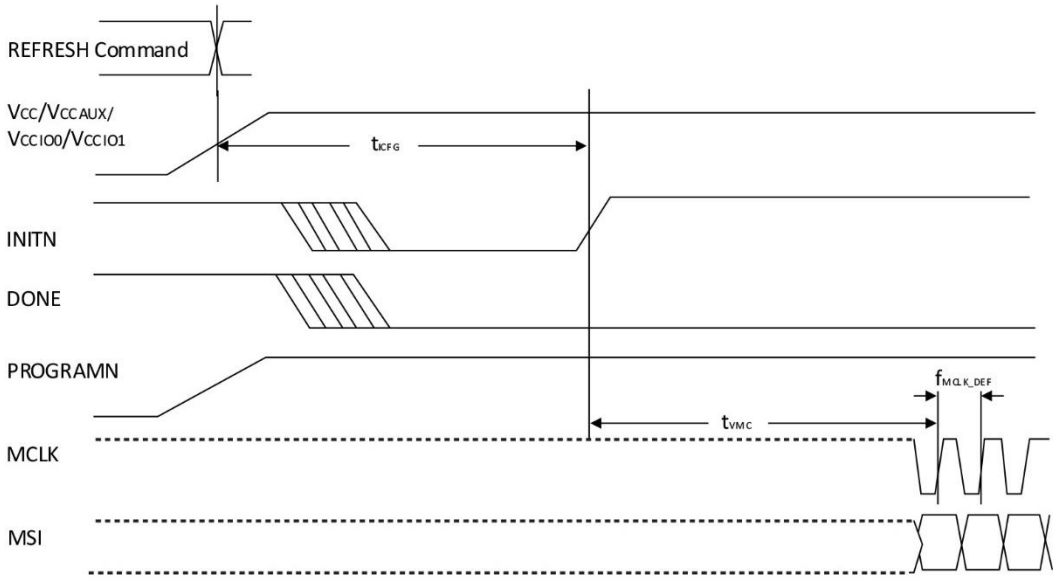


Figure 3.13. Master SPI POR/REFRESH Timing

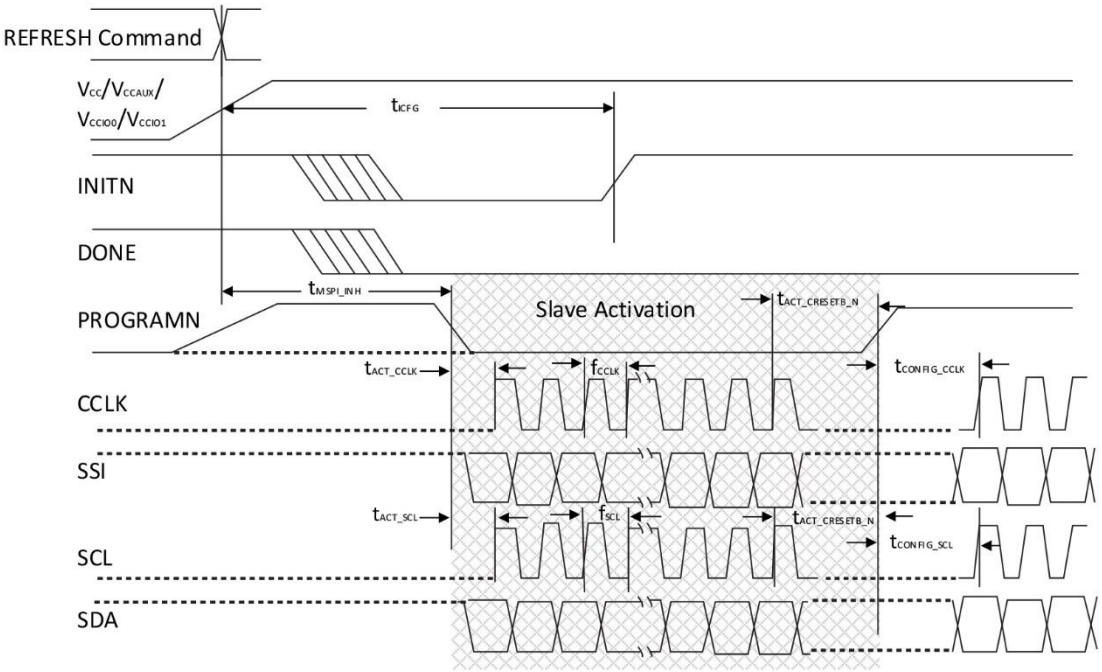


Figure 3.14. Slave SPI/I²C/I3C POR/REFRESH Timing

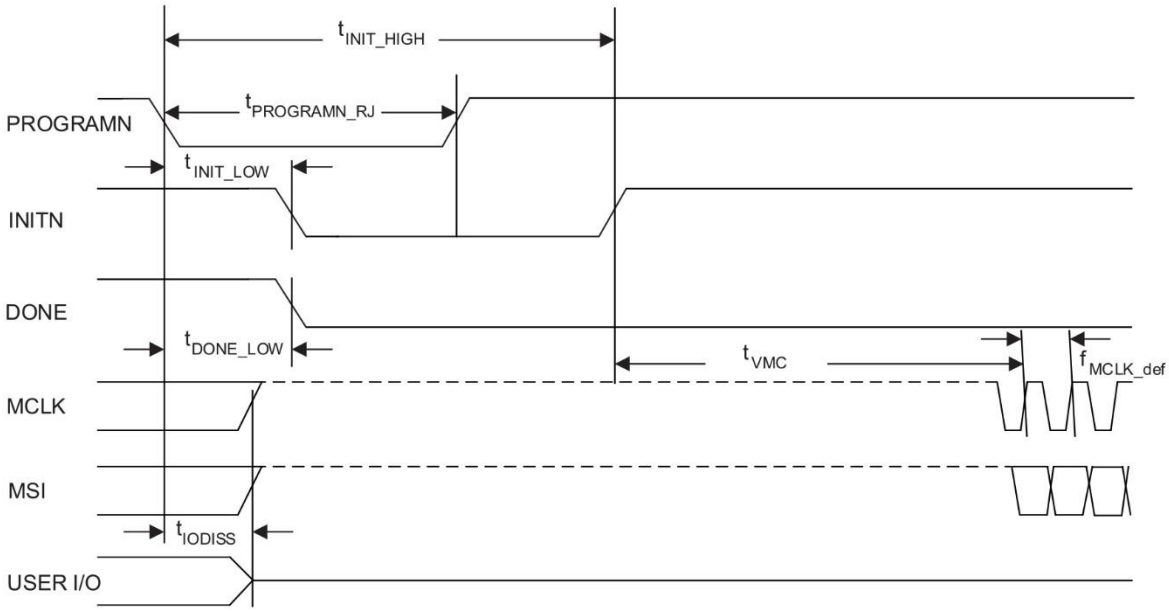


Figure 3.15. Master SPI PROGRAMNTiming

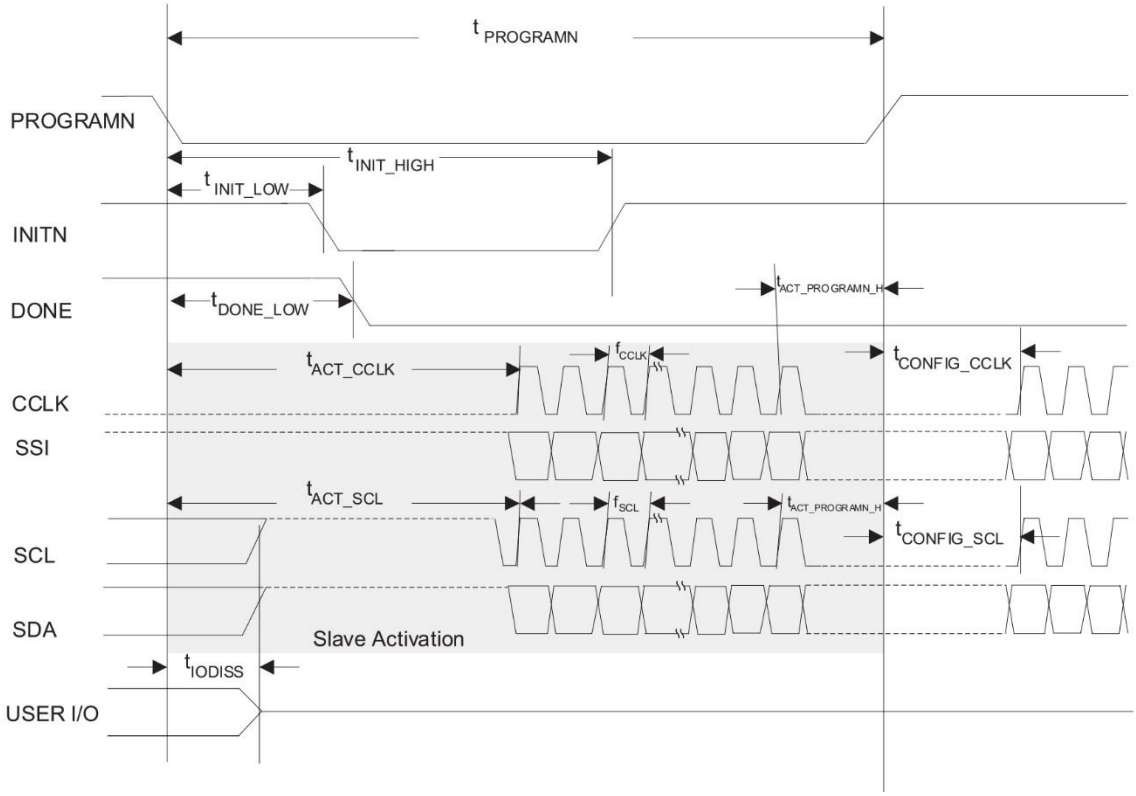


Figure 3.16. Slave SPI/I²C/I3C PROGRAMN Timing

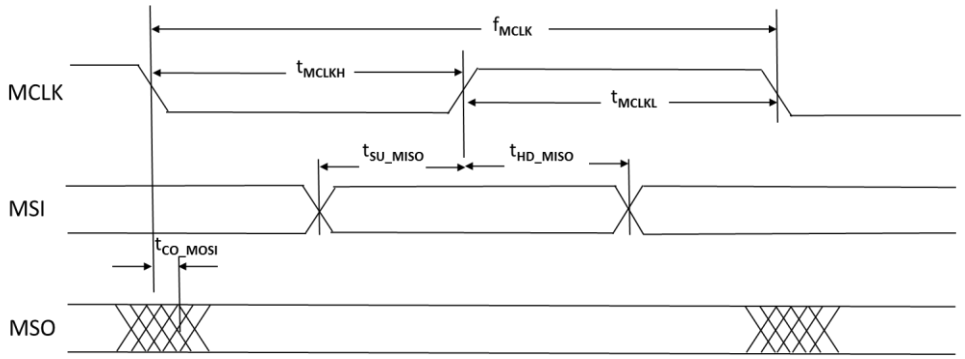


Figure 3.17. Master SPI Configuration Timing

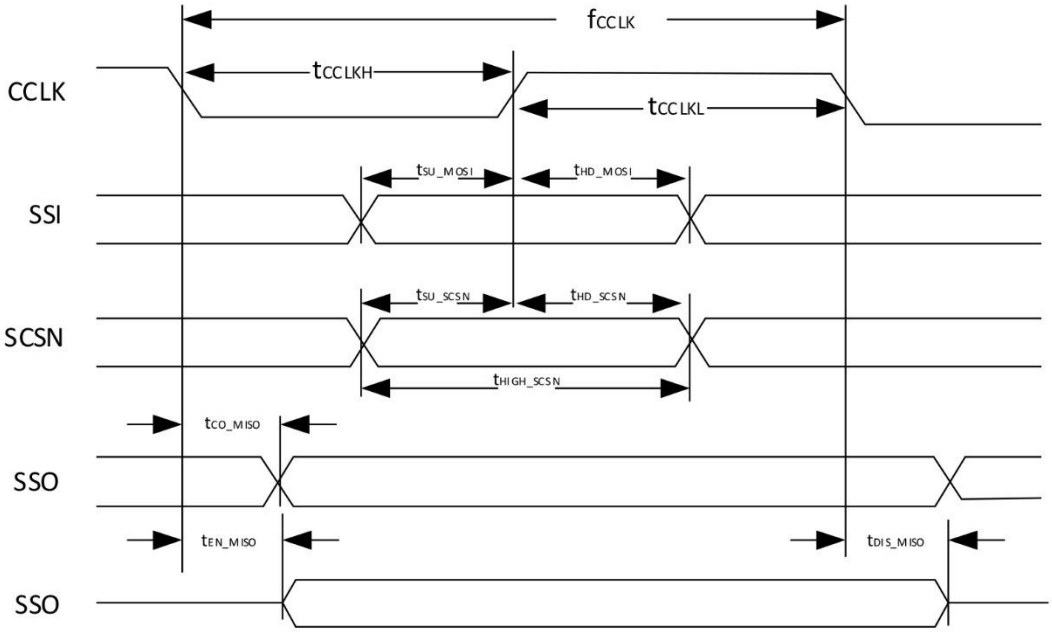


Figure 3.18. Slave SPI Configuration Timing

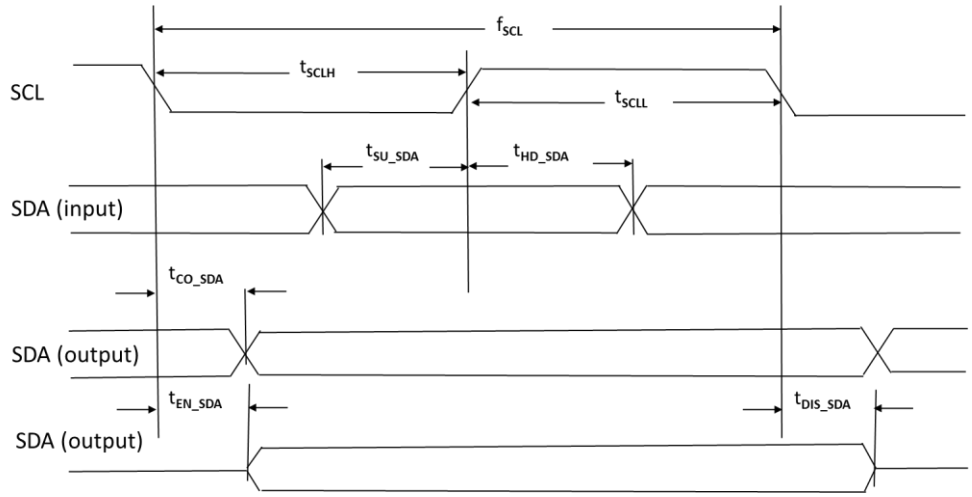


Figure 3.19. I²C/I3C Configuration Timing

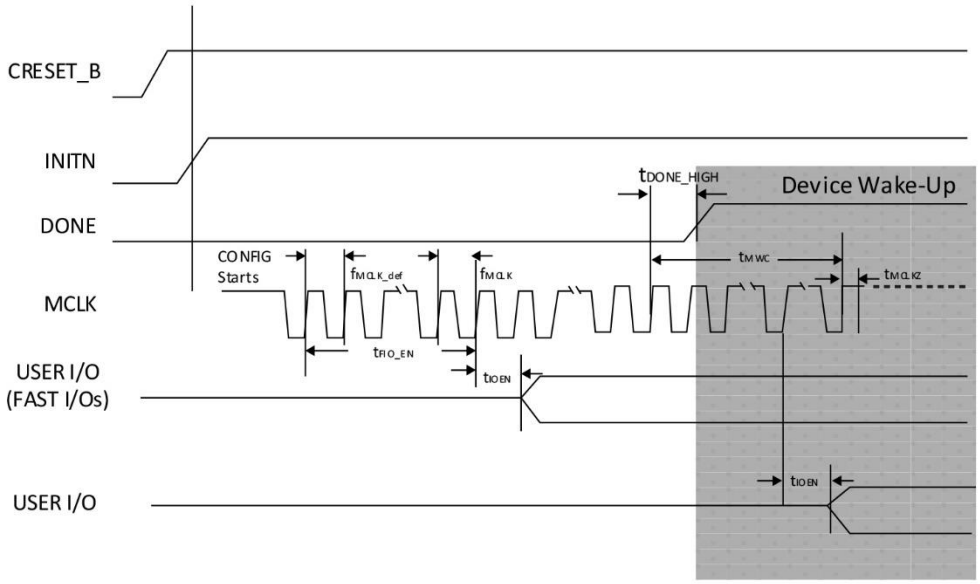


Figure 3.20. Master SPI Wake-Up Timing

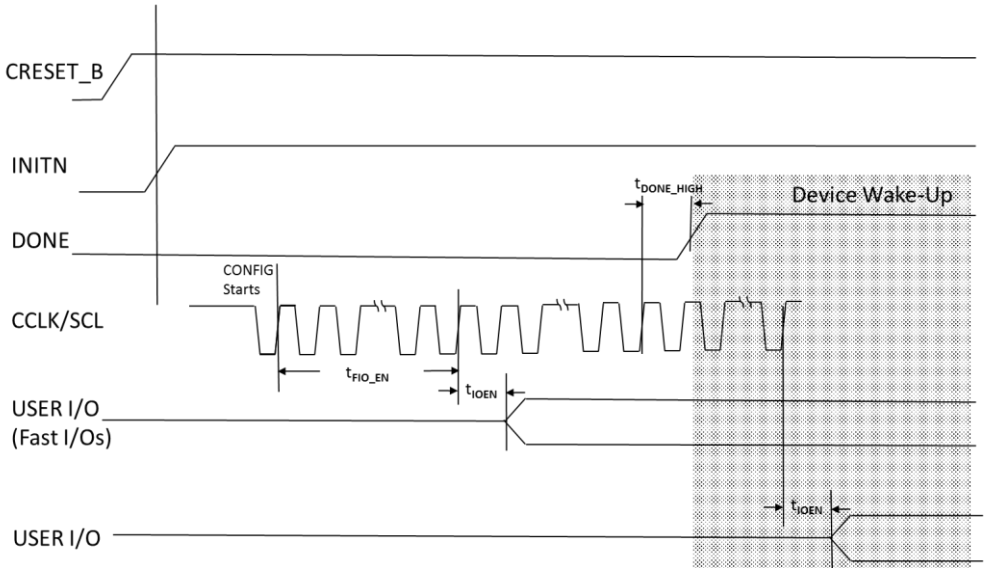


Figure 3.21. Slave SPI/I²C/I3C Wake-Up Timing

Certus™-NX-RT JTAG Port Timing Characteristics

Over recommended operating conditions.

Table 3.39. JTAG Port Timing Specifications

Symbol	Description	Min	Typ	Max	Unit
f_{MAX}	TCK clock frequency	—	—	25	MHz
t_{BTCPH}	TCK clock pulse width high	20	—	—	ns
t_{BTCPL}	TCK clock pulse width low	20	—	—	ns
t_{BTS}	TCK TAP setup time	5	—	—	ns
t_{BTH}	TCK TAP hold time	5	—	—	ns
t_{BTRF}	TAP controller TDO rise/fall time*	100	—	—	mV/ns
t_{BTCO}	TAP controller falling edge of clock to valid output	—	—	14	ns
$t_{BTCODIS}$	TAP controller falling edge of clock to valid disable	—	—	14	ns
t_{BTCOEN}	TAP controller falling edge of clock to valid enable	—	—	14	ns
t_{BTCRS}	BSCAN test capture register setup time	8	—	—	ns
t_{BTCRH}	BSCAN test capture register hold time	25	—	—	ns
t_{BUTCO}	BSCAN test update register, falling edge of clock to valid output	—	—	25	ns
$t_{BTUODIS}$	BSCAN test update register, falling edge of clock to valid disable	—	—	25	ns
$t_{BTUPOEN}$	BSCAN test update register, falling edge of clock to valid enable	—	—	25	ns

***Note:** Based on default IO setting of slow slew rate.

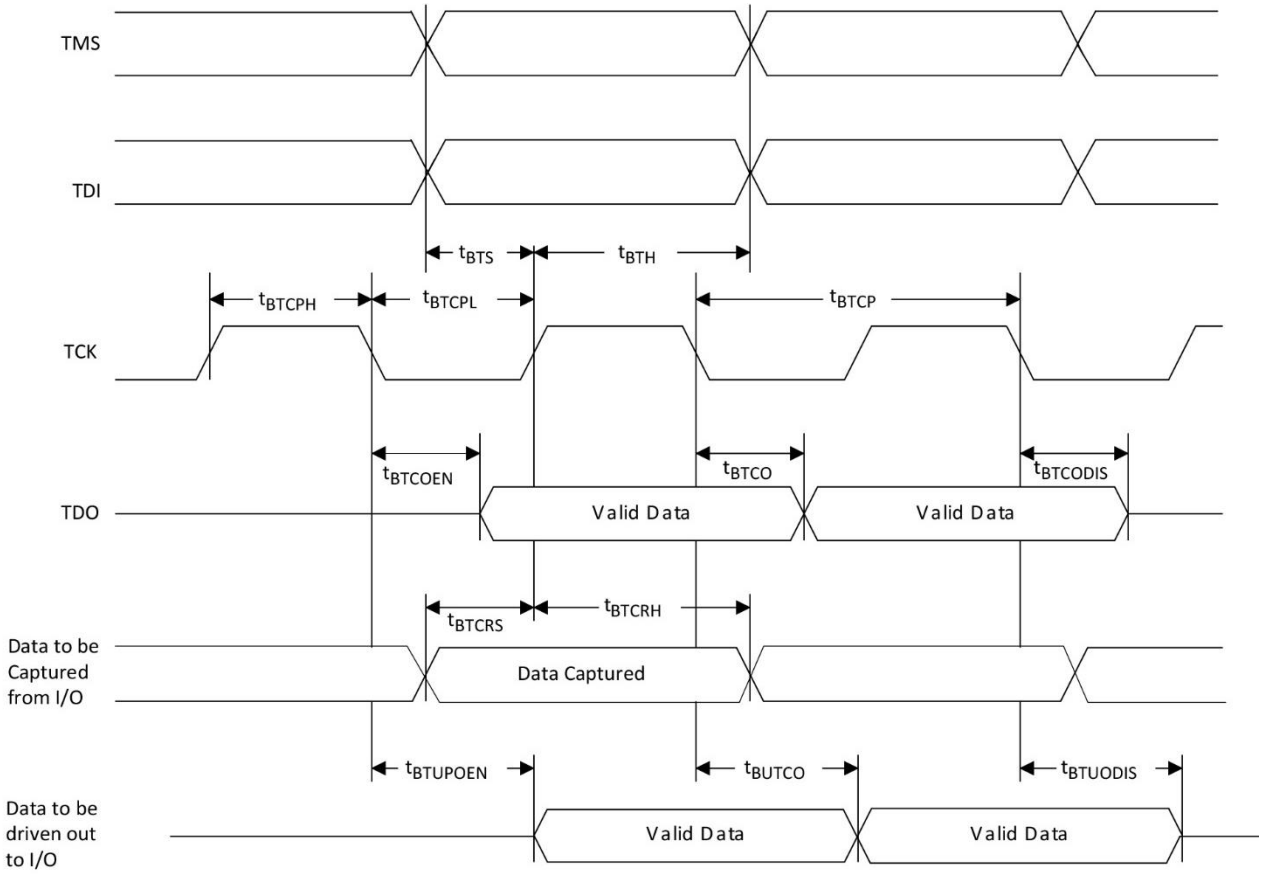
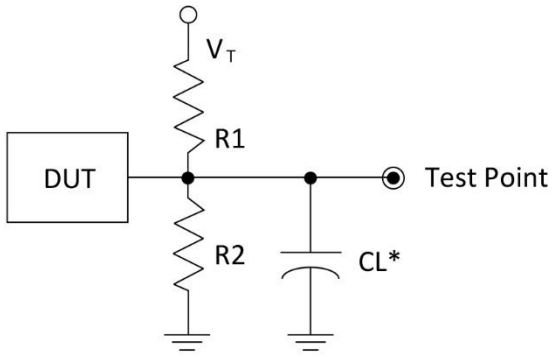


Figure 3.22. JTAG Port Timing Waveforms

Certus™-NX-RT Switching Test Characteristics

Figure 3.23 shows the output test load that is used for AC testing. The specific values for resistance, capacitance, voltage, and other test conditions are listed in Table 3.40.



*CL Includes Test Fixture and Probe Capacitance

Figure 3.23. Output Test Load, LVTTL and LVCMOS Standards

Table 3.40. Test Fixture Required Components, Non-Terminated Interfaces

Test Condition	R_1	R_2	C_L	Timing Ref.	V_T
LVTTL and other LVCMOS settings (L ≥ H, H ≥ L)	∞	∞	0 pF	LVCMOS 3.3 = 1.5 V	—
				LVCMOS 2.5 = $V_{CCIO}/2$	—
				LVCMOS 1.8 = $V_{CCIO}/2$	—
				LVCMOS 1.5 = $V_{CCIO}/2$	—
				LVCMOS 1.2 = $V_{CCIO}/2$	—
LVCMOS 2.5 I/O (Z ≥ H)	∞	1 MΩ	0 pF	$V_{CCIO}/2$	—
LVCMOS 2.5 I/O (Z ≥ L)	1 MΩ	∞	0 pF	$V_{CCIO}/2$	V_{CCIO}
LVCMOS 2.5 I/O (H ≥ Z)	∞	100	0 pF	V_{OH} – 0.10	—
LVCMOS 2.5 I/O (L ≥ Z)	100	∞	0 pF	V_{OL} + 0.10	V_{CCIO}

Note: Output test conditions for all other interfaces are determined by the respective standards.

Pinout Information

Signal Descriptions*

Signal Name	Bank	Type	Description
Power and GND			
V_{SS}	—	GND	Ground for internal FPGA logic and I/O
V_{SSSD}	—	GND	Ground for the SerDes block
V_{SSADC}	—	GND	Ground for ADC block
V_{CC}, V_{CCECLK}	—	Power	Power supply pins for core logic. V_{CC} is connected to 1.0 V (nom.) supply voltage. Power On Reset (POR) monitors this supply voltage.
V_{CCAUXA}	—	Power	Auxiliary power supply pin for internal analog circuitry. This supply is connected to 1.8 V (nom.) supply voltage. POR monitors this supply voltage.
V_{CCAUX}	—	Power	Auxiliary power supply pin for I/O Bank 0, Bank 1, Bank 2, Bank 6, and Bank 7. This supply is connected to 1.8 V (nom.) supply voltage, and is used for generating stable drive current for the I/O.
$V_{CCAUXHx}$	3–5	Power	Auxiliary power supply pin for I/O Bank 3, Bank 4, and Bank 5. This supply is connected to 1.8 V (nom.) supply voltage, and is used for generating stable current for the differential input comparators and stable drive current for the I/O.
V_{CCIOx}	0–7	Power	Power supply pins for I/O bank x. For x = 0, 1, 2, 6, and 7, V_{CCIO} can be connected to (nom.) 1.2 V, 1.5 V, 1.8 V, 2.5 V, or 3.3 V. For x = 3, 4, and 5, V_{CCIO} can be connected to (nom.) 1.0 V, 1.2 V, 1.35 V, 1.5 V, or 1.8 V. There are dedicated and shared configuration pins in banks 0 and 1. POR monitors these banks supply voltages.
$V_{CCADC18}$	—	Power	1.8 V (nom.) power supply for the ADC block.
V_{CCSD0}	—	Power	1.0 V (nom.) power supply for the SerDes block.
$V_{CCPLLSD0}$	—	Power	1.8 V (nom.) power supply for the PLL in the SerDes block.
$V_{CCAUXSD}$	—	Power	1.8 V (nom.) auxiliary power supply for the SerDes block.
Dedicated Pins			
Dedicated Configuration I/O Pin			
JTAG_EN	1	Input	LVCMOS input pin. This input selects the JTAG shared GPIO to be used for JTAG 0 = GPIO 1 = JTAG
Dedicated ADC I/O Pins			
ADC_REFA, ADC_REFB	—	Input	ADC reference voltage, for each of the 2 ADC converters
ADC_DP/NA, ADC_DP/NB	—	Input	Dedicated ADC input pairs, for each of the 2 ADC converters
Dedicated SerDes I/O Pins			
SD0_RXDP/N	—	Input	SerDes Data Differential Input Pairs
SD0_TXDP/N	—	Output	SerDes Data Differential Output Pairs
SD0_REFCLKP/N	—	Input	SerDes Reference Clock Differential Input Pairs

Signal Name	Bank	Type	Description
SD0_REXT	—	Input	SerDes External Reference Resistor Input. Resistor connects between to this pin and SD0_REFRET pin. This is used to adjust the on-chip differential termination impedance, based on the external resistance value: R_{EXT} = 909 Ω, R_{DIFF} = 80 Ω R_{EXT} = 976 Ω, R_{DIFF} = 85 Ω R_{EXT} = 1.02 kΩ, R_{DIFF} = 90 Ω R_{EXT} = 1.15 kΩ, R_{DIFF} = 100 Ω
SD0_REFRET	—	Input	SerDes Reference Return Input. These pins should be AC coupled to the $V_{CCPLLSD0}$ supply.
Misc. Pins			
NC		—	No connect.
RESERVED		—	This pin is reserved and should not be connected to anything on the board.
General Purpose I/O Pins			
P[T/R/B/L] [Number]_[A/B]	T = 0 R = 1, 2 B = 3, 4, 5 L = 6, 7	Input, Output, Bi-Dir	Programmable User I/O: • [T/B/L/R] indicates the package pin/ball is in T (Top), B (Bottom), L (Left), or R (Right) edge of the device. • [Number] identifies the PIO [A/B] pair. • [A/B] shows the package pin/ball is A or B signal in the pair. PIO A and PIO B are grouped as a pair. • Each A/B pair in the bottom banks supports true differential input and output buffers. When configured as differential input, differential termination of 100 Ω can be selected. • Each A/B pair in the top, left and right banks does not support true differential input or output buffer. It supports all single-ended inputs and outputs, and can be used for emulated differential output buffer. • Some of these user-programmable I/O are used during configuration, depending on the configuration mode. You need to make appropriate connection on the board to isolate the 2 different functions before/after configuration. • Some of these user-programmable I/O are shared with special function pins. These pins, when not used as special purpose pins, can be programmed as I/O for user logic. • During configuration the user-programmable I/O are tri-stated with an internal weak pull-down resistor enabled. If any pin is not used (or not bonded to a package pin), it is tri-stated and default to have weak pull-down enabled after configuration.
Shared Configuration Pins[1, 2] **1. These pins can be used for configuration during configuration mode. When configuration is completed, these pins can be used as GPIO, or shared function in GPIO. When these pins are used in dual function, you need to isolate the signal paths for the dual functions on the board.** **2. The pins used are defined by the configuration modes detected. Slave SPI or I²C/I3C modes are detected during slave activation. Pins that are not used in the configuration mode selected are tri-stated during configuration and can connect directly as GPIO in user's function.**			
PRxxx/SDA/USER_SDA	1	Input, Output, Bi-Dir	Configuration: I²C/I3C Mode: SDA signal User Mode: PRxxx: GPIO User_SDA: SDA signal for I²C/I3C interface

Signal Name	Bank	Type	Description
PRxxx/SCL/USER_SCL	1	Input, Output, Bi-Dir	Configuration: I²C/I3C Mode: SCL signal User Mode: PRxxx: GPIO User_SDA: SCL signal for I²C/I3C interface
PRxxx/TDO/SSO	1	Input, Output, Bi-Dir	Configuration: Slave SPI Mode: Slave Serial Output User Mode: PRxxx: GPIO TDO: When JTAG_EN = 1, used as TDO signal for JTAG
PRxxx/TDI/SSI	1	Input, Output, Bi-Dir	Configuration: Slave SPI Mode: Slave Serial Input User Mode: PRxxx: GPIO TDI: When JTAG_EN = 1, used as TDI signal for JTAG
PRxxx/TMS/SCSN	1	Input, Output, Bi-Dir	Configuration: Slave SPI Mode: Slave Chip Select User Mode: PRxxx: GPIO TMS: When JTAG_EN = 1, used as TMS signal for JTAG
PRxxx/TCK/SCLK	1	Input, Output, Bi-Dir	Configuration: Slave SPI Mode: Slave Clock Input User Mode: PRxxx: GPIO TCK: When JTAG_EN = 1, used as TCK signal for JTAG
PTxxx/MCSNO	0	Input, Output, Bi-Dir	Configuration: Master SPI Mode: Chip Select Output User Mode: PTxxx: GPIO
PTxxx/MD3	0	Input, Output, Bi-Dir	Configuration: Master Quad SPI Mode: I/O3 User Mode: PTxxx: GPIO
PTxxx/MD2	0	Input, Output, Bi-Dir	Configuration: Master Quad SPI Mode: I/O2 User Mode: PTxxx: GPIO
PTxxx/MSI/MD1	0	Input, Output, Bi-Dir	Configuration: Master SPI Mode: Master Serial Input Master Quad SPI Mode: I/O1 User Mode: PTxxx: GPIO

Signal Name	Bank	Type	Description
PTxxx/MSO/MD0	0	Input, Output, Bi-Dir	Configuration: Master SPI Mode: Master Serial Output Master Quad SPI Mode: I/O0 User Mode: PTxxx: GPIO
PTxxx/MCSN/PCLKT0_1	0	Input, Output, Bi-Dir	Configuration: Master SPI Mode: Master Chip Select Output User Mode: PTxxx: GPIO PCLKT0_0: Top PCLK Input
PTxxx/MCLK/PCLKT0_0	0	Input, Output, Bi-Dir	Configuration: Master SPI Mode: Master Clock Output User Mode: PTxxx: GPIO PCLKT0_1: Top PCLK Input
PTxxx/PROGRAMN	0	Input, Output, Bi-Dir	Configuration: PROGRAMN: Initiate configuration sequence when asserted LOW. User Mode: PTxxx: GPIO
PTxxx/INITN	0	Input, Output, Bi-Dir	Configuration: INITN: Open Drain I/O pin. This signal is driven to LOW when configuration sequence is started, to indicate the device is in initialization state. This signal is released after initialization is completed, and the configuration download can start. You can keep drive this signal LOW to delay configuration download to start. User Mode: PTxxx: GPIO
PTxxx/DONE	0	Input, Output, Bi-Dir	Configuration: DONE: Open Drain I/O pin. This signal is driven to LOW during configuration time. It is released to indicate the device has completed configuration. You can keep drive this signal LOW to delay the device to wake up from configuration. User Mode: PTxxx: GPIO
Shared User GPIO Pins[1,2,3,4] **1. Shared User GPIO pins are pins that can be used as GPIO, or functional pins that connect directly to specific functional blocks, when device enters into User Mode.** **2. Declaring on assigning the pin as GPIO or specific functional pin is done by configuration bitstream, except JTAG pins.** **3. JTAG pins are controlled by JTAG_EN signal. When JTAG_EN = 1, the pins are used for JTAG interface. When JTAG = 0, the pins are used as GPIO or specific functional pin defined by configuration bitstream.** **4. Refer to package pin file.**			
Shared JTAG Pins			
PRxxx/TDO/ yyyy	1	Input, Output, Bi-Dir	User Mode: PRxxx: GPIO TDO: When JTAG_EN = 1, used as TDO signal for JTAG yyyy: Other possible selectable specific functional

Signal Name	Bank	Type	Description
PRxxx/TDI/yyyy	1	Input, Output, Bi-Dir	User Mode: PRxxx: GPIO TDI: When JTAG_EN = 1, used as TDI signal for JTAG yyyy: Other possible selectable specific functional
PRxxx/TMS/ yyyy	1	Input, Output, Bi-Dir	User Mode: PRxxx: GPIO TMS: When JTAG_EN = 1, used as TMS signal for JTAG yyyy: Other possible selectable specific functional
PRxxx/TCK/ yyyy	1	Input, Output, Bi-Dir	User Mode: PRxxx: GPIO TCK: When JTAG_EN = 1, used as TCK signal for JTAG Yyyy: Other possible selectable specific functional
Shared CLOCK Pins[1] **Some PCLK pins can also be used as GPLL reference clock input pin. Refer to Nexus sysCLOCK PLL Design and Usage Guide (FPGA-TN-02095).**			
PBxxx/PCLK[T,C][3,4,5]_[0-3]/yyyy	3, 4, 5	Input, Output, Bi-Dir	User Mode: PBxxx: GPIO PCLK: Primary Clock or GPLL Refclk signal [T,C] = True/Complement when using differential signaling [3,4,5] = Bank [0-3] Up to 4 signals in the bank yyyy: Other possible selectable specific functional
PTxxx/PCLKT0_[0-1]/yyyy	0	Input, Output, Bi-Dir	User Mode: PTxxx: GPIO PCLKT: Primary Clock or GPLL Refclk signal (Only Single Ended) [0-1] Up to 2 signals in the bank yyyy: Other possible selectable specific functional
PRxxx/PCLKT[1,2]_[0-2]/yyyy	1, 2	Input, Output, Bi-Dir	User Mode: PRxxx: GPIO PCLKT: Primary Clock or GPLL Refclk signal (Only Single Ended) [0-2] Up to 3 signals in the bank yyyy: Other possible selectable specific functional
PLxxx/PCLKT[6,7]_[0-2]/yyyy	6, 7	Input, Output, Bi-Dir	User Mode: PLxxx: GPIO PCLKT: Primary Clock or GPLL Refclk signal (Only Single Ended) [0-2] Up to 3 signals in the bank yyyy: Other possible selectable specific functional
PBxxx/LRC_GPLL[T,C]_IN/yyyy	3	Input, Output, Bi-Dir	User Mode: PBxxx: GPIO LRC_GPLL: Lower Right GPLL Refclk signal (PLLCK) [T,C] = True/Complement when using differential signaling yyyy: Other possible selectable specific functional

Signal Name	Bank	Type	Description
PBxxx/LLC_GPLL[T,C]_IN/yyyy	5	Input, Output, Bi-Dir	User Mode: PBxxx: GPIO LLC_GPLL: Lower Left GPLL Refclk signal (PLLCK) [T,C] = True/Complement when using differential signaling yyyy: Other possible selectable specific functional
PLxxx/ULC_GPLLT_IN/yyyy	7	Input, Output, Bi-Dir	User Mode: PLxxx: GPIO ULC_GPLL: Upper Left GPLL Refclk signal (Only Single Ended) (PLLCK) yyyy: Other possible selectable specific functional
PRxxx/yyyy	1	Input, Output, Bi-Dir	User Mode: PRxxx: GPIO yyyy: Other possible selectable specific functional
Shared VREF Pins			
PBxxx/V_{REF}[3,4,5]_[1-2]/yyyy	3, 4, 5	Input, Output, Bi-Dir	User Mode: PBxxx: GPIO V_{REF}: Reference Voltage for DDR memory function [3,4,5] = Bank [1-2] Up to V_{REF}s for each bank yyyy: Other possible selectable specific functional
Shared ADC Pins			
PBxxx/ADC_C[P,N]nn/yyyy	3, 4, 5	Input, Output, Bi-Dir	User Mode: PBxxx: GPIO ADC_C: ADC Channel Inputs [P,N] = Positive or Negative Input nn = ADC Channel number (0 – 15) yyyy: Other possible selectable specific functional
Shared Comparator Pins			
PBxxx/COMP[1-3][P,N]/yyyy	3, 5	Input, Output, Bi-Dir	User Mode: PBxxx: GPIO COMP: Differential Comparator Input [P,N] = Positive or Negative Input [1-3] = Input to Comparators 1-3 yyyy: Other possible selectable specific functional
Shared SGMII Pins			
PBxxx/SGMII_RX[P,N] [0-1]/yyyy	3, 5	Input, Output, Bi-Dir	User Mode: PBxxx: GPIO SGMII_RX: Differential SGMII RX Inputs [P,N] = Positive or Negative Input [0-1] = Input to SGMII RX0 or RX1 yyyy: Other possible selectable specific functional

***Note:** Not all signals are available as external pins in all packages. Refer to the Pinout List file for various package details.

Pin Information Summary

Certus™-NX-RT Family

Pin Information Summary		UT24C407 caBGA256
User I/O Pins		
General Purpose Inputs/Outputs per Bank	Bank 0	12
	Bank 1	21
	Bank 2	28
	Bank 3	32
	Bank 4	32
	Bank 5	10
	Bank 6	28
	Bank 7	22
Total Single-Ended User I/O		185
Differential Input / Output Pairs	Bank 0	0
	Bank 1	0
	Bank 2	0
	Bank 3	32
	Bank 4	32
	Bank 5	10
	Bank 6	0
	Bank 7	0
Total Differential I/O		74
Power Pins		
V_{CC}, V_{CCECLK}		5
V_{CCAUXA}		1
V_{CCAUX}		2
$V_{CCAUXHx}$		3
$V_{CCAUXSD}$		1
V_{CCIO}	Bank 0	1
	Bank 1	1
	Bank 2	1
	Bank 3	1
	Bank 4	1
	Bank 5	1
	Bank 6	1
	Bank 7	1
V_{CCSD0}		1
$V_{CCPLLSD0}$		1
$V_{CCADC18}$		1
Total Power Pins		23

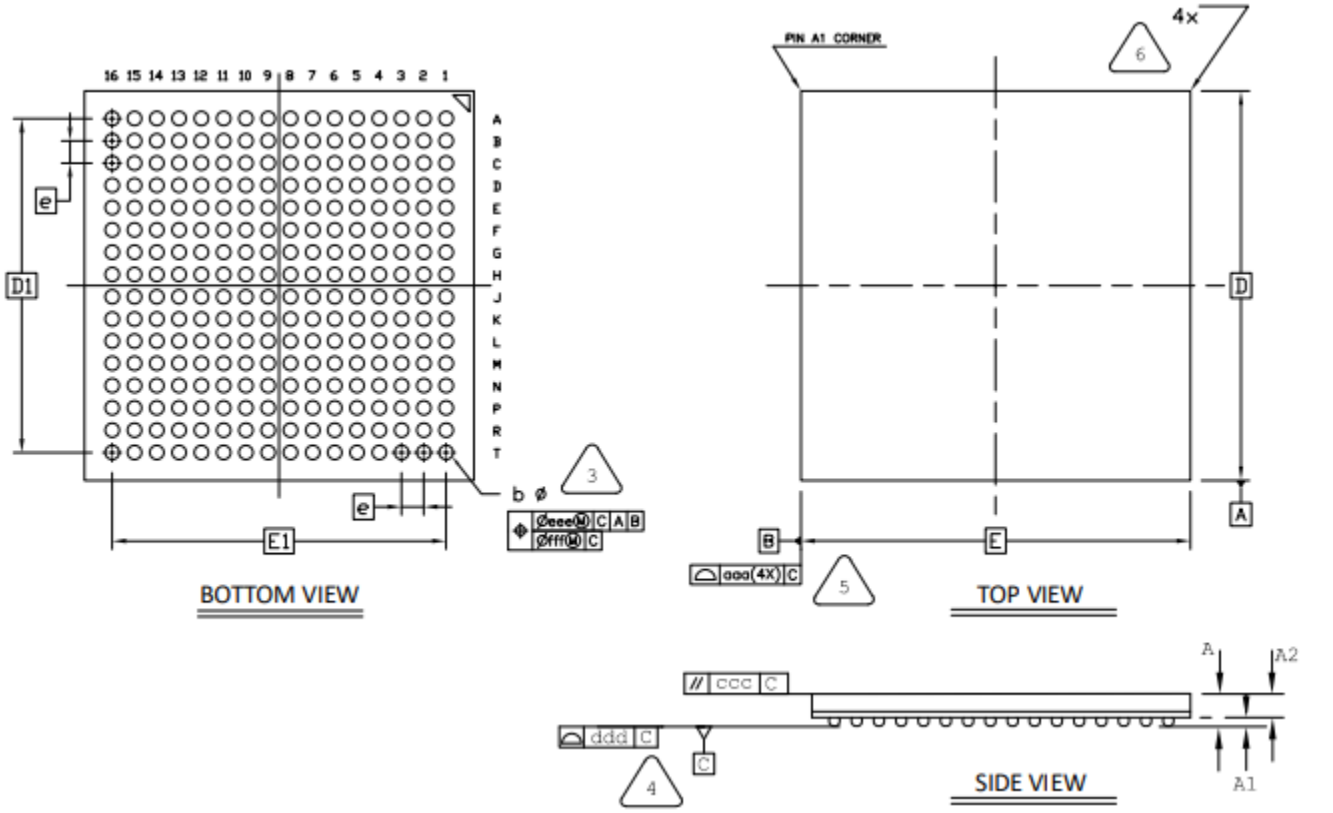
GND Pins		
V_{SS}		27
V_{SSADC}		1
V_{SSSD}		5
Total GND Pins		33
Dedicated Pins		
Dedicated ADC Channels (pairs)		2
Dedicated ADC Reference Voltage Pins		2
Dedicated Misc Pins		
JTAGEN		1
NC		—
RESERVED		—
Total Dedicated Pins		1
Shared Pins		
Shared Configuration Pins	Bank 0	10
	Bank 1	6
	Bank 2	0
	Bank 3	0
	Bank 4	0
	Bank 5	0
	Bank 6	0
	Bank 7	0
Shared JTAG Pins	Bank 0	0
	Bank 1	4
	Bank 2	0
	Bank 3	0
	Bank 4	0
	Bank 5	0
	Bank 6	0
	Bank 7	0
Shared PCLK Pins	Bank 0	2
	Bank 1	3
	Bank 2	3
	Bank 3	8
	Bank 4	8
	Bank 5	8
	Bank 6	3
	Bank 7	3
Shared GPLL Pins	Bank 0	0
	Bank 1	0

	Bank 2	0
	Bank 3	2
	Bank 4	0
	Bank 5	2
	Bank 6	0
	Bank 7	2
Shared V_{REF} Pins	Bank 0	0
	Bank 1	0
	Bank 2	0
	Bank 3	2
	Bank 4	2
	Bank 5	2
	Bank 6	0
	Bank 7	0
Shared ADC Channels (pairs)	Bank 0	0
	Bank 1	0
	Bank 2	0
	Bank 3	12
	Bank 4	0
	Bank 5	4
	Bank 6	0
	Bank 7	0
Shared Comparator Channels (pairs)	Bank 0	0
	Bank 1	0
	Bank 2	0
	Bank 3	3
	Bank 4	0
	Bank 5	3
	Bank 6	0
	Bank 7	0

Shared SGMII Channels (pairs)	Bank 0	0
	Bank 1	0
	Bank 2	0
	Bank 3	0
	Bank 4	0
	Bank 5	1
	Bank 6	0
	Bank 7	0

Package Information

256-Ball caBGA Package

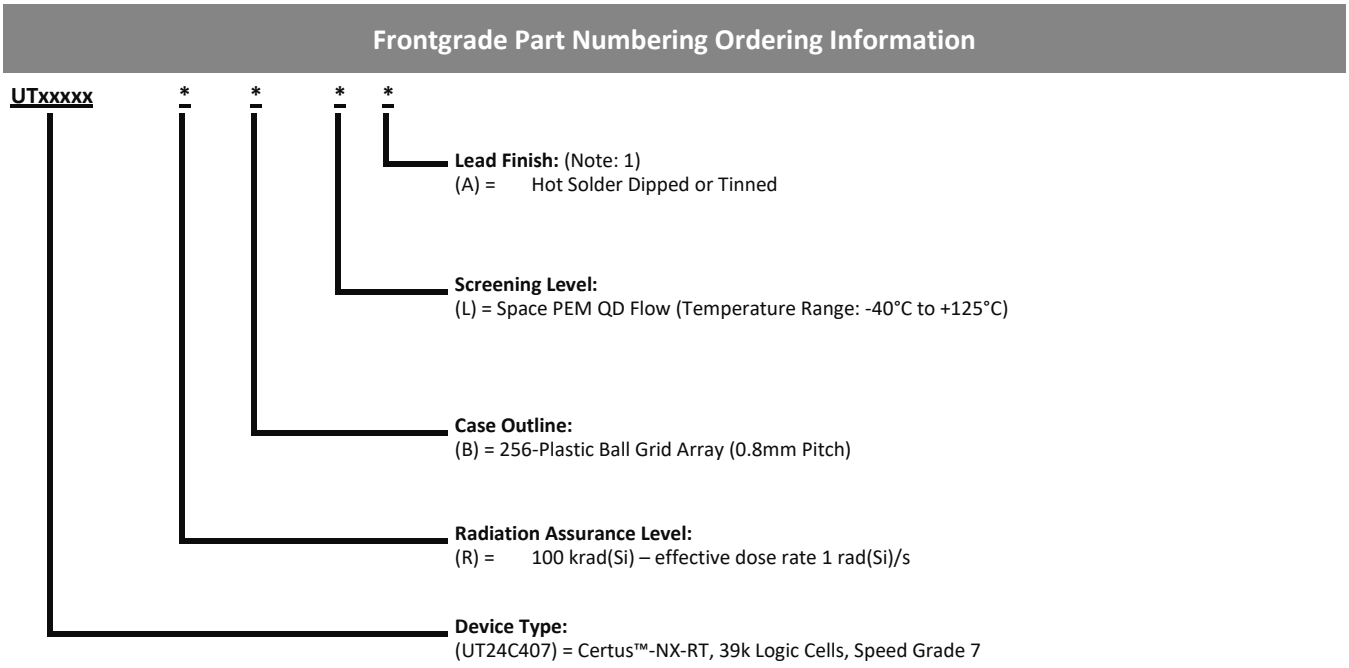


Notes:

1. Dimensions and tolerances per ANSI Y14.5M
2. All dimensions are in millimeters
3. Dimension “b” is measured at the maximum solder ball diameter, parallel to primary datum Ⓒ
4. Primary datum Ⓒ and seating plane are defined by the spherical crowns of the solder balls
5. Bilateral tolerance zone is applied to each side of the package body
6. Exact shape and size of this feature is optional
7. Reference JEDEC MO-275A_01.

Symbol	Min.	Nom.	Max.
A	-	-	1.70
A1	0.25	-	-
A2	-	-	1.45
D/E	14.0 BSC		
D1/E1	12.0 BSC		
n	256		
b	0.45	0.50	0.55
e	0.80		
aaa	0.15		
ccc	0.20		
ddd	0.20		
eee	0.15		
fff	0.08		

Ordering Information



NOTES:

1. Space PEM QD devices only

References

For more information, refer to the following documents:

- sysCLOCK PLL Design and Usage Guide for Nexus Platform (FPGA-TN-02095)
- sysDSP Usage Guide for Nexus Platform (FPGA-TN-02096)
- sysCONFIG Usage Guide for Nexus Platform (FPGA-TN-02099)
- sysI/O Usage Guide for Nexus Platform (FPGA-TN-02067)
- Soft Error Detection (SED)/Correction (SEC) Usage Guide for Nexus Platform (FPGA-TN-02076)
- Memory Usage Guide for Nexus Platform (FPGA-TN-02094)
- ADC Usage Guides for Nexus Platform (FPGA-TN-02129)
- Certus™-NX-RT High-Speed I/O Interface (FPGA-TN-02216)
- Power Management and Calculation for Certus™-NX-RT Devices (FPGA-TN-02214)
- Certus™-NX-RT 40K Pinout File (FPGA-SC-02004)
- Lattice Memory Mapped Interface and Lattice Interrupt Interface User Guide (FPGA-UG-02039)
- sub-LVDS Signaling Using Lattice Devices (FPGA-TN-02028)
- Multi-Boot Usage Guide for Nexus Platform (FPGA-TN-02145)
- TransFR Usage Guide for Nexus Platform (FPGA-TN-02173)
- I2C Hardened IP Usage Guide for Nexus Platform (FPGA-TN-02142)

For package information, refer to the following documents:

- PCB Layout Recommendations for BGA Packages (FPGA-TN-02024)
- Solder Reflow Guide for Surface Mount Devices (FPGA-TN-02041)
- Thermal Management (FPGA-TN-02044)
- Package Diagrams (FPGA-DS-02053)
- High Speed PCB Design Considerations (FPGA-TN-02148)
- Advanced Configuration Security Usage Guide for Nexus Platform (FPGA-TN-02176)
- Hardware Checklist (FPGA-TN-02151)

For further information on interface standards, refer to the following websites:

- JEDEC Standards (LVTTL, LVCMOS, SSTL) – www.jedec.org
- PCI – www.pcisig.com

Revision History

Date	Revision #	Author	Change Description	Page #
01/26/2022	2.0.0		Initial Release	
02/08/2022	2.0.1		Added Operational Environment section, fixed broken document references, fixed image quality and header formatting.	
04/26/2022	2.0.2		Removed prototype characters from Frontgrade Part Number Ordering Information section; Added Certus-NX Part Number Ordering Information section for ordering prototype parts	
02/07/2023	2.0.3		Minor changes to Operational Environment section.	
04/14/2023	2.0.4		Section 6.2 removed for eliminating proto offering	
1/18/2024	2.0.5		JGB: Pin Information Summary section; changed table to show correct number of dedicated pins for ADC channels. The ADC is available and those two entries should read: • Dedicated ADC Channels (pairs) 2 • Dedicated ADC Reference Voltage pins 2 • Remove all references to Lattice part number LFD2NX-40, replaced with UT24C407.	
3/25/2024	2.0.6	JGB	Additions and corrections to Table 3.2: 1. spelling on irradiation in note 1 2. Add note to indicate SEL tested at 125C	
12/13/2024	2.0.7	BZB	Corrected package drawing information to point to latest Lattice drawing Updated ordering information from Space Automotive to Space PEM QD	

Datasheet Definitions

	Definition
Advanced Datasheet	Frontgrade reserves the right to make changes to any products and services described herein at any time without notice. The product is still in the development stage and the **datasheet is subject to change**. Specifications can be **TBD** and the part package and pinout are **not final.**
Preliminary Datasheet	Frontgrade reserves the right to make changes to any products and services described herein at any time without notice. The product is in the characterization stage and prototypes are available.
Datasheet	Product is in production and any changes to the product and services described herein will follow a formal customer notification process for form, fit or function changes.